

# Averaging Step

$$n_t \langle i_{out} \rangle = \frac{2}{T_s} \int_0^{T_s/2} n_t i_{out}(t) dt$$

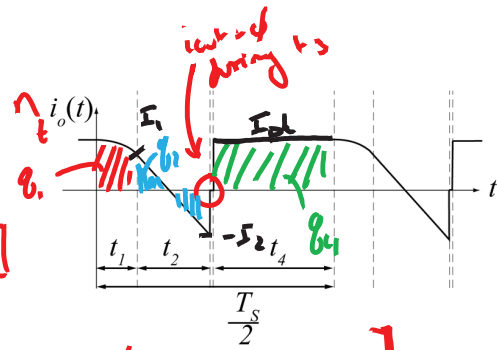
$$n_t \langle i_{out} \rangle = \frac{2}{T_s} [\theta_1 + \theta_2 + \theta_3 + \theta_4]$$

$$= \frac{2}{T_s} \left[ C_p(2V_g) + t_2 \frac{I_1 - I_2}{2} + \phi + t_4 I_{pk} \right]$$

$$J = \frac{n_t \langle i_{out} \rangle}{I_{base}} = \frac{2}{T_s} \left[ \frac{2C_p V_g}{V_g} \sqrt{\frac{L I_{pk}}{C_p}} + t_2 \frac{J_1 - J_2}{2} + t_4 J_{pk} \right]$$

$$J = \frac{2}{T_s} \frac{1}{\omega_0} \left[ 2 + \theta_2 \frac{J_1 - J_2}{2} + \theta_4 J_{pk} \right]$$

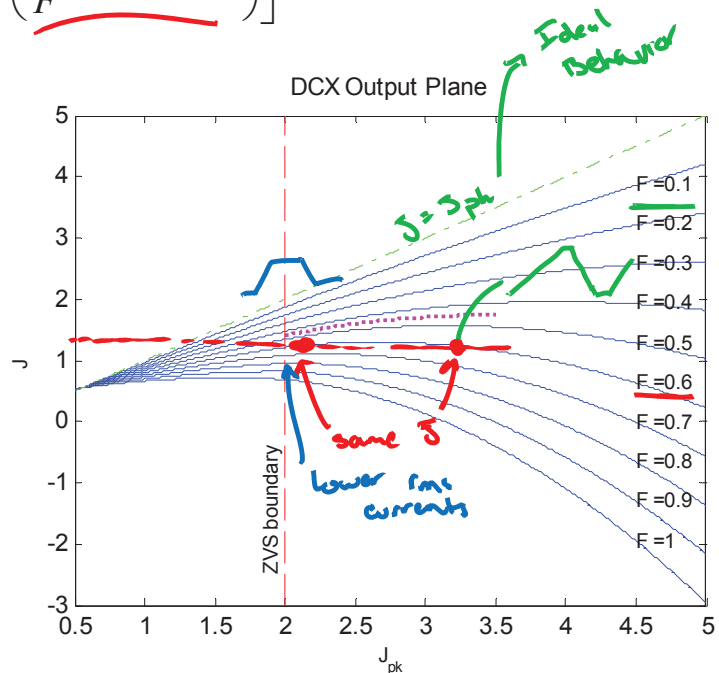
$$\frac{f_s}{2\pi f_0} = \frac{F}{\pi} \rightarrow J = \frac{F}{\pi} \left[ 2 + \theta_2 \frac{J_1 - J_2}{2} + \theta_4 J_{pk} \right]$$



# Output Plane

$$F = \frac{f_s}{f_0}$$

$$J = \frac{n \langle i_{out} \rangle}{I_{base}} = \frac{F}{\pi} \left[ 2 + \frac{1}{4} (J_1^2 - J_2^2) + J_p \left( \frac{\pi}{F} - \alpha - \theta_2 - \delta \right) \right]$$

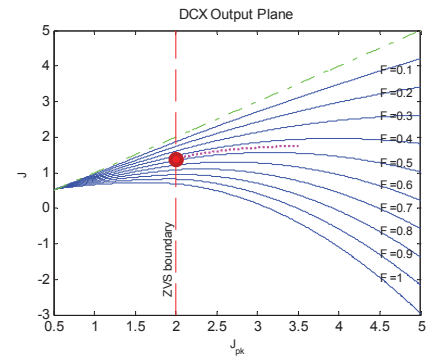
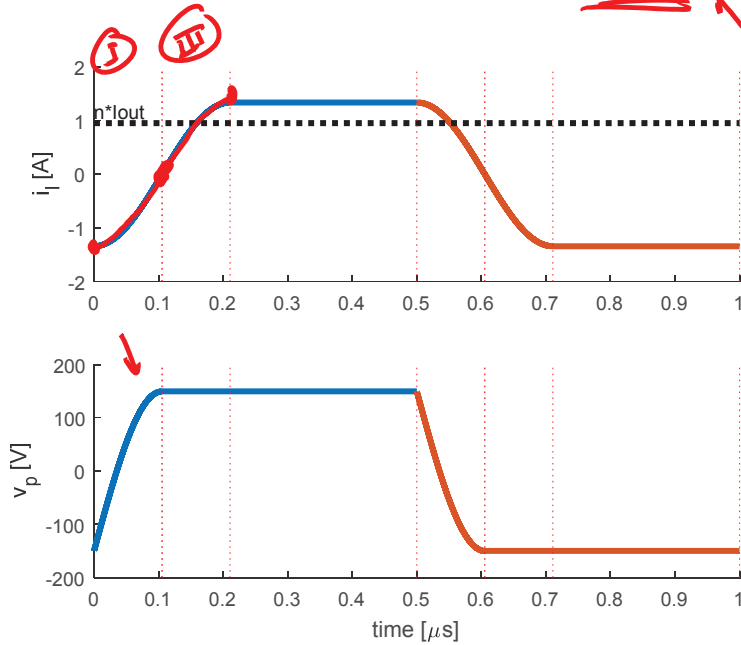


# Example Waveforms

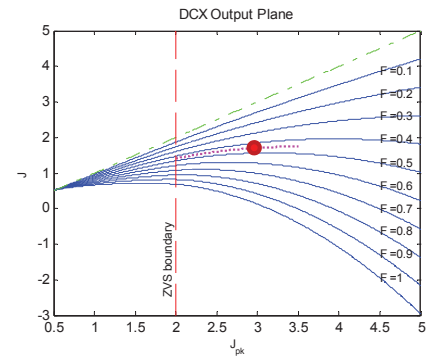
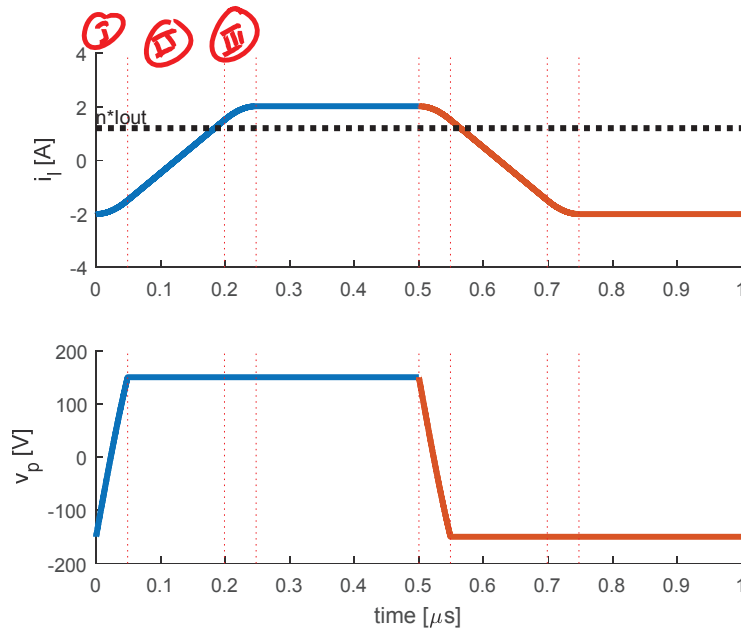
$f_s = 1\text{MHz}$   
 $V_g = 150\text{V}$

$V = 12\text{V}$   
 $L_p = 15\text{mH}$

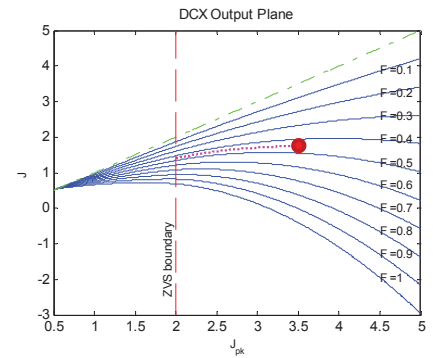
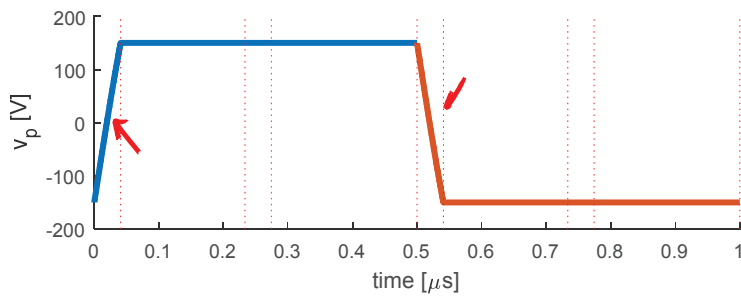
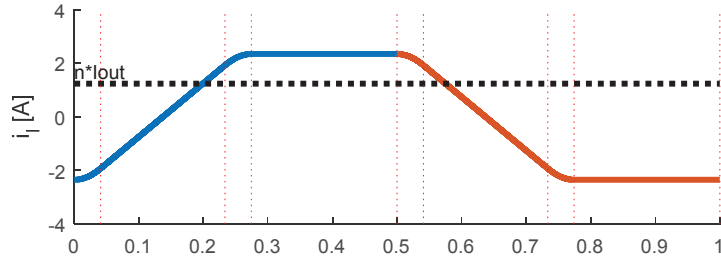
$C_p = 300\text{pF}$   
 $C_s = \frac{L_p}{Z^2}$



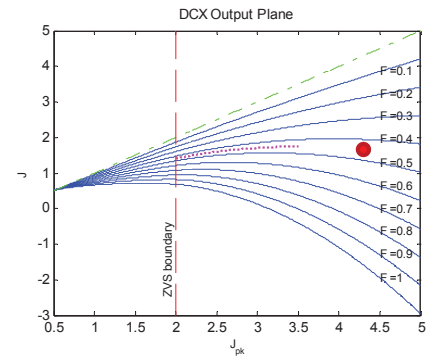
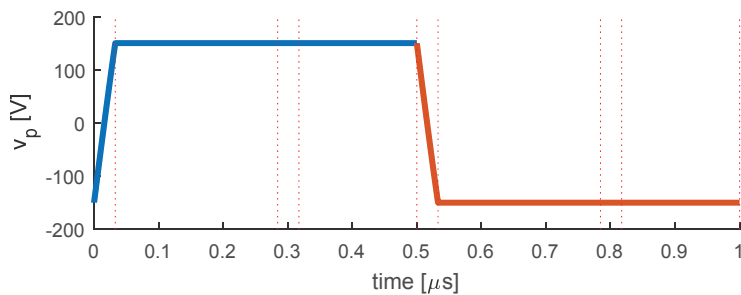
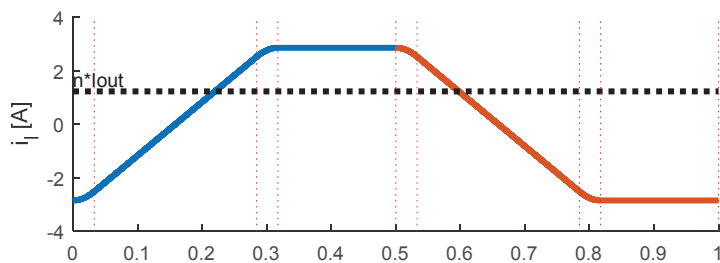
# Example Waveforms



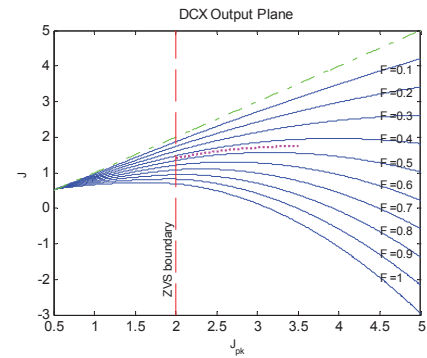
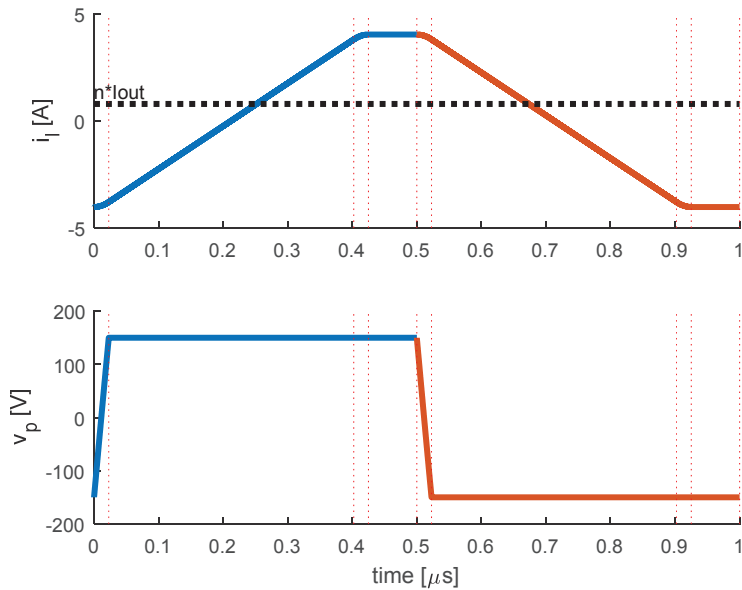
# Example Waveforms



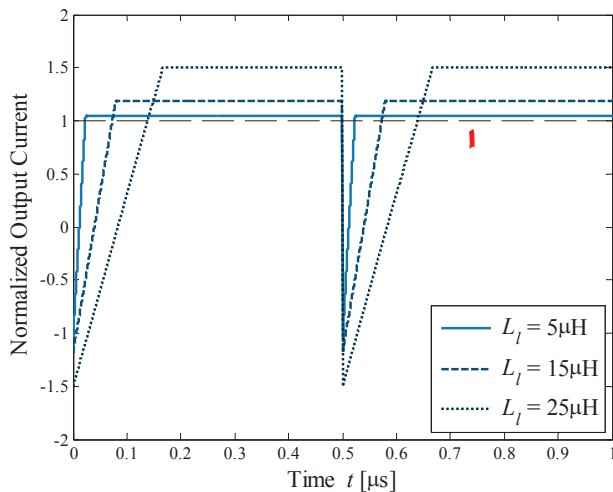
# Example Waveforms



# Example Waveforms

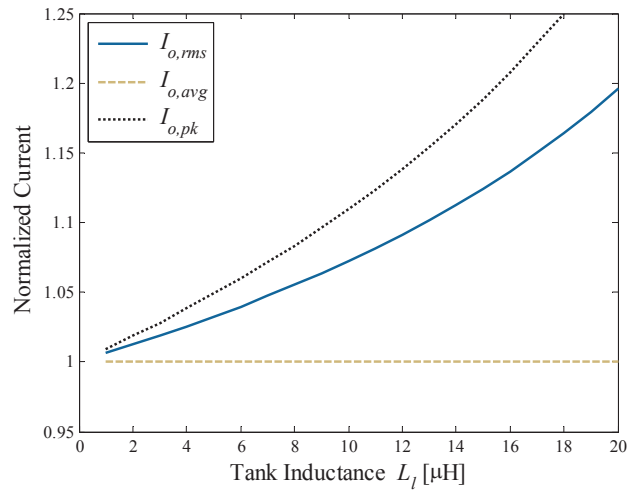
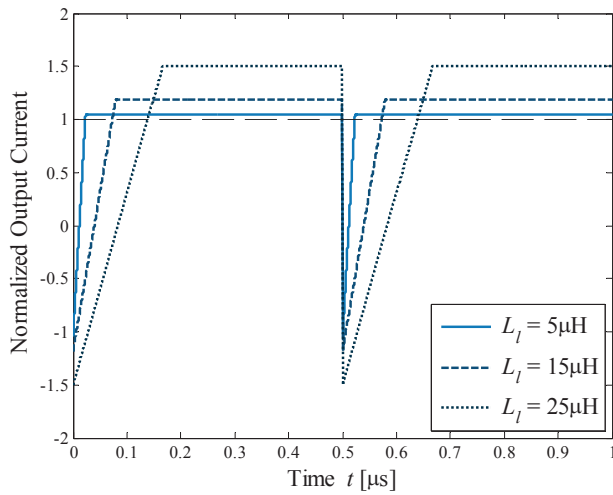


# Output Current Vs. Inductance

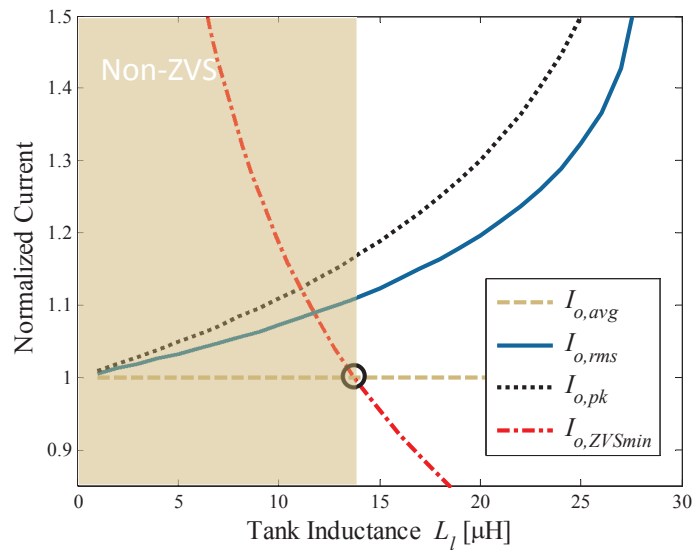


low (small) inductance is best for minimizing conduction losses.

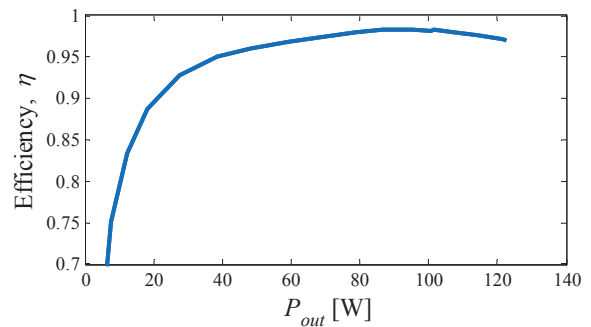
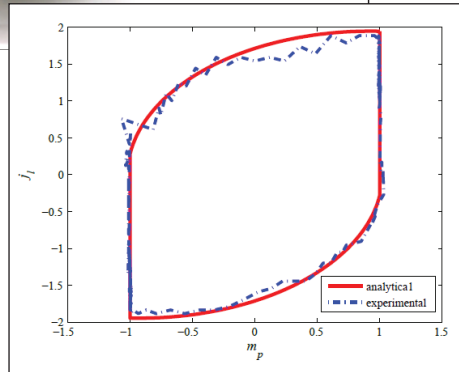
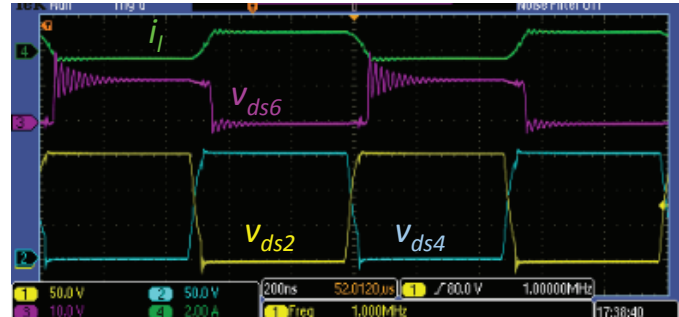
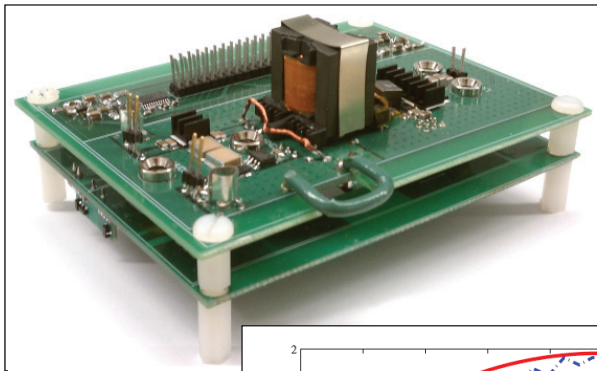
# Output Current Vs. Inductance



# Constraints on Inductance



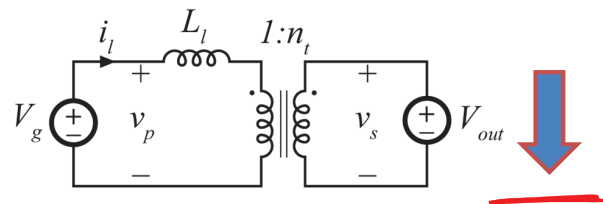
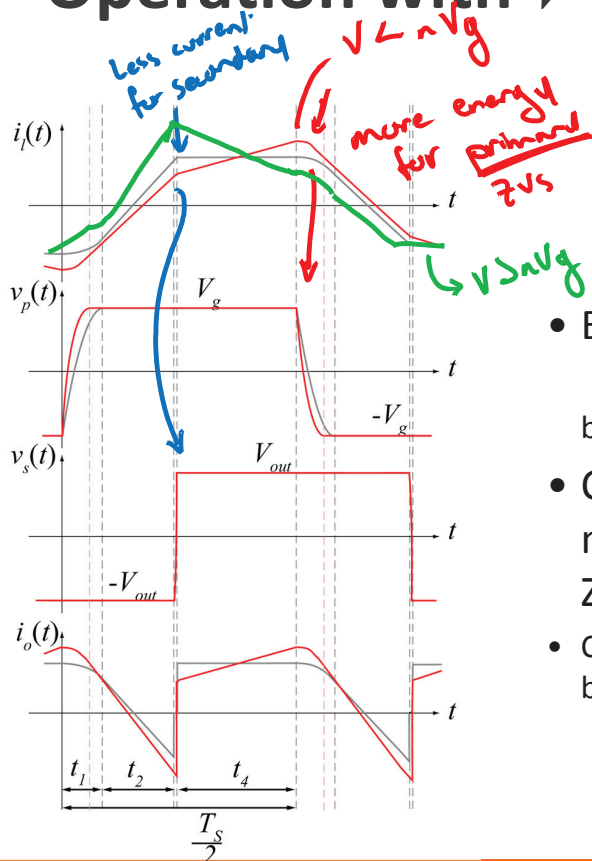
# DAB: Experimental Results



D. Costinett, D. Maksimovic, and R. Zane, "Design and control for high efficiency in high step-down dual active bridge converters operating at high switching frequency," IEEE Trans. Power Electron., vol. PP, no. 99, p. 1, 2012.



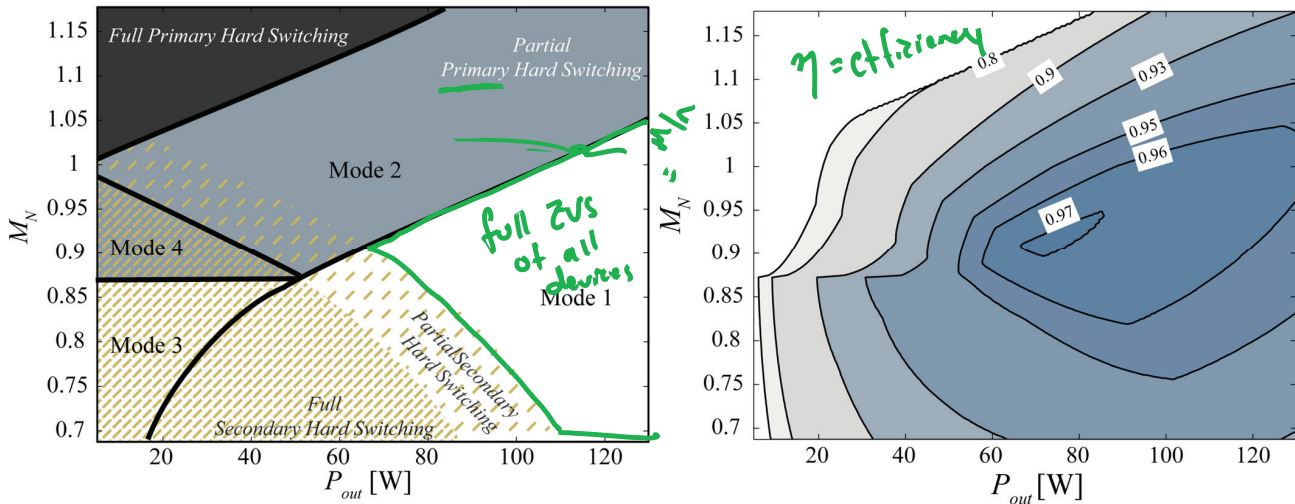
## Operation with $V \neq nV_g$



- E.g. Decrease to  $M_N < 1$  by decreasing output voltage
- Current now ramping, causing more energy available for primary ZVS, but higher RMS currents
- Can use behavior to extend ZVS range of one bridge



# Soft Switching Range with Varying $V_{out}$



D. Costinett, D. Maksimovic, and R. Zane, "Design and control for high efficiency in high step-down dual active bridge converters operating at high switching frequency," IEEE Trans. Power Electron., vol. PP, no. 99, p. 1, 2012.



## Application Example: Automotive

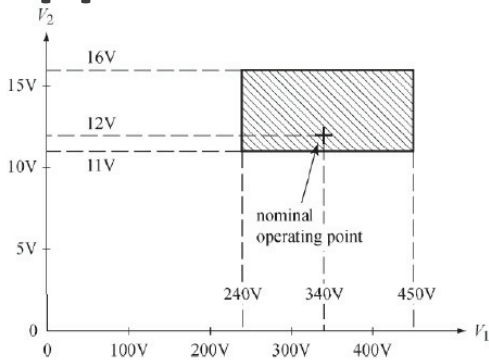


Fig. 1. Converter operating voltage ranges required for automotive application.

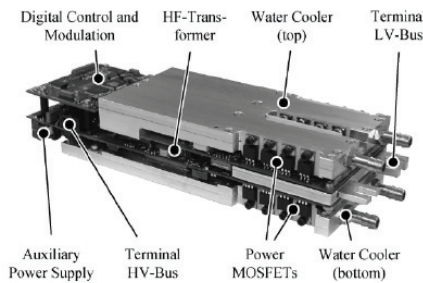
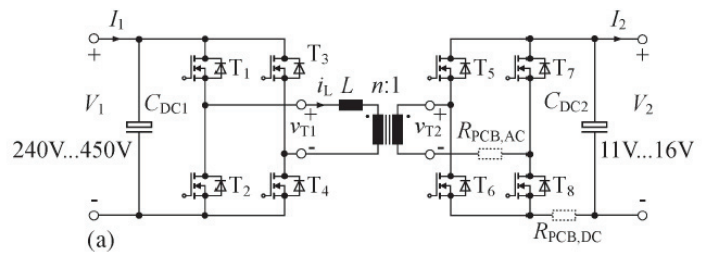


Fig. 3. Automotive DAB converter (273 × 90 × 53 mm).



(a)

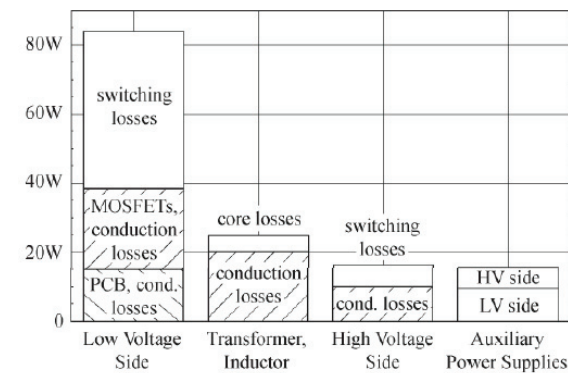


Fig. 13. Calculated distribution of the power losses for operation at  $V_1 = 340$  V,  $V_2 = 12$  V, and  $P_2 = 2$  kW.

\*F. Krismer, J.W.Kolar, "Accurate Power Loss Model Derivation of a High-Current Dual Active Bridge Converter for an Automotive Application, IEEE Trans. On Industrial Electronics, March 2010



# Alternate Modulation Schemes

