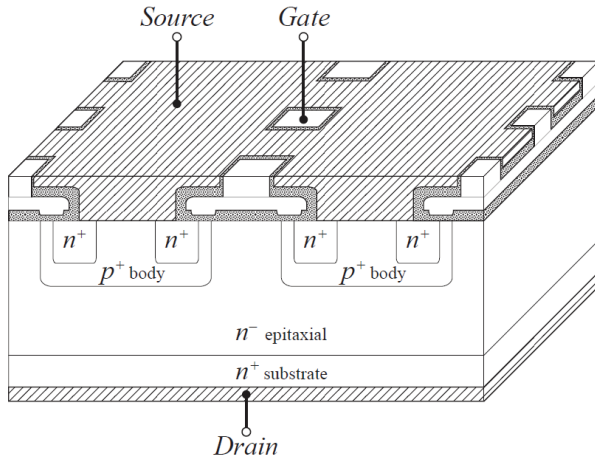
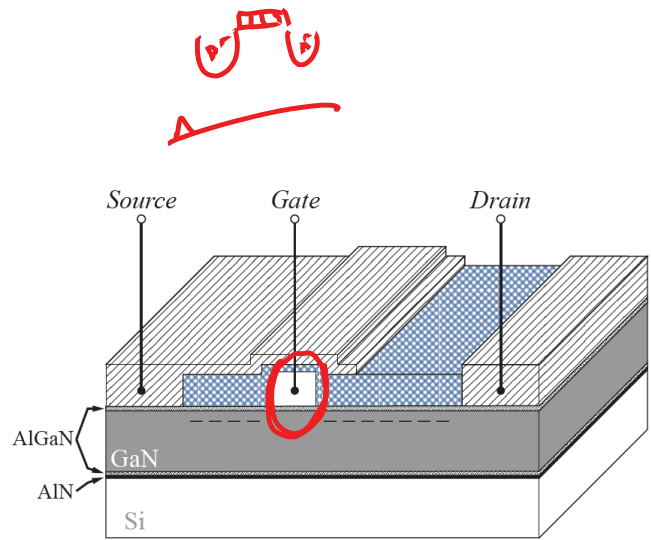


GaN Devices

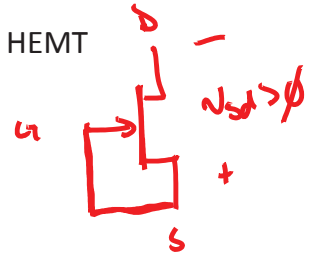


Vertical Silicon Power MOSFET



Lateral GaN HEMT

- No body diode (reverse conduction due to $V_{gd} > V_{gd,th} \approx 2V$)
 - Use antiparallel (schottky) diode or precise dead time
- Significantly faster switching



Designing with GaN

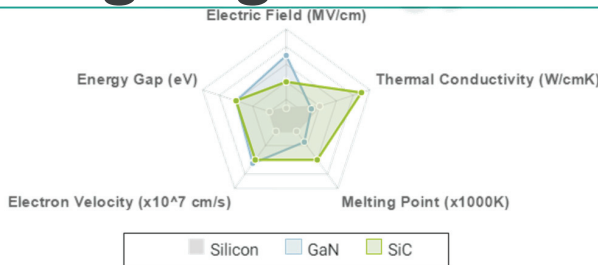
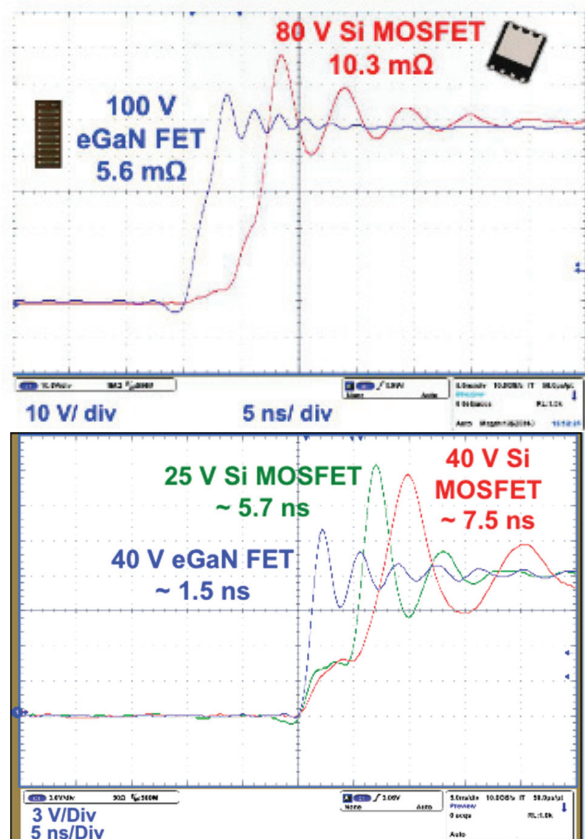


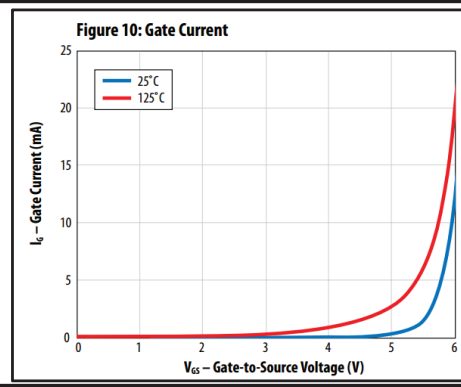
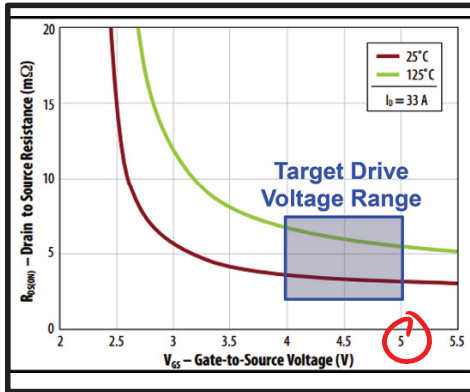
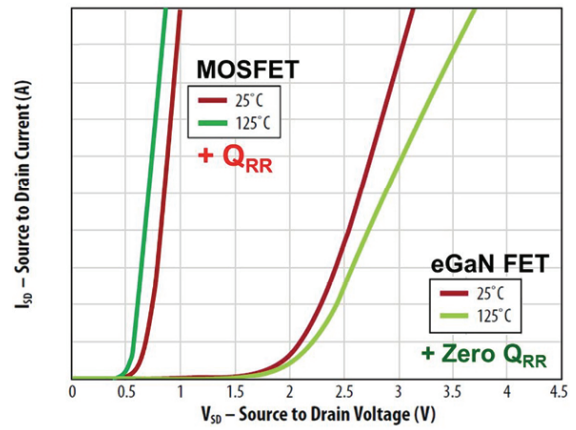
Fig. 1: Material properties of silicon, silicon carbide, and gallium nitride.

- Because of high electric breakdown field and high electron velocity, GaN devices with comparable R_{on} can be significantly smaller and switch must faster.
- Need **very** good layout to prevent ringing from causing overvoltage and device failure.
- More information:
 - <http://potential.eecs.utk.edu/About.php?topic=PowerSemiconductors>



GaN Design Issues

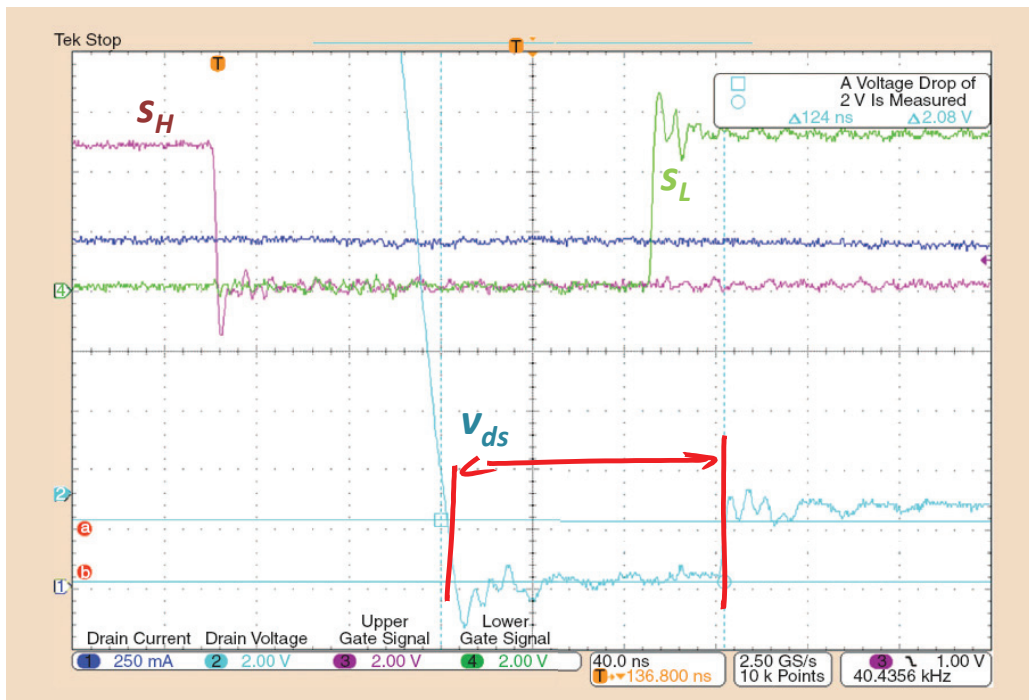
1. → Reverse conduction mechanism
2. → Sensitivity to parasitics
3. → Gate robustness
4. → Small size -> Thermal, soldering difficult



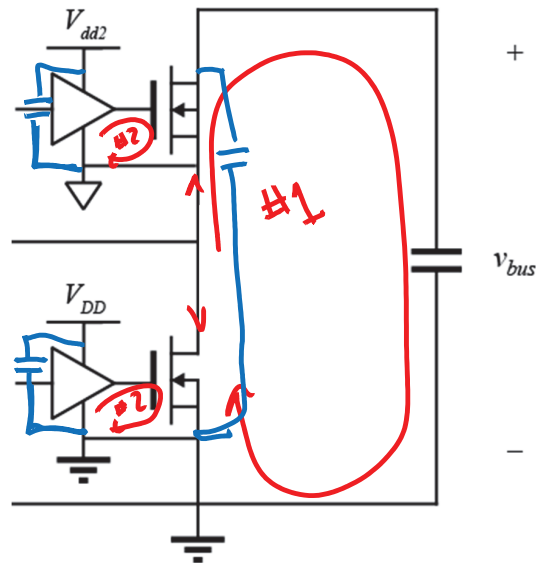
EPC

THE UNIVERSITY OF
TENNESSEE
KNOXVILLE

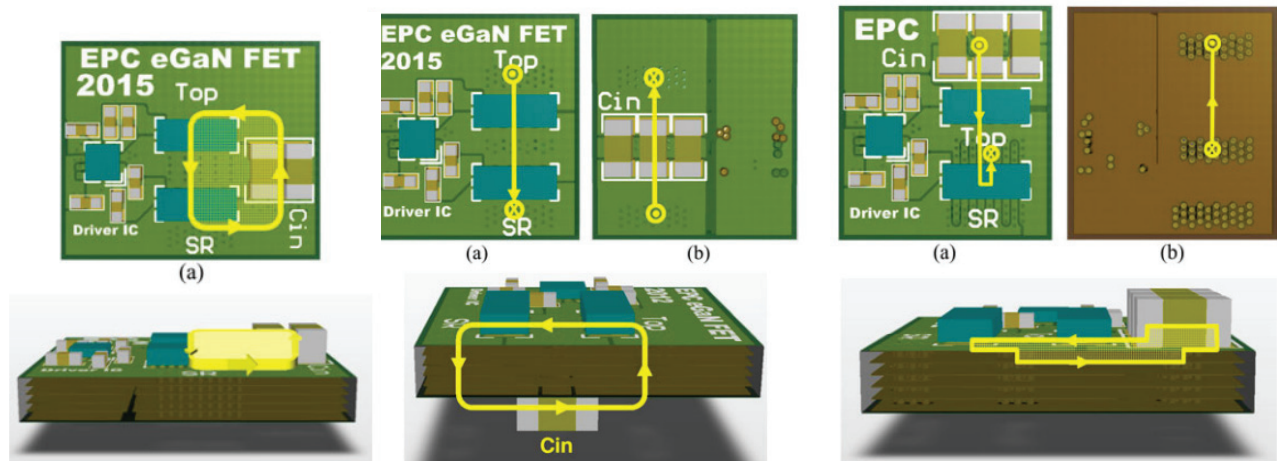
Reverse Conduction



Bridge Layout



Half Bridge Layout: Example

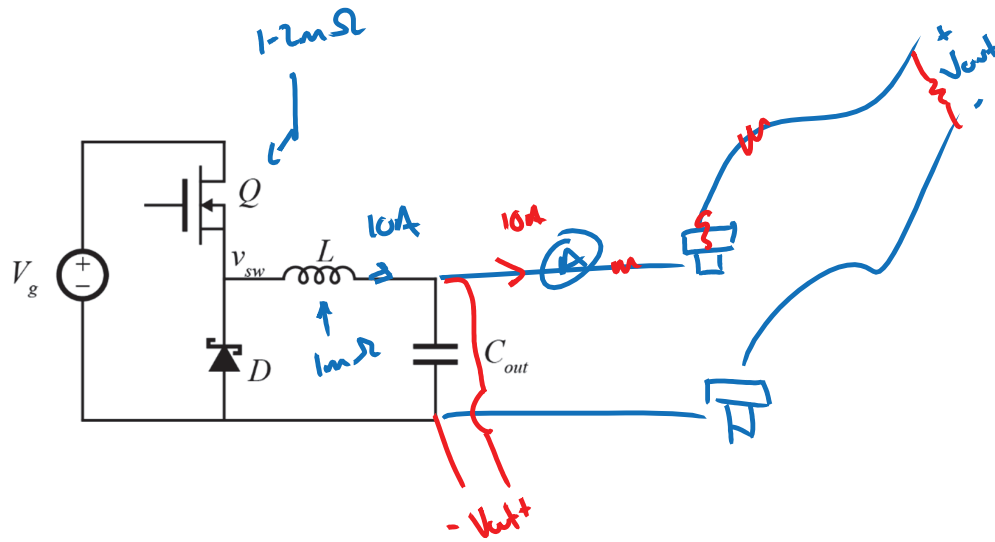


Lateral

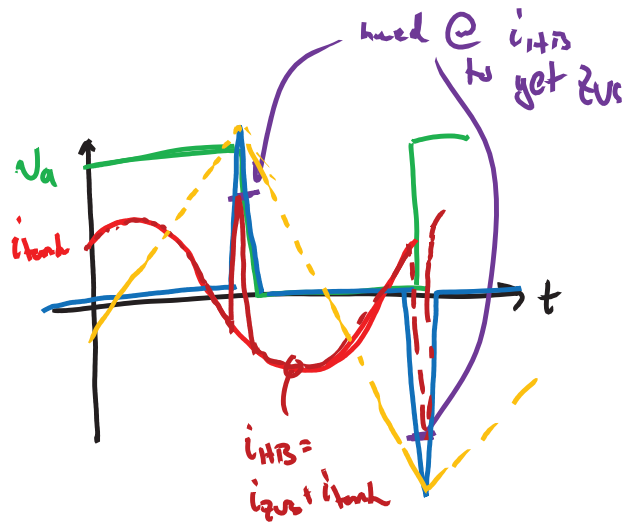
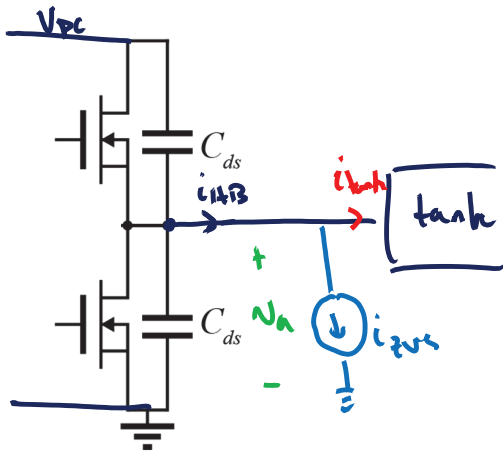
Vertical

"Optimal"

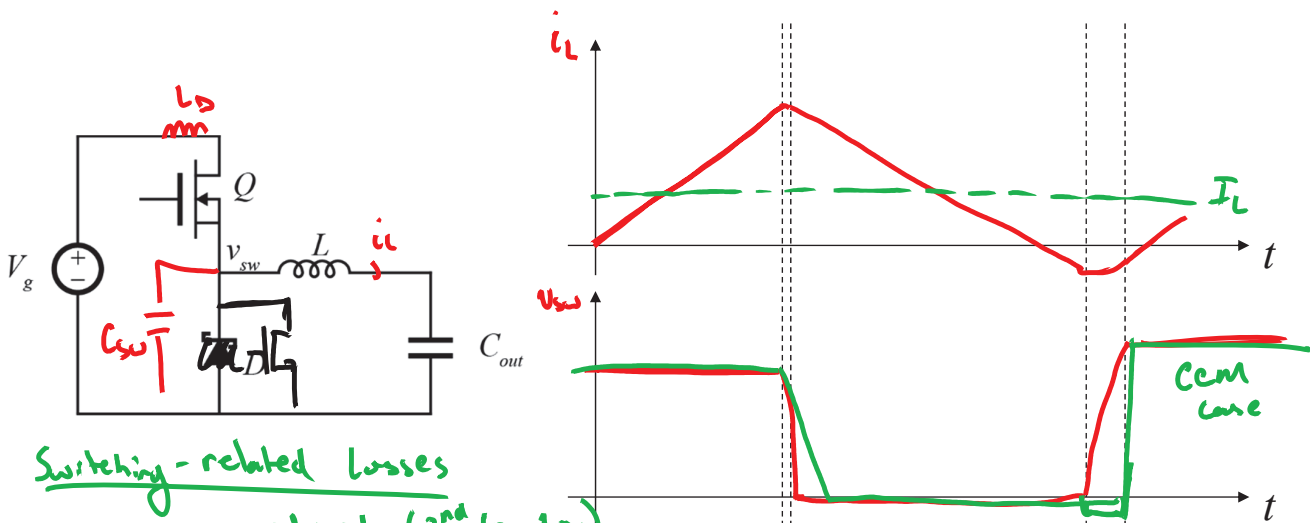
Kelvin Measurement



ZVS Assist Circuits



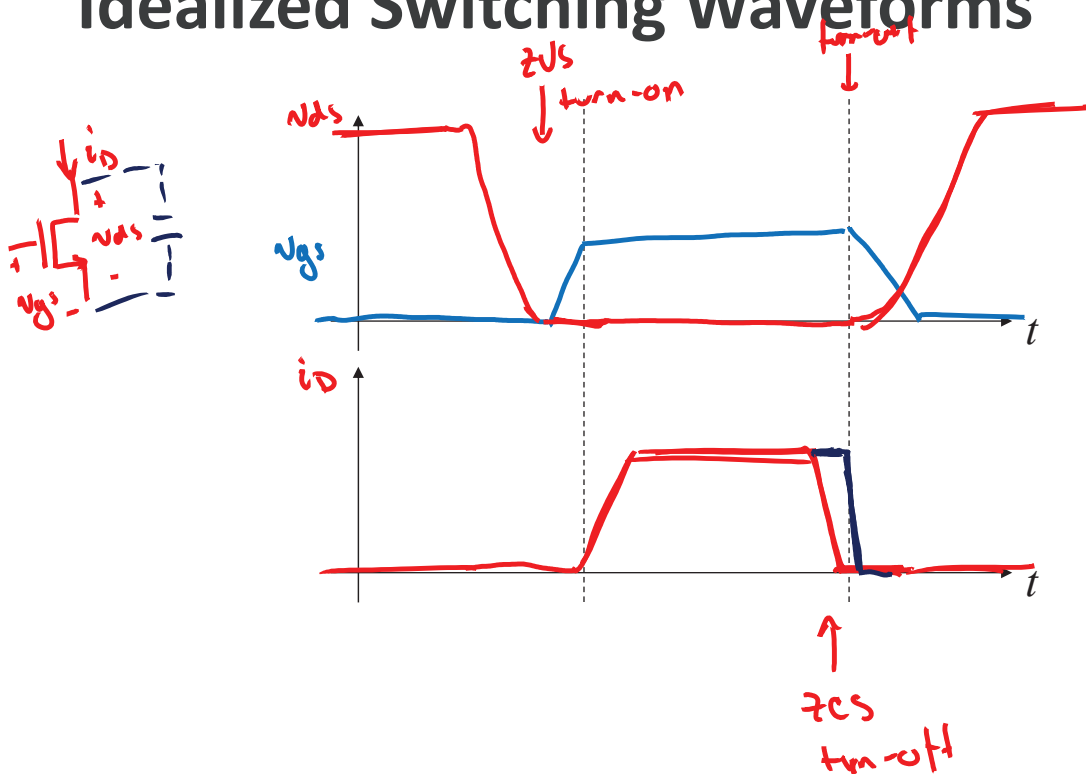
Remaining Switching Losses



Switching-related losses

1. $C_{sw} \rightarrow$ reduced (2nd transition)
2. Overlap \rightarrow
 - turn-on \rightarrow reduced
 - turn-off \rightarrow $\left(\frac{I_L \ll i_{sw}}{C_{sw} \cdot v_{gs}}\right)$
 L_D may increase (!)
3. Reverse Recovery \rightarrow eliminated
4. $Q_g \rightarrow$ no effect
5. Body diode conduction \rightarrow reduce (dt setting)
6. Shoot-through \rightarrow alleviated
7. Additional parasitics (L_D) ?

Idealized Switching Waveforms



Class-E Amplifier

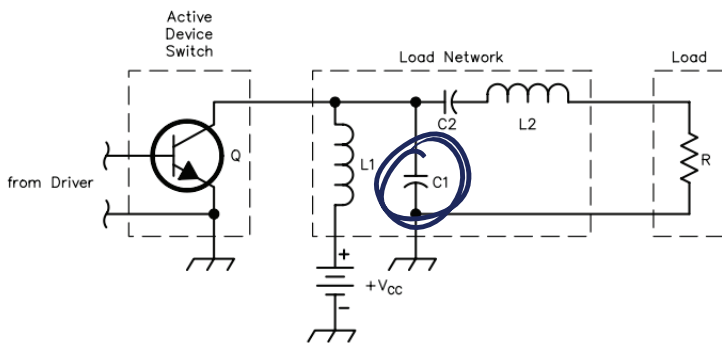


Fig 2—Schematic of a low-order Class-E amplifier.

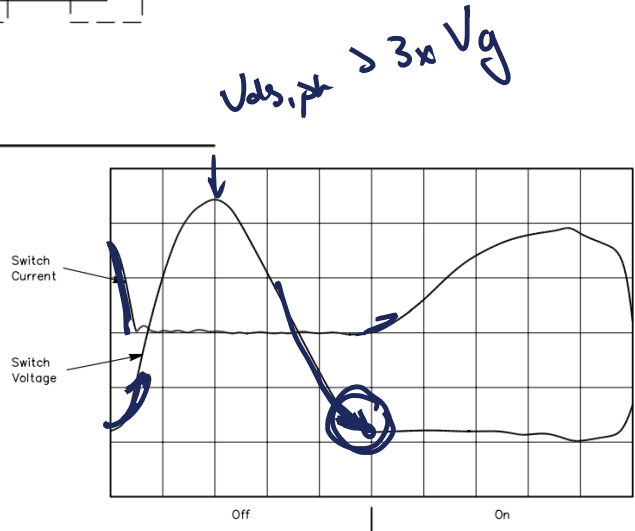
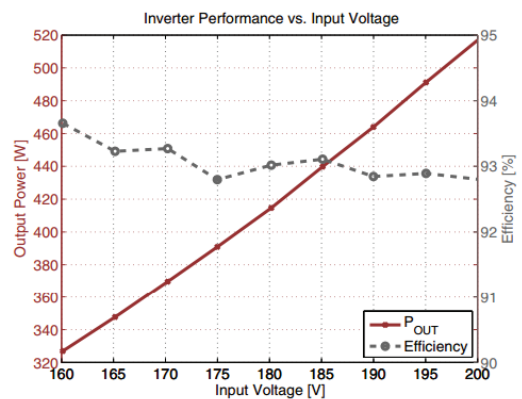
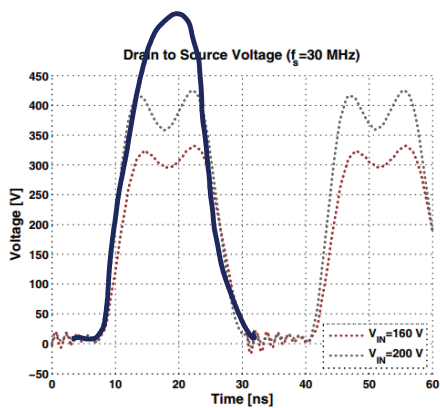
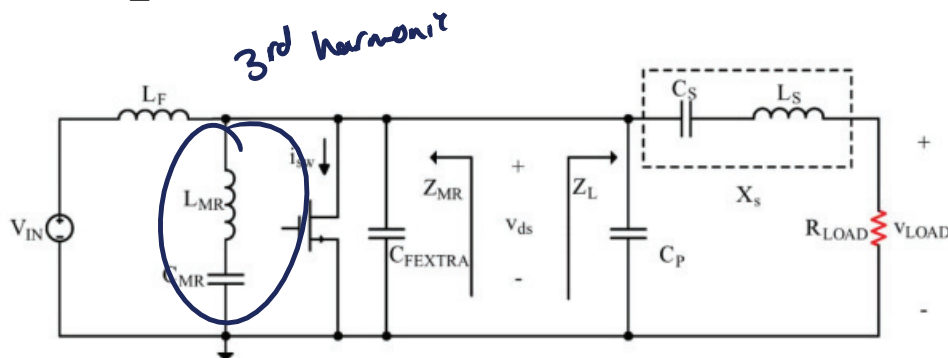


Fig 3—Actual transistor voltage and current waveforms in a low-order Class-E amplifier.

N. O. Sokal, "Class-E RF Power Amplifiers," 2001



Class Φ_2 Inverter



J. M. Rivas, O. Leitermann, Y. Han, A. D. Sagneri, and D. J. Perreault, "A High-Frequency Resonant Inverter Topology With Low-Voltage Stress", 2008



VHF DC-DC Converter

Φ_2 Boost Converter

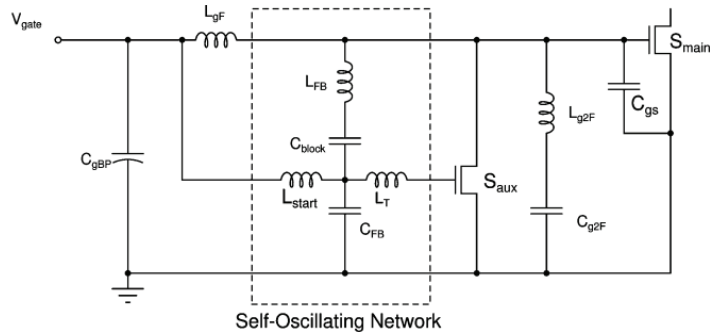
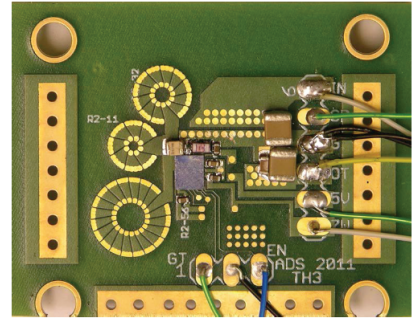
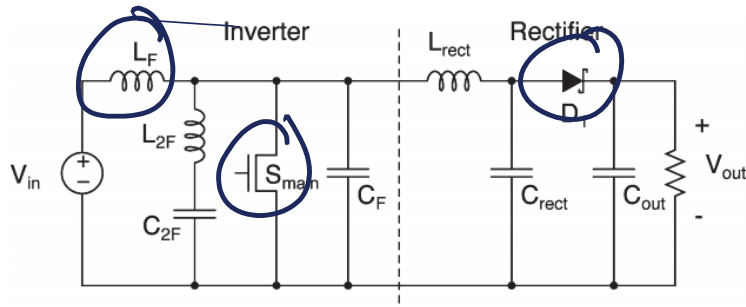


Fig. 5. Trapezoidal resonant gate drive circuit with self-oscillating network. The converter is enabled by applying the voltage V_{gate} , and disabled by setting V_{gate} to zero. This gate driver is employed in the 110-MHz converter (Fig. 9).