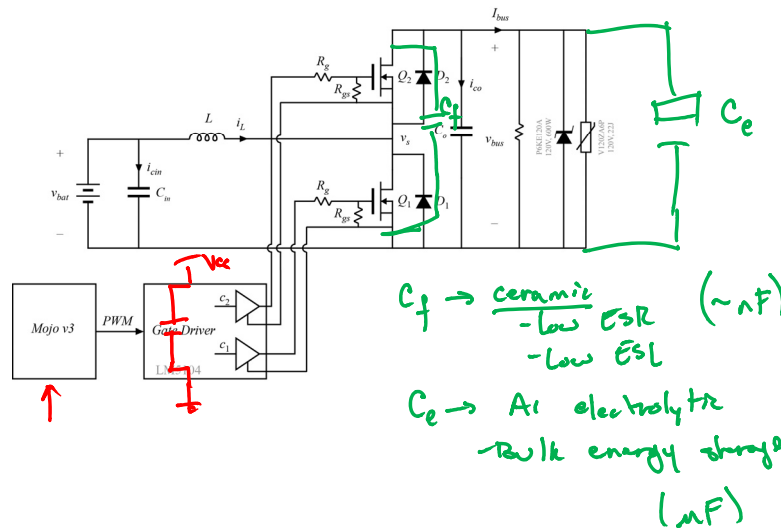




## Experiment 3: Open Loop Boost



## Power Converter Layout: Buck Example

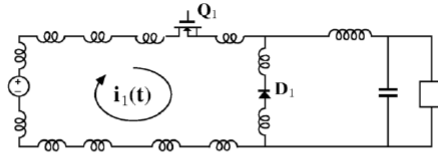
Use loop analysis

$$V_p = L_p \frac{di_1}{dt}$$

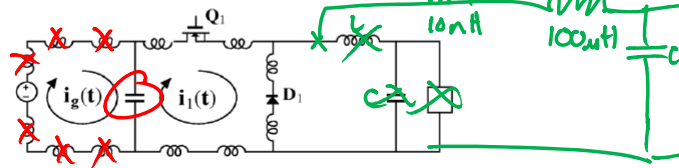
switched input current  $i_1(t)$  contains large high frequency harmonics  
 —hence inductance of input loop is critical  
 inductance causes ringing, voltage spikes, switching loss, generation of B- and E-fields, radiated EMI  
 the second loop contains a filter inductor, and hence its current  $i_2(t)$  is nearly dc  
 —hence additional inductance is not a significant problem in the second loop



Parasitic inductances of input loop explicitly shown:



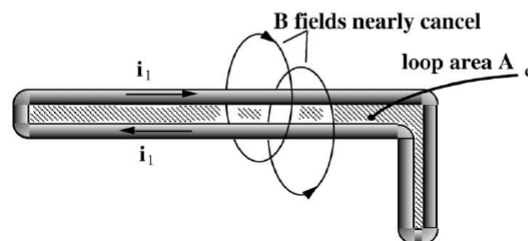
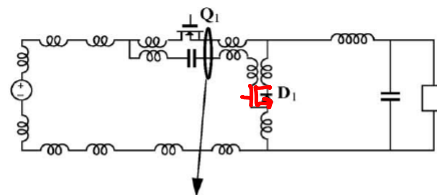
Addition of bypass capacitor confines the pulsating current to a smaller loop:

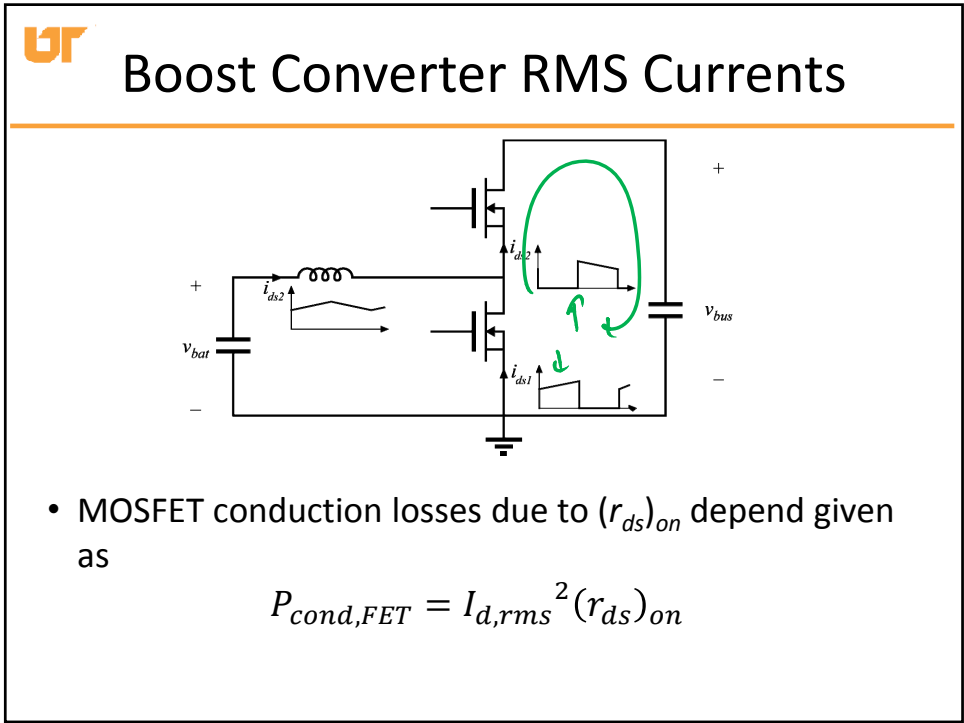
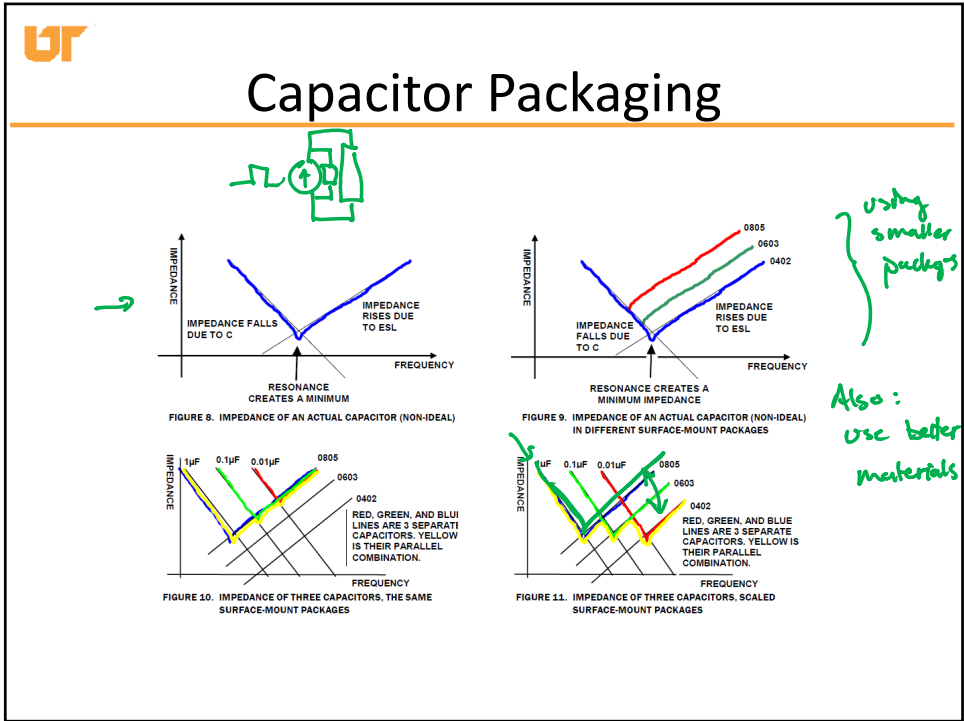


high frequency currents are shunted through capacitor instead of input source



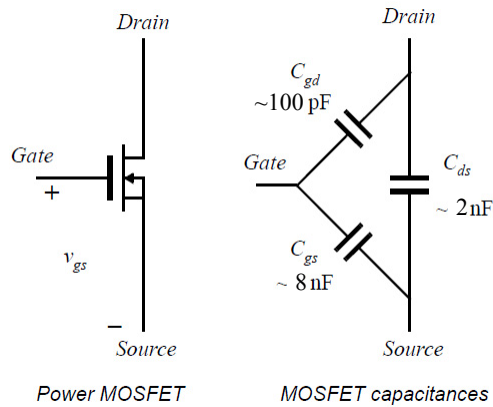
Even better: minimize area of the high frequency loop, thereby minimizing its inductance







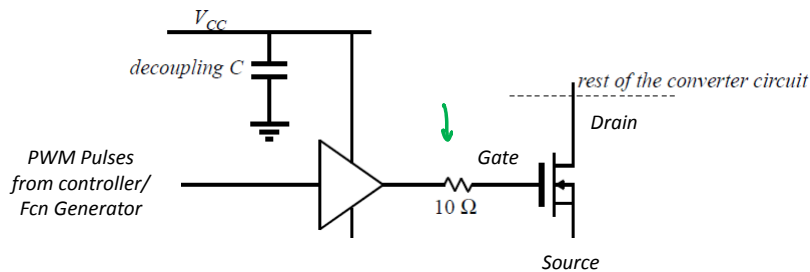
## Driving a Power MOSFET Switch



- MOSFET is off when  $v_{gs} < V_{th} \approx 3 \text{ V}$
- MOSFET fully on when  $v_{gs}$  is sufficiently large (10-15 V)
- Warning: MOSFET gate oxide breaks down and the device fails when  $v_{gs} > 20 \text{ V}$ .
- Fast turn on or turn off (10's of ns) requires a large spike (1-2 A) of gate current to charge or discharge the gate capacitance
- MOSFET gate driver is a logic buffer that has high output current capability



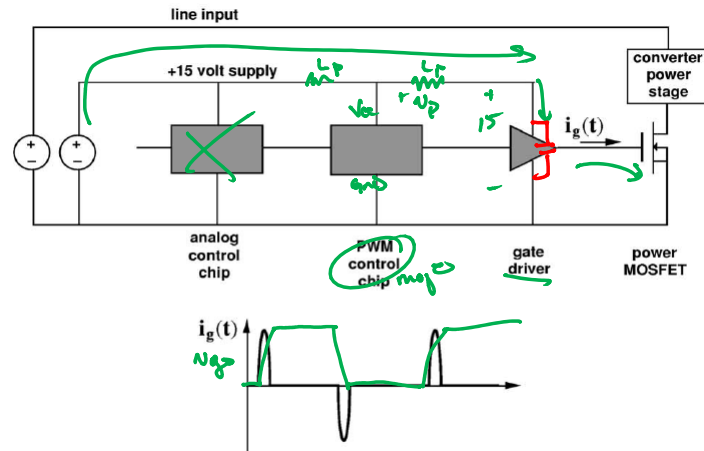
## Driving a Power MOSFET Switch



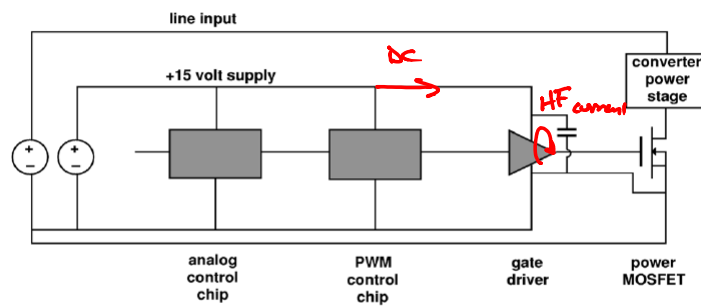
- MOSFET gate driver is used as a logic buffer with high output current ( $\sim 1.8 \text{ A}$ ) capability
- The amplitude of the gate voltage equals the supply voltage  $V_{CC}$
- Decoupling capacitors are necessary at all supply pins of LM5104 (and all ICs)
- Gate resistance used to slow  $dv/dt$  at switch node



## Gate Driver Example



### Solution: bypass capacitor and close coupling of gate and return leads

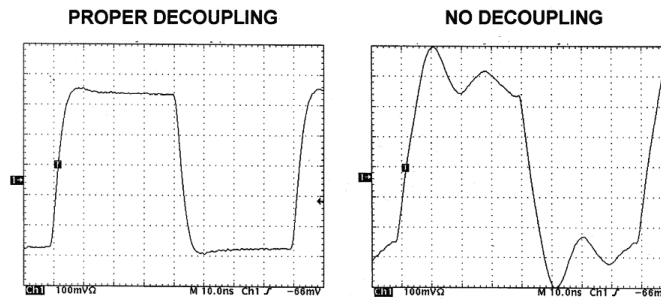


High frequency components of gate drive current are confined to a small loop

A dc component of current is still drawn output of 15V supply, and flows past the control chips. Hence, return conductor size must be sufficiently large



## Op-amp Pulsed Decoupling

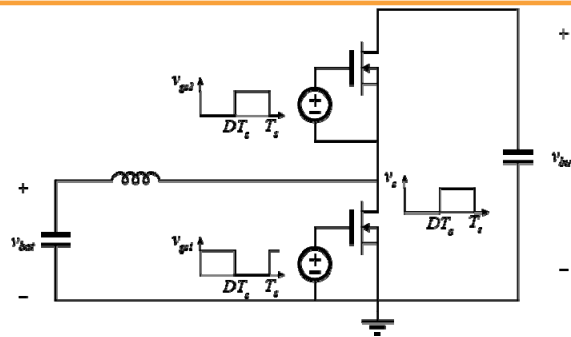


VERTICAL SCALE: 100mV/div  
HORIZONTAL SCALE: 10ns/div

Figure 12.68: Effects of Inadequate Decoupling on the Phase Response of the AD9631 Op Amp



## Half Bridge Gate Drive Waveforms

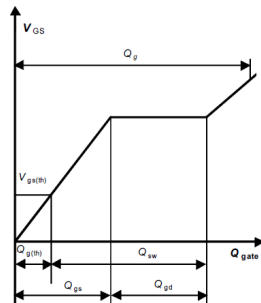


- Gate driver chip must implement  $v_{gs}$  waveforms
- Sources will have pulsating currents and need decoupling



## MOSFET Gate Charge

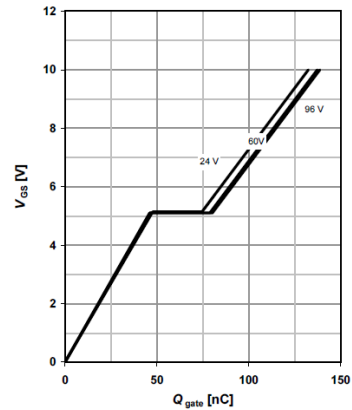
- Charge is supplied to both  $C_{gs}$  and  $C_{gd}$  in order to move gate voltage and switch MOSFET
- Would like to supply the charge in minimum time to quickly switch FET
- Results in high peak currents



14 Typ. gate charge

$$V_{GS} = f(Q_{gate}); I_D = 100 \text{ A pulsed}$$

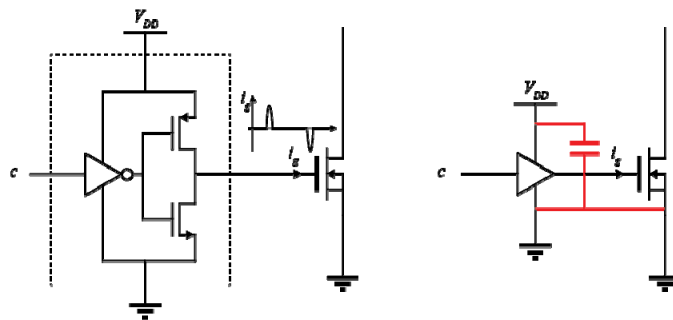
parameter:  $V_{DD}$



16 Gate charge waveforms



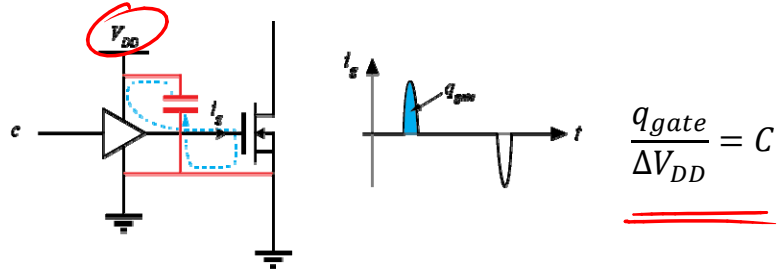
## Gate Drive Implementation



- Gate driver is cascades back half-bridges of decreasing size to obtain quick rise times
- Reminder: keep loops which handle pulsating current small by decoupling and making close connections



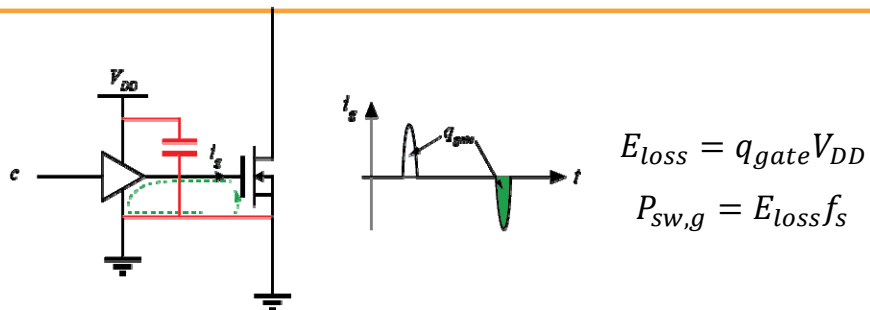
## Capacitor Sizing Notes



- Area of current pulse is total charge supplied to gate of capacitor
- All charge must be supplied from gate drive decoupling capacitor



## Gate Drive Losses

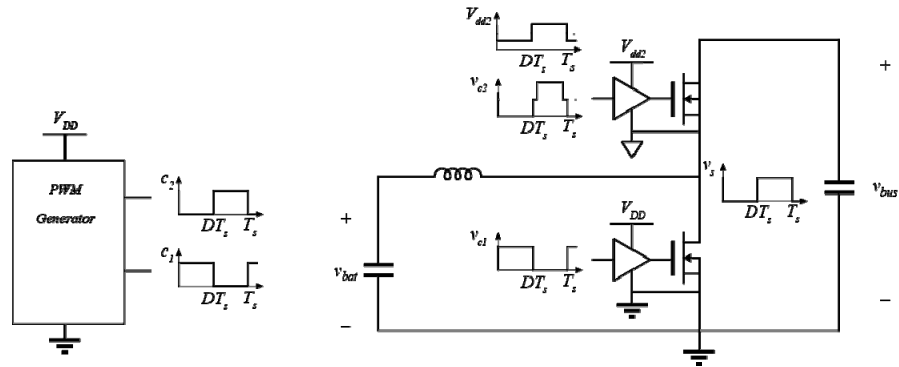


- Gate charge is supplied through driver resistance during switch turn-on
- Gate charge is dissipated in gate driver on switch turn-off





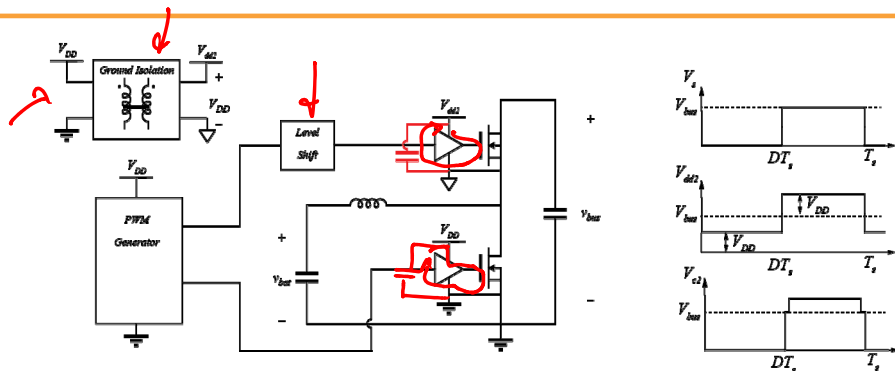
## High Side Signal Ground



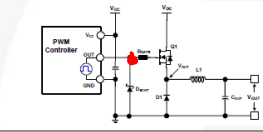
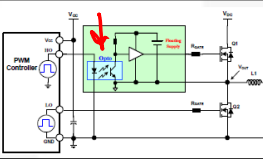
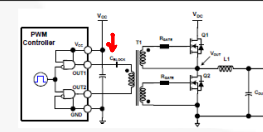
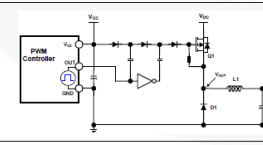
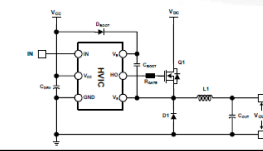
- Gate driver chip must implement  $v_{gs}$  waveforms
- Issue: source of  $Q_2$  is not grounded




## Generating Floating Supply



- Isolated supplies sometimes used; Isolated DC-DC, batteries
- Bootstrap concept: capacitor can be charged when  $V_s$  is low, then switched

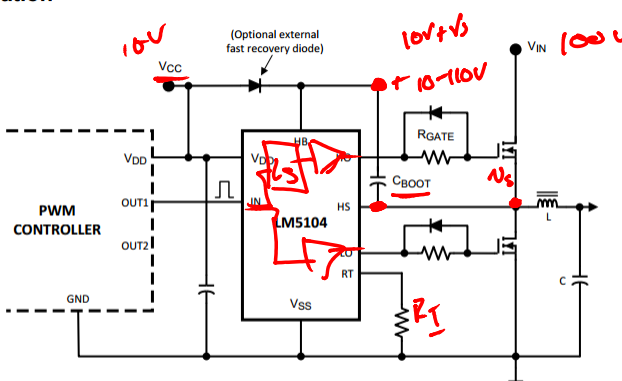
<p>Direct Drive</p>		<p>Easiest high-side application the MOSFET and can be driven directly by the PWM controller or by a ground referenced driver, but it must meet two conditions, as follows:</p> $V_{CC} < V_{GS,MAX} \text{ and } V_{DC} < V_{CC} - V_{GS,Miller}$
<p>Floating Supply Gate Drive</p>		<p>Cost impact of isolated supply is significant. Opto-coupler tends to be relatively expensive, limited in bandwidth, and noise sensitive.</p>
<p>Transformer Coupled Drive</p>		<p>Gives full gate control for an indefinite period of time, but is somewhat limited in switching performance. This can be improved with added complexity.</p>
<p>Charge Pump Drive</p>		<p>The turn-on times tend to be long for switching applications. Inefficiencies in the voltage multiplication circuit may require more than low stages of pumping.</p>
<p>Bootstrap Drive</p>		<p>Simple and inexpensive with limitations; such as, the duty cycle and on-time are both constrained by the need to refresh the bootstrap capacitor. Requires level shift, with the associated difficulties.</p> <p style="text-align: right;"><i>Fairchild Semi App Note AN-6076</i></p>



# LM5104 Gate Driver

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## 8.2 Typical Application

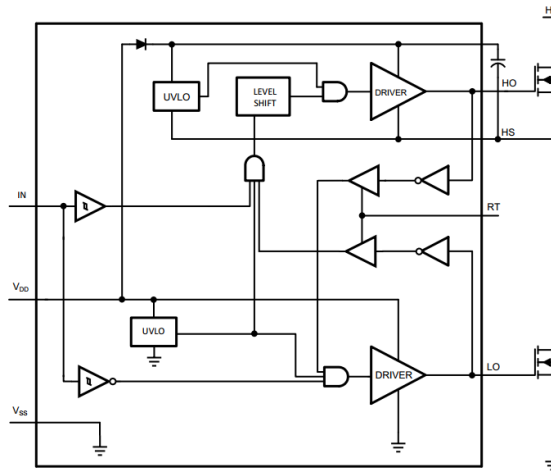


**Figure 15. LM5104 Driving MOSFETs Connected in Synchronous Buck Configuration**



## LM5104 Internal Block Diagram

### 7.2 Functional Block Diagram



## Bootstrap Diode Loss

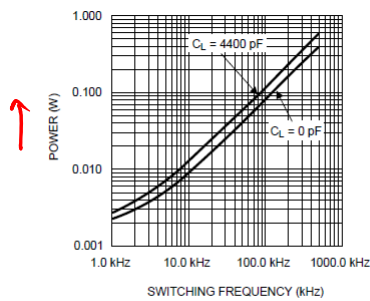


Figure 5. Diode Power Dissipation  $V_{IN} = 80V$

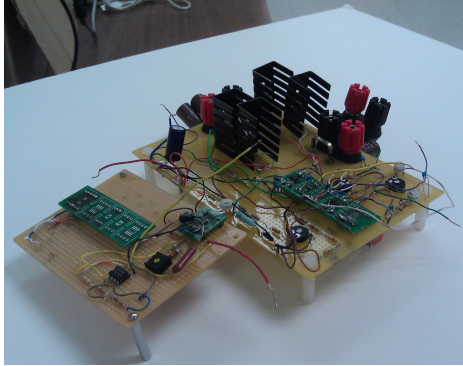
- ~~Conduction losses~~ due to pulsating currents are relatively small
- Switching losses are significant
- Diode capacitance and reverse recovery play a role



## Wiring Advice

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- Take the time to wire board cleanly



- Very difficult to find a short in sloppy-wired circuit