

ECE 325 – Electric Energy System Components

4- Transformers

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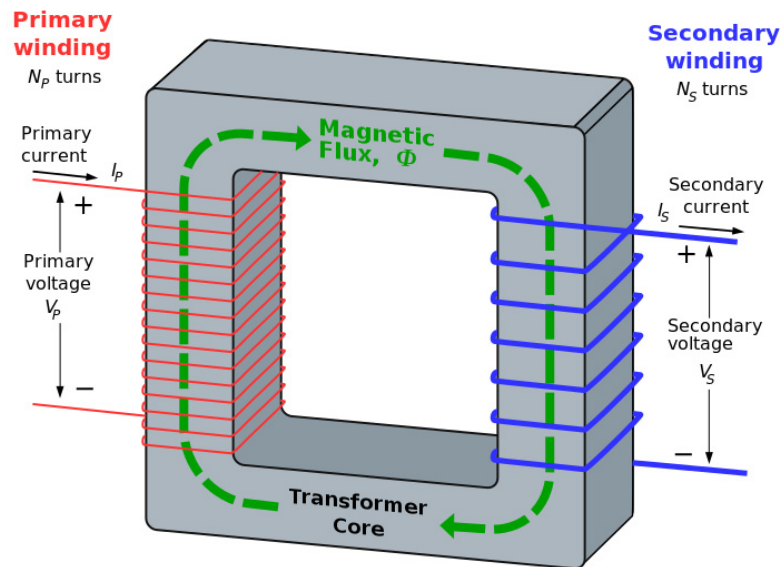
Power Transformer

- Ideal transformers

- Winding resistance is negligible
- No leakage flux
- Permeability of the core is infinite (zero magnetizing current I_p is needed to produce flux)
- No core loss

- Real transformers:

- Windings resistance is not negligible
- Have leakage flux (windings do not link the same flux)
- Permeability of the core is finite
- Have core losses (hysteresis losses and eddy current losses due to time varying flux)



Ideal Transformer

- Primary winding** (assuming sinusoidal flux):

Because there is no flux leakage,

$$\phi_1 = \phi_2 = \phi_m \text{ (mutual flux)}$$

$$\phi_m = \Phi_{\max} \cos \omega t$$

$$e_g = e_1 = N_1 \frac{d\phi_m}{dt} = \omega N_1 \Phi_{\max} (-\sin \omega t) = 2\pi f N_1 \Phi_{\max} \cos(\omega t + 90^\circ)$$

$$= E_{1\max} \cos(\omega t + 90^\circ) \quad \text{Leading } \phi_m \text{ by } 90^\circ$$

$$E_{1\max} = 2\pi f N_1 \Phi_{\max} = \sqrt{2} \times 4.44 f N_1 \Phi_{\max}$$

- Secondary winding:**

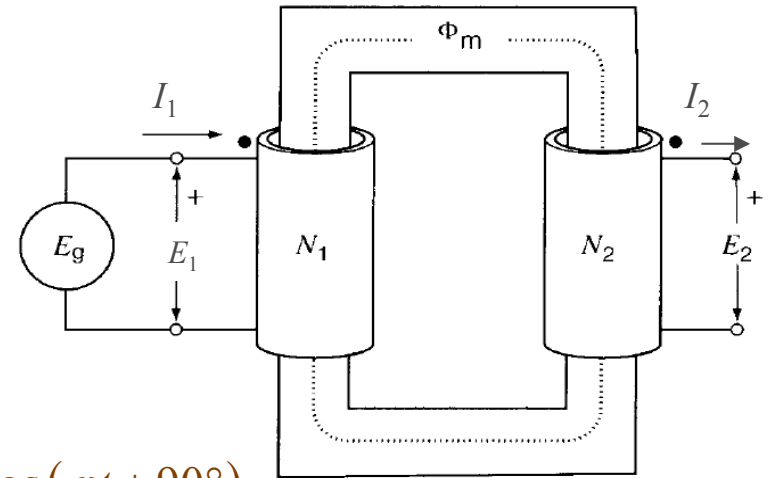
Because there is no flux leakage

$$e_2 = N_2 \frac{d\phi_m}{dt} = -\omega N_2 \Phi_{\max} \sin \omega t$$

$$= E_{2\max} \cos(\omega t + 90^\circ)$$

$$E_{2\max} = 2\pi f N_2 \Phi_{\max} = \sqrt{2} \times 4.44 f N_2 \Phi_{\max}$$

- Voltage/turns ratio** $a = E_1/E_2 = N_1/N_2$



$$\begin{aligned} E_g = E_1 &= \frac{E_{1\max}}{\sqrt{2}} \angle 90^\circ \\ &= 4.44 f N_1 \Phi_{\max} \angle 90^\circ \end{aligned}$$

$$\begin{aligned} E_2 &= \frac{E_{2\max}}{\sqrt{2}} \angle 90^\circ \\ &= 4.44 f N_2 \Phi_{\max} \angle 90^\circ \end{aligned}$$

No-load and Load Conditions

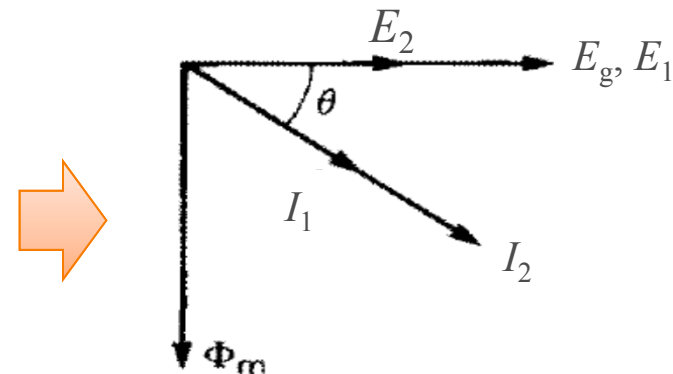
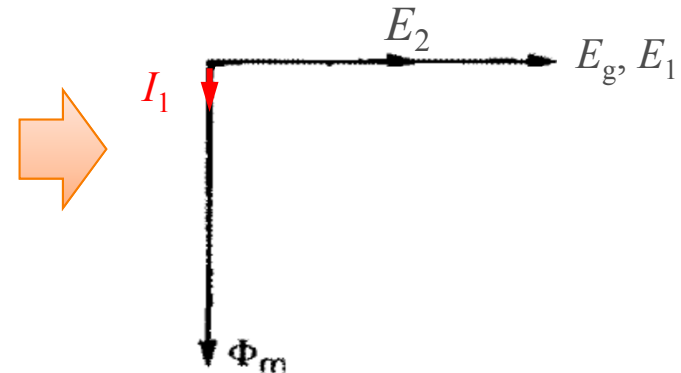
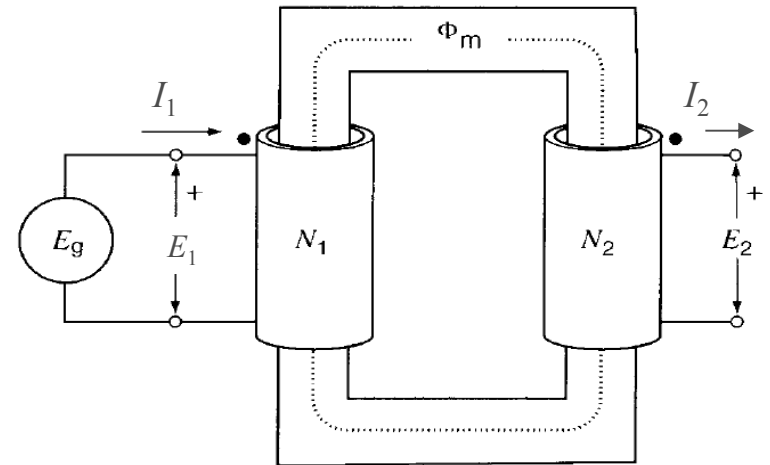
- **Under no-load conditions**, because of infinitely permeable core, no magnetizing current is required to produce flux ϕ_m .

$$I_1 = I_2 = 0$$

- **Under load conditions** ($I_1 \neq 0$, $I_2 \neq 0$, power factor angle θ) because of the infinitely permeable core, there is an exact mmf balance

$$F_1 = I_1 N_1 = I_2 N_2 = F_2$$

$$a = \frac{E_1}{E_2} = \frac{I_2}{I_1} = \frac{N_1}{N_2}$$



Circuit Symbol for an Ideal Transformer

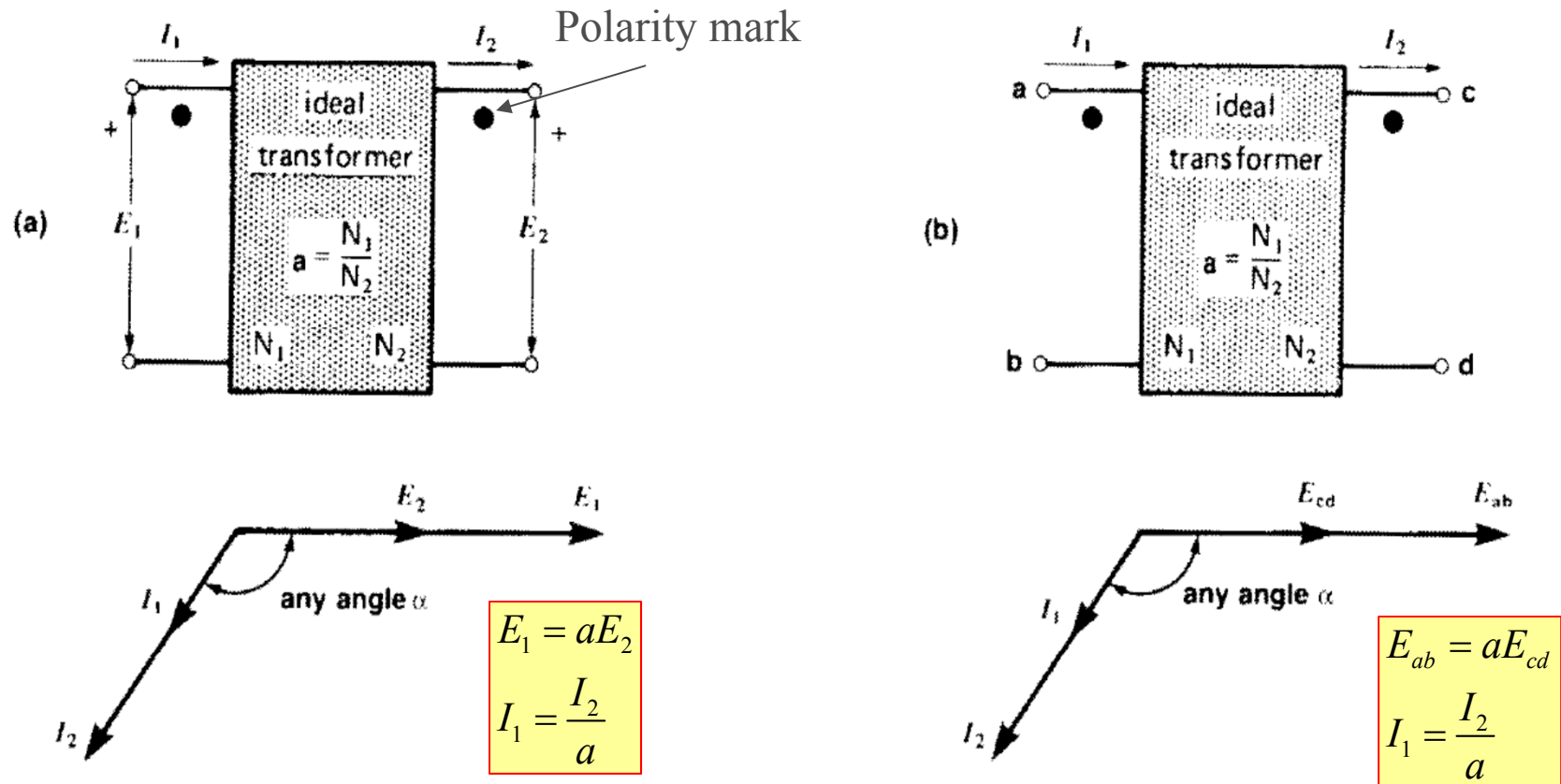


Figure 9.11

- Symbol for an ideal transformer and phasor diagram using sign notation.
- Symbol for an ideal transformer and phasor diagram using double-subscript notation.

Impedance Ratio

$$Z = \frac{E_2}{I_2}$$

$$\begin{aligned} E_1 &= aE_2 \\ I_1 &= I_2 / a \end{aligned}$$

$$Z_x = \frac{E_1}{I_1} = \frac{aE_2}{I_2 / a} = a^2 Z$$

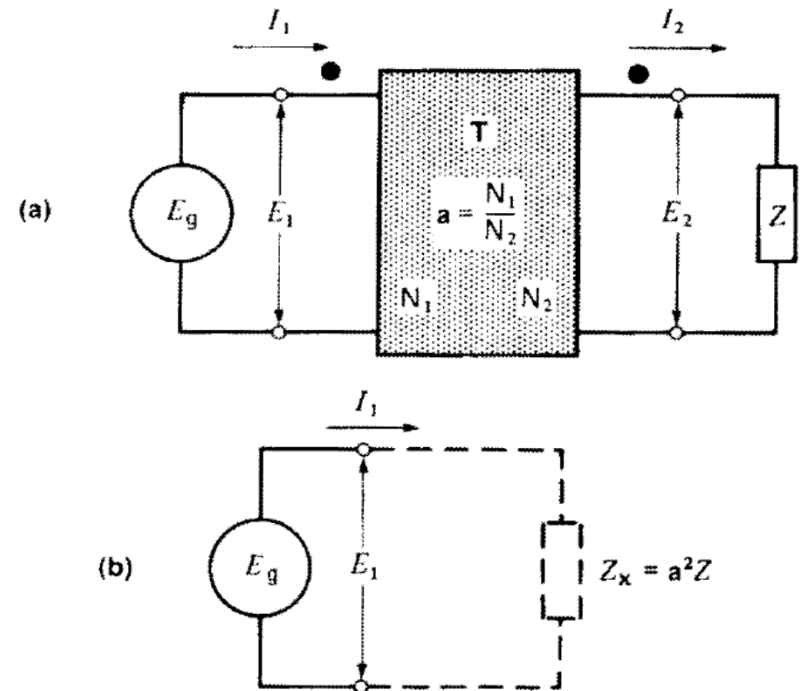


Figure 9.12

a. Impedance transformation using a transformer.
b. The impedance seen by the source differs from Z .

- **Impedance transformation**

- The impedance seen by the source (primary side) is a^2 times the real impedance (secondary side)
- Thus, an ideal transformer has the amazing ability to increase or decrease the value of an impedance

Shifting Impedances (P←S)

$$Z_p \leftarrow \boxed{a^2} \leftarrow Z_s$$

$$E_p \leftarrow \boxed{a} \leftarrow E_s$$

$$I_p \leftarrow \boxed{1/a} \leftarrow I_s$$

