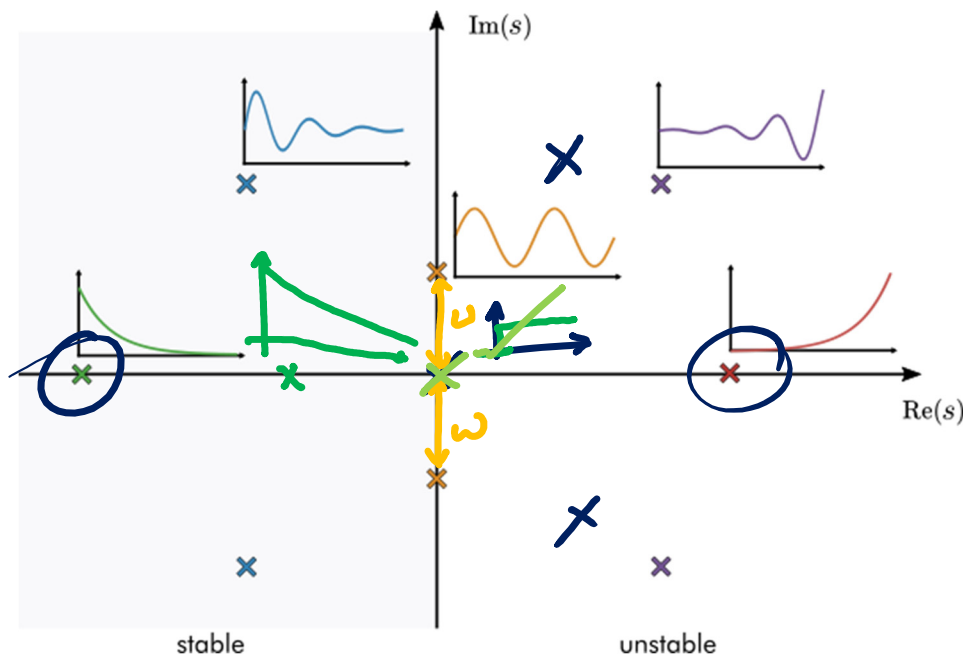


Poles-Zero Plot



Takeaways

- complex poles/zeros always show up in conjugate pairs (required for real-valued time-domain function)

- If all poles in open LHP system is stable / signal is bounded

- If all poles in open LHP $j\omega$ -axis is in region of convergence (ch 15 - frequency response)

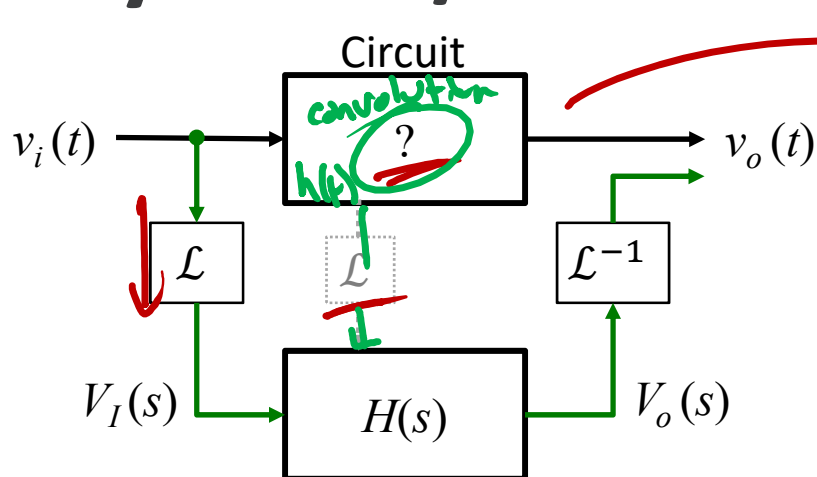
If $F(s)$ has poles at $-\sigma \pm j\omega$

$$F(s) = \frac{N(s)}{\dots (s - (-\sigma + j\omega)) (s - (-\sigma - j\omega)) \dots}$$

$$F(s) = \dots + \frac{k}{s - (-\sigma + j\omega)} + \frac{k^*}{s - (-\sigma - j\omega)} + \dots$$

$$f(t) = \dots + 2|k| e^{-\sigma t} \cos(\omega t - \Delta/k) + \dots$$

System I/O Relationship



201 approach: solve ODEs

202: $V_I(s) = \mathcal{L}\{v_i(t)\}$

$V_O(s) = H(s)V_I(s)$

$v_o(t) = \mathcal{L}^{-1}\{V_O(s)\} = \mathcal{L}^{-1}\{H(s)V_I(s)\}$

$v_o(t) = \mathcal{L}^{-1}\{H(s)V_I(s)\}$

$\int_0^{\infty} h(t-\tau)w_i(\tau)d\tau$ → property for Laplace transform of a product of two s-domain functions?

$\mathcal{L}^{-1}\{V_I(s)\} = v_i(t)$

$\mathcal{L}^{-1}\{H(s)\} = h(t)$

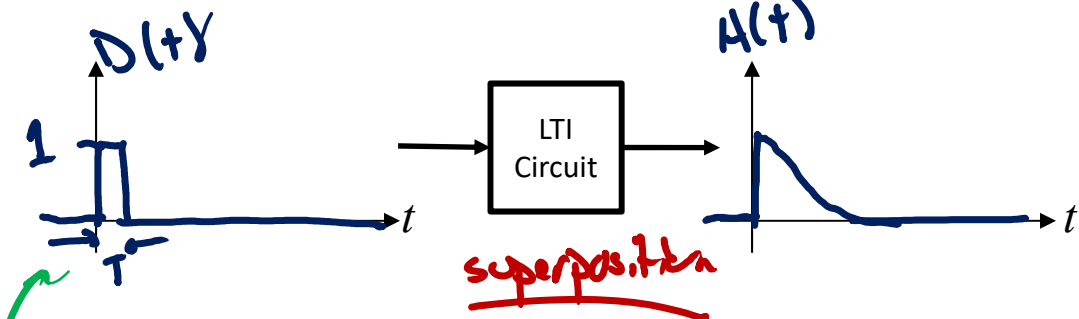
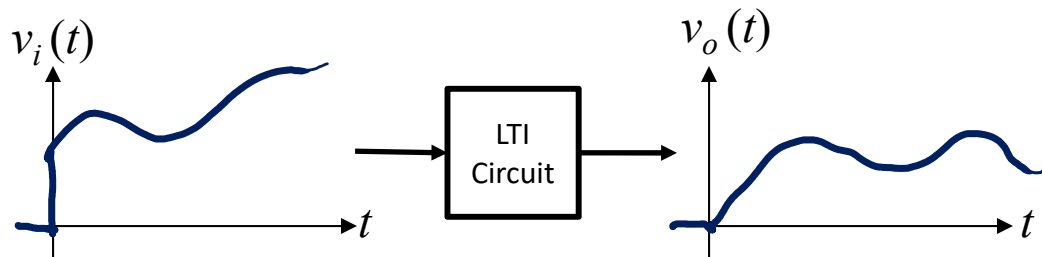
if $V_I(s) = 1 \iff v_i(t) = \delta(t)$

then

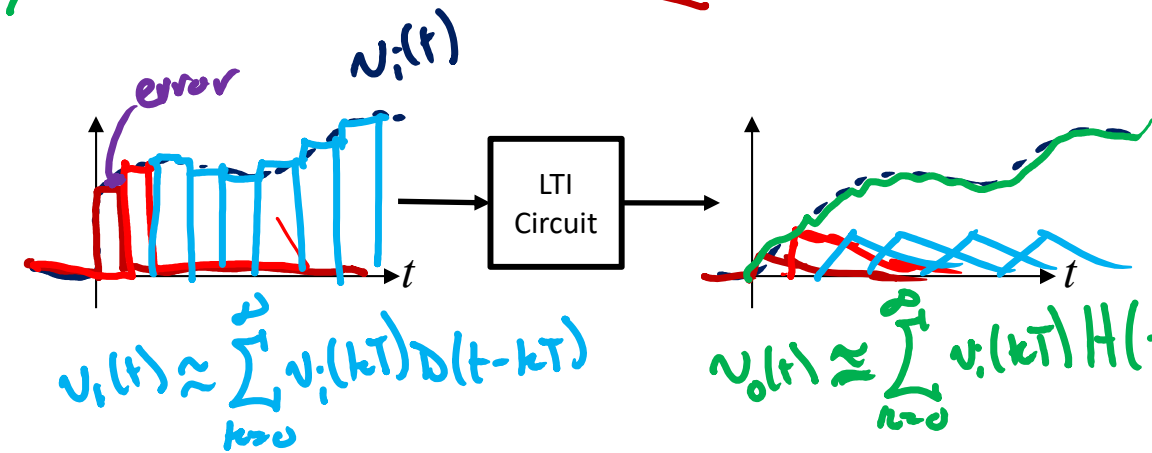
$v_o(t) = \mathcal{L}^{-1}\{H(s) \cdot 1\} = \mathcal{L}^{-1}\{H(s)\}$

$\mathcal{L}^{-1}\{H(s)\} = h(t) \equiv \text{"impulse response"}$

Convolution



superposition



as $T \rightarrow 0$

$$v_o(t) = \int_0^{\infty} v_i(\tau) \delta(t-\tau) d\tau$$

sifting property of $\delta(t)$

$$v_o(t) = \int_0^{\infty} v_i(\tau) h(t-\tau) d\tau$$

$$= \int_0^{\infty} v_i(t-\tau) h(\tau) d\tau$$

Convolution integral

(formally $\int_{-\infty}^{\infty} v_i(\tau) h(t-\tau) d\tau$
but with unilateral Laplace
zero for $t < 0$)

The Convolution Integral

Laplace Transform property:

$$V_o(s) = V_i(s) H(s)$$



$$v_o(t) = \int_0^{\infty} h(t-\tau) v_i(\tau) d\tau = h(t) * v_i(t)$$

short-hand
↓

$$v_o(t) = \int_0^{\infty} h(-\tau+t) v_i(\tau) d\tau$$

flip
↑
shift
at $t=2$

$$v_o(t=2) = \int_0^{\infty} h(-\tau+2) v_i(\tau) d\tau$$

