ECE 522 - Power Systems Analysis II Spring 2021

Generator Modeling

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Outline

Modeling of synchronous generators for Stability Studies

- Synchronous Machine Modeling
- Simplified Models for Stability Studies
- Materials:
 - Kundur's Chapters 3-5
 - Saadat's Chapter 8

Synchronous Generators



Types of Rotors

- Salient pole rotors
 - Have concentrated windings on poles and non-uniform air gap
 - Short axial length and large diameter to extract the maximum power from a waterfall
 - On hydraulic turbines operated at low speeds at 50-300 r/min (having a large number of poles)
 - Have a squirrel-cage windings (damper windings) embedded in the pole-faces to help damp out speed oscillations

12 poles salient-pole rotor (12 MW)

• Cylindrical/round rotors

- Distributed winding and uniform air gap
- Large axial length and small diameter to limit the centrifugal forces
- Steam and gas turbines, operated at high speeds, typically 1800-3600r/min (4 or 2-pole)
- Eddy in the solid steal rotor gives damping effects
- 70% of large synchronous generators (150-1500MVA)



Round rotor generator under construction

(Source: http://emadrlc.blogspot.com)

Stator and Rotor Windings

Armature windings:

• a-a', b-b' and c-c' windings

Rotor windings:

- Field windings
 - Field winding F-F' produces a flux on the d-axis.
- Damper windings
 - Two amortisseur/damper windings D-D' and Q-Q' respectively on d- and q-axes, which could be actual windings or effective parts of the solid steal rotor.
 - For a round-rotor machine, for symmetry, consider two windings on each axis by adding a second damper winding G-G' to the q-axis

Total number of windings:

- Salient pole: 3+3 (a, b, c, F, D, Q)
- Round-rotor: 3+4 (a, b, c, F, D, Q, G)

Rotor position

- θ is the displacement of d-axis from the axis of a-a'
- Consider a reference frame rotating synchronously with d and q-axes at speed ω_r (assumed to be along with the axis of a-a' at t=0 in the figure)
- δ is the displacement of q-axis from the rotating reference axis



Voltage and Flux Equations (Salient-pole machine)

- The main objective of synchronous machine modeling is to find the minimum set of constant parameters ٠ by which voltage and flux equations are in the simplest form for desired accuracy.
- All windings are modeled as magnetically coupled circuits with inductances depending on θ . •



Stator-to-rotor mutual inductances (l_{aF} , l_{bD} , l_{aO})

Each rotor winding (F, D, Q) is considered a load in its circuit having ψ and *i* in the same direction.

Each stator winding (a, b, c) is considered a source in its circuit having ψ and *i* in opposite directions.

$$e_{a} = -R_{a}i_{a} + \frac{d\psi_{a}}{dt}$$

$$i \qquad \psi_{a} = -\sum_{a} l_{ab}i_{b} + \sum_{b} \sum_{a} k_{b}i_{b}$$

$$= -\sum_{k=a,b,c} l_{ak} i_k + \sum_{j=F,D,Q} l_{aj} i_j$$

Stator mutual inductances (l_{ab}, l_{bc}, l_{ac})

e_a		R_a	0	0	0	0	0	$ -i_a $		Ψ_a
e_b		0	R_{b}	0	0	0	0	$-i_b$		$\psi_{\scriptscriptstyle b}$
e_{c}		0	0	R_{c}	0	0	0	$-i_c$	d	ψ_{c}
e_{F}	=	0	0	0	R_{F}	0	0	i_F	$+\frac{1}{dt}$	$\psi_{\scriptscriptstyle F}$
$e_D = 0$		0	0	0	0	R_{D}	0	i_D		$\psi_{\scriptscriptstyle D}$
$e_Q = 0$		0	0	0	0	0	R_{Q}	i_Q		ψ_{Q}
$\begin{bmatrix} \mathbf{e}_{abc} \\ \mathbf{e}_{FDQ} \end{bmatrix}$	=	$\begin{bmatrix} \mathbf{R}_{ab} \\ 0 \end{bmatrix}$	ec F	0 R _{FDQ}][- i	$-\mathbf{i}_{abc}$	$\left] + \frac{d}{dt} \right]$	$\begin{bmatrix} \Psi_{abc} \\ \Psi_{FD} \end{bmatrix}$		
$\begin{bmatrix} \psi_a \\ \psi_b \end{bmatrix}$		l_{aa} l_{ba}	l_{ab} l_{bb}	l_{a} l_{b}		aF bF	l_{aD} l_{bD}	$\begin{bmatrix} l_{aQ} \\ l_{bQ} \end{bmatrix}$	$\begin{bmatrix} -i\\ -i \end{bmatrix}$	a

$$\begin{bmatrix} \psi_{a} \\ \psi_{b} \\ \psi_{c} \\ \psi_{F} \\ \psi_{P} \\ \psi_{Q} \end{bmatrix} = \begin{bmatrix} l_{aa} & l_{ab} & l_{ac} & l_{aF} & l_{aD} & l_{aQ} \\ l_{ba} & l_{bb} & l_{bc} & l_{bF} & l_{bD} & l_{bQ} \\ l_{ca} & l_{cb} & l_{cc} & l_{cF} & l_{cD} & l_{cQ} \\ l_{Fa} & l_{Fb} & l_{Fc} & l_{FF} & l_{FD} & l_{FQ} \\ l_{Da} & l_{Db} & l_{Dc} & l_{DF} & l_{DD} & l_{DQ} \\ l_{Qa} & l_{Qb} & l_{Qc} & l_{QF} & l_{QD} & l_{QQ} \end{bmatrix} \begin{bmatrix} l_{a} \\ l_{a} \\ -i_{b} \\ -i_{c} \\ i_{F} \\ i_{D} \\ i_{Q} \end{bmatrix}$$

$\left[\Psi_{abc} \right]$	_	\mathbf{L}_{SS}	\mathbf{L}_{SR}	$\begin{bmatrix} -\mathbf{i}_{abc} \end{bmatrix}$
$[\Psi_{FDQ}]$	_	L_{RS}	\mathbf{L}_{RR}	i _{FDQ}

Variations of Self- and Mutual Inductances



 $l_{xy} = N_x \times N_y \times P_{xy} = l_{yx}$ (Symmetric)

Each winding is either stationary (on stator) or rotating (on rotor) together with the magnet: $x, y \in \{a, b, c, F, D, Q\}$

What is each l_{xy} ?



 $N_x \times N_y$: effective coupling

 P_{xy} : permeance of the mutual flux path (mainly influenced by air gap)

- Stator self/mutual inductances (e.g. l_{aa} and l_{ab}): $N_x \times N_y$ is constant; P_{xy} is a periodic function (assumed to be cosine) of θ with a period 180° due to periodic air gap changes $P_{xy} \approx P_0 + P(\theta + \alpha) \rightarrow l_{xy} = l_0 + l_2 \cos 2(\theta + \alpha)$
- Stator to Rotor Mutual Inductances (e.g. l_{aF}): P_{xy} between stator and rotor windings is almost constant; however, their effective coupling is a periodic function of θ with period 360°; the flux leakage can be ignored ($l_0=0$).

 $N_x \times N_y \approx N_0 \cos(\theta + \beta) \rightarrow l_{xy} = l_1 \cos(\theta + \beta)$

• Rotor Inductances are all constant using a reference revolving with the rotor

Stator self-inductances (l_{aa} , l_{bb} , l_{cc}) and Mutual Inductances (l_{ab} , l_{bc} , l_{ac})

• l_{aa} is equal to the ratio of flux linking phase *a* winding to current i_a , with zero currents in all other circuits (maximum P_{aa} at θ =0° or 180°)

 $l_{aa} = L_s + L_m \cos 2\theta$ $l_{bb} = L_s + L_m \cos 2(\theta - 120^\circ)$ $l_{cc} = L_s + L_m \cos 2(\theta + 120^\circ)$ $L_s > L_m \ge 0$



• l_{ab} <0 since windings *a* and *b* have 120° (>90°) displacement (maximum P_{ab} at θ = -30° or 150°)

$$l_{ab} = -M_s - L_m \cos 2(\theta + 30^\circ) = l_{ba}$$
$$l_{bc} = -M_s - L_m \cos 2(\theta - 90^\circ) = l_{bc}$$
$$l_{ca} = -M_s - L_m \cos 2(\theta + 150^\circ) = l_{ac}$$
$$M_s \approx L_s/2$$



Stator to Rotor Mutual Inductances

$(\mathbf{L}_{SR}: l_{aF} \ l_{bF} \ l_{cF} \ l_{aD} \ l_{bD} \ l_{cD} \ l_{aQ} \ l_{bQ} \ l_{cQ})$

- The rotor sees a constant permeance if neglecting variations in the air gap due to stator slots
- When the flux linking a stator winding and a rotor winding reaches the maximum when they aligns with each other and is 0 when they are displaced by 90°

 $\begin{array}{c} b \\ q(Q) \\ \theta \\ \theta \\ c \\ \end{array}$

(maximum $N_a \times N_F$ at $\theta = 0^\circ$)

d-axis	$l_{aF} = l_{Fa} = M_F \cos\theta$	$l_{aD} = l_{Da} = M_D \cos\theta$
	$l_{bF} = l_{Fb} = M_F \cos(\theta - 120^\circ)$	$l_{bD} = l_{Db} = M_D \cos(\theta - 120^\circ)$
	$l_{cF} = l_{Fc} = M_F \cos(\theta + 120^\circ)$	$l_{cD} = l_{Dc} = M_D \cos(\theta + 120^\circ)$

$$l_{aQ} = l_{Qa} = -M_Q \sin\theta$$
$$l_{bQ} = l_{Qb} = -M_Q \sin(\theta - 120^\circ)$$
$$l_{cQ} = l_{Qc} = -M_Q \sin(\theta + 120^\circ)$$

What if we define q-axis lagging d-axis by 90°? (no negative signs)

For Salient-pole Rotors



For round rotors, which of the curves will be different?

Rotor Inductances (L_{RR}: l_{FF} , l_{DD} , l_{QQ} , l_{FD} , l_{FQ} , l_{DQ})

- They are all constant
 - Rotor self inductances

$$\begin{split} l_{FF} &\triangleq L_F \\ l_{DD} &\triangleq L_D \\ l_{QQ} &\triangleq L_Q \end{split}$$

Rotor mutual inductances

$$l_{FD} = l_{DF} \triangleq M_R$$
$$l_{FQ} = l_{QF} = 0$$
$$l_{DQ} = l_{QD} = 0$$

Summary

$$\begin{bmatrix} \Psi_{abc} \\ \Psi_{FDQ} \end{bmatrix} = \begin{bmatrix} \mathbf{L}_{SS} & \mathbf{L}_{SR} \\ \mathbf{L}_{RS} & \mathbf{L}_{RR} \end{bmatrix} \begin{bmatrix} -\mathbf{i}_{abc} \\ \mathbf{i}_{FDQ} \end{bmatrix} \qquad \mathbf{L}_{SS} = \begin{bmatrix} L_s + L_m \cos 2\theta & -M_s - L_m \cos 2\left(\theta + \frac{\pi}{6}\right) & -M_s - L_m \cos 2\left(\theta + \frac{\pi}{6}\right) \\ -M_s - L_m \cos 2\left(\theta + \frac{\pi}{6}\right) & L_s + L_m \cos 2\left(\theta - \frac{2\pi}{3}\right) & -M_s - L_m \cos 2\left(\theta - \frac{\pi}{2}\right) \\ -M_s - L_m \cos 2\left(\theta + \frac{5\pi}{6}\right) & -M_s - L_m \cos 2\left(\theta - \frac{\pi}{2}\right) & L_s + L_m \cos 2\left(\theta + \frac{2\pi}{3}\right) \end{bmatrix}$$

$$L_{SR} = \begin{bmatrix} M_F \cos\theta & M_D \cos\theta & -M_Q \sin\theta \\ M_F \cos\left(\theta - \frac{2\pi}{3}\right) & M_D \cos\left(\theta - \frac{2\pi}{3}\right) & -M_Q \sin\left(\theta - \frac{2\pi}{3}\right) \\ M_F \cos\left(\theta + \frac{2\pi}{3}\right) & M_D \cos\left(\theta + \frac{2\pi}{3}\right) & -M_Q \sin\left(\theta + \frac{2\pi}{3}\right) \end{bmatrix} \qquad L_{RS} = L^T_{SR} \qquad L_{RR} = \begin{bmatrix} L_F & M_R & 0 \\ M_R & L_D & 0 \\ 0 & 0 & L_Q \end{bmatrix}$$

Further thinking:

- L_{RR} is constant because it is in a reference frame rotating with the rotor
- \mathbf{L}_{SS} and \mathbf{L}_{SR} are $\theta(t)$ dependent due to variations of $N_x \times N_y$ or P_{xy} caused by the rotation of the rotor relative to the stator

$$\psi_{FDQ} = -\mathbf{L}_{RS}(\theta) \times \mathbf{i}_{abc} + \mathbf{L}_{RR} \times \mathbf{i}_{FDQ}$$
 Only $\mathbf{L}_{SR}(\theta)$ is explicitly dependent of θ

$$\Leftrightarrow \mathbf{L}_{RS}(\theta) \times \mathbf{i}_{abc} = -\psi_{FDQ} + \mathbf{L}_{RR} \times \mathbf{i}_{FDQ}$$
 RHS is in the rotor reference and independent of θ .

LHS =
$$[\mathbf{L}_{RS}(\theta) \mathbf{P}^{-1}(\theta)] \times [\mathbf{P}(\theta) \mathbf{i}_{abc}] \stackrel{\text{def}}{=} \mathbf{L}'_{RS} \times \mathbf{i}'$$

$$\mathbf{L}_{RS}(\theta) \times \mathbf{i}_{abc} = \begin{bmatrix} M_F \\ M_D \\ M_Q \end{bmatrix} \begin{bmatrix} \cos\theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3}) \\ \cos\theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3}) \\ -\sin\theta & -\sin(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \\ -\sin\theta & -\sin(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

$$= \begin{bmatrix} M_F \\ M_D \\ M_Q \end{bmatrix} \begin{bmatrix} i_a \cos\theta + i_b \cos(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \\ i_a \cos\theta + i_b \cos(\theta - \frac{2\pi}{3}) & +i_c \cos(\theta + \frac{2\pi}{3}) \\ -i_a \sin\theta + i_b \sin(\theta - \frac{2\pi}{3}) & +i_c \sin(\theta + \frac{2\pi}{3}) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

$$i_d$$

$$= \begin{bmatrix} M_F \\ M_D \\ M_Q \end{bmatrix} \begin{bmatrix} i_a \cos\theta + i_b \cos(\theta - \frac{2\pi}{3}) + i_c \cos(\theta + \frac{2\pi}{3}) \\ i_a \cos\theta + i_b \cos(\theta - \frac{2\pi}{3}) + i_c \cos(\theta + \frac{2\pi}{3}) \\ -i_a \sin\theta + i_b \sin(\theta - \frac{2\pi}{3}) + i_c \sin(\theta + \frac{2\pi}{3}) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

$$i_d$$

$$= \begin{bmatrix} M_F \\ M_D \\ M_Q \end{bmatrix} \begin{bmatrix} i_a \cos\theta + i_b \cos(\theta - \frac{2\pi}{3}) + i_c \cos(\theta + \frac{2\pi}{3}) \\ -i_a \sin\theta + i_b \sin(\theta - \frac{2\pi}{3}) + i_c \sin(\theta + \frac{2\pi}{3}) \\ -i_c \sin\theta + i_b \sin(\theta - \frac{2\pi}{3}) + i_c \sin(\theta + \frac{2\pi}{3}) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

• For balanced steady-state conditions:

$$i_{a} = I_{m} \sin \omega_{s}t, \quad i_{b} = I_{m} \sin(\omega_{s}t - 2\pi/3), \quad i_{c} = I_{m} \sin(\omega_{s}t + 2\pi/3), \quad \theta = \omega_{r}t + \delta \pi/2, \quad \omega_{r} \approx \omega_{s}$$

$$i_{a} \cos \theta + i_{b} \cos(\theta - \frac{2\pi}{3}) + i_{c} \cos(\theta + \frac{2\pi}{3}) = I_{m} \sin \omega_{s}t \cos \theta + I_{m} \sin(\omega_{s}t - \frac{2\pi}{3})\cos(\theta - \frac{2\pi}{3}) + I_{m} \sin(\omega_{s}t + \frac{2\pi}{3})\cos(\theta + \frac{2\pi}{3})$$

$$= \frac{3I_{m}}{2}\sin(\omega_{s}t - \theta) = \frac{3I_{m}}{2}\sin(\omega_{s}t - \omega_{r}t - \delta + \frac{\pi}{2}) \approx \frac{3I_{m}}{2}\cos\delta \quad \sim \mathbf{i}d$$

$$i_{a} \sin \theta + i_{b} \sin(\theta - \frac{2\pi}{3}) + i_{c} \sin(\theta + \frac{2\pi}{3}) = I_{m} \sin \omega_{s}t \cos(\theta - \frac{\pi}{2}) + I_{m} \sin(\omega_{s}t - \frac{2\pi}{3})\cos(\theta - \frac{2\pi}{3} - \frac{\pi}{2}) + I_{m} \sin(\omega_{s}t + \frac{2\pi}{3})\cos(\theta + \frac{2\pi}{3} - \frac{\pi}{2})$$

$$= \frac{3I_{m}}{2}\sin(\omega_{s}t - \theta + \frac{\pi}{2}) = \frac{3I_{m}}{2}\sin(\omega_{s}t - \omega_{r}t - \delta + \pi) \approx \frac{3I_{m}}{2}\sin\delta \quad \sim \mathbf{i}q$$

$$\mathbf{i}_{a} + \mathbf{i}_{b} + \mathbf{i}_{c} = \mathbf{0} \qquad \sim \mathbf{i}_{0}$$
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Park's (dq0) Transformation

$$\mathbf{i}_{0dq} = \mathbf{P}\mathbf{i}_{abc} \begin{bmatrix} i_0 \\ i_d \\ i_q \end{bmatrix} = \begin{bmatrix} k_0 \\ k_d \\ k_q \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 \\ \cos\theta & \cos(\theta - 2\pi/3) & \cos(\theta + 2\pi/3) \\ -\sin\theta & -\sin(\theta - 2\pi/3) & -\sin(\theta + 2\pi/3) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix} = \begin{bmatrix} k_0 & k_0 & k_0 \\ k_d \cos(\theta - 2\pi/3) & k_d \cos(\theta + 2\pi/3) \\ -k_q \sin(\theta - 2\pi/3) & -k_q \sin(\theta - 2\pi/3) & -k_q \sin(\theta + 2\pi/3) \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix}$$

For balanced steady-state conditions:
$$i_a = I_m \sin \omega_s t$$
$$i_b = I_m \sin(\omega_s t - 2\pi/3)$$
$$i_c = I_m \sin(\omega_s t - 2\pi/3)$$
$$i_c = I_m \sin(\omega_s t - 2\pi/3)$$
$$i_c = I_m \sin(\omega_s t + 2\pi/3)$$
$$\theta = \omega_r t + \delta - \pi/2, \ \omega_r \approx \omega_s$$

• If we expect the transformation to be power invariant:

$$P_{3\phi} = e_a i_a + e_b i_b + e_c i_c = \mathbf{e}_{abc}^T \mathbf{i}_{abc} = (\mathbf{P}^{-1} \mathbf{e}_{0dq})^T \mathbf{P}^{-1} \mathbf{i}_{0dq} = \mathbf{e}_{0dq}^T \mathbf{P}^{-1} \mathbf{i}_{0dq} = e_0 i_0 + e_d i_d + e_q i_q$$

$$\mathbf{P}^{-T} \mathbf{P}^{-1} = \mathbf{I} \rightarrow \mathbf{P}^{-1} = \mathbf{P}^T \text{ or } \mathbf{P}^T \mathbf{P} = \mathbf{I} \quad (\text{orthogonal matrix}) \qquad \qquad k_d = k_q = \sqrt{\frac{2}{3}} \text{ and } k_0 = \frac{1}{\sqrt{3}}$$

$$\mathbf{P} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2} \\ \cos\theta & \cos(\theta - 2\pi/3) & \cos(\theta + 2\pi/3) \\ -\sin\theta & -\sin(\theta - 2\pi/3) & -\sin(\theta + 2\pi/3) \end{bmatrix} \mathbf{P}^{-1} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1/\sqrt{2} & \cos\theta & -\sin\theta \\ 1/\sqrt{2} & \cos(\theta - 2\pi/3) & -\sin(\theta - 2\pi/3) \\ 1/\sqrt{2} & \cos(\theta + 2\pi/3) & -\sin(\theta + 2\pi/3) \end{bmatrix}$$

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What if we define q-axis lagging d-axis by 90°?

$$\begin{array}{c}
\mathbf{Flux Equations} \\
\mathbf{i}_{0dq} = \mathbf{P}_{i_{abc}} \\
\mathbf{\psi}_{odq} = \mathbf{P}_{\psi_{abc}} \\
\end{array} \quad \left(\begin{array}{c}
\mathbf{i}_{0dq} \\
\mathbf{i}_{DD} \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\end{array} \right) = \left(\begin{array}{c}
\mathbf{P} & 0 \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\end{array} \right) = \left(\begin{array}{c}
\mathbf{P} & 0 \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\mathbf{f}_{DD} \\
\end{array} \right) = \left(\begin{array}{c}
\mathbf{P} & 0 \\
\mathbf{f}_{DD} \\
\mathbf{f}_{D$$

$$L_0 = L_s - 2M_s, \quad L_d = L_s + M_s + \frac{3}{2}L_m, \quad L_q = L_s + M_s - \frac{3}{2}L_m, \quad k = \sqrt{\frac{3}{2}}L_m$$

Voltage Equations $\begin{vmatrix} \mathbf{e}_{abc} \\ \mathbf{e}_{EDO} \end{vmatrix} = \begin{vmatrix} \mathbf{R}_{abc} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{EDO} \end{vmatrix} \begin{vmatrix} -\mathbf{i}_{abc} \\ \mathbf{i}_{EDO} \end{vmatrix} + \frac{d}{dt} \begin{vmatrix} \Psi_{abc} \\ \Psi_{EDO} \end{vmatrix}$ $\begin{bmatrix} \mathbf{P}^{-1} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{bmatrix} \begin{bmatrix} \mathbf{e}_{0dq} \\ \mathbf{e}_{FDQ} \end{bmatrix} = \begin{bmatrix} \mathbf{R}_{abc} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{FDQ} \end{bmatrix} \begin{bmatrix} \mathbf{P}^{-1} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{bmatrix} \begin{vmatrix} -\mathbf{i}_{0dq} \\ \mathbf{i} \end{vmatrix} + \frac{d}{dt} \begin{bmatrix} \Psi_{abc} \\ \Psi_{FDO} \end{bmatrix} \qquad \begin{bmatrix} \mathbf{P} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{bmatrix} \begin{bmatrix} \Psi_{abc} \\ \psi_{FDQ} \end{bmatrix} \neq \begin{bmatrix} \Psi_{0dq} \\ \psi_{FDQ} \end{bmatrix}$ $\begin{vmatrix} \mathbf{e}_{0dq} \\ \mathbf{e}_{EDO} \end{vmatrix} = \begin{vmatrix} \mathbf{P} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{vmatrix} \begin{vmatrix} \mathbf{R}_{abc} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{EDO} \end{vmatrix} \begin{vmatrix} \mathbf{P}^{-1} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{vmatrix} \begin{vmatrix} -\mathbf{i}_{0dq} \\ \mathbf{i}_{EDO} \end{vmatrix} + \begin{vmatrix} \mathbf{P} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{vmatrix} \begin{vmatrix} \mathbf{P}^{-1} & \mathbf{0} \\ \mathbf{0} & \mathbf{I} \end{vmatrix} \begin{vmatrix} \Psi_{0dq} \\ \Psi_{EDO} \end{vmatrix}$ $\left[\begin{array}{c} \mathbf{e}_{0dq} \\ \mathbf{e}_{FDQ} \end{array} \right] = \left[\begin{array}{c} \mathbf{R}_{0dq} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{FDQ} \end{array} \right] \left| \begin{array}{c} -\mathbf{i}_{0dq} \\ \mathbf{i} \end{array} \right| + \left[\begin{array}{c} \mathbf{P} \frac{d}{dt} \mathbf{P}^{-1} \mathbf{\Psi}_{0dp} \\ \mathbf{\Psi}_{FDO} \end{array} \right] + \left[\begin{array}{c} \mathbf{P} \frac{d}{dt} \mathbf{P}^{-1} \mathbf{\Psi}_{0dp} \\ \mathbf{Q} \end{array} \right]$ Note: **P** is NOT constant $\mathbf{P}\frac{d}{dt}\mathbf{P}^{-1} = \mathbf{P}\frac{d\theta}{dt}\frac{d}{d\theta}\mathbf{P}^{-1} = \omega_{r}\mathbf{P}\frac{d}{d\theta}\mathbf{P}^{-1} = \frac{2\omega_{r}}{3}\begin{bmatrix}1/\sqrt{2} & 1/\sqrt{2} & 1/\sqrt{2}\\\cos\theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3})\\-\sin\theta & -\sin(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3})\end{bmatrix} \times \begin{vmatrix}0 & -\sin\theta & -\cos\theta\\0 & -\sin(\theta - \frac{2\pi}{3}) & -\cos(\theta - \frac{2\pi}{3})\\0 & -\sin(\theta + \frac{2\pi}{3}) & -\cos(\theta + \frac{2\pi}{3})\end{vmatrix} = \begin{bmatrix}0 & 0 & 0\\0 & 0 & -\omega_{r}\\0 & \omega_{r} & 0\end{bmatrix}$ $\mathbf{P}\frac{d}{dt}\mathbf{P}^{-1}\psi_{0dq} = \begin{vmatrix} 0 & \mathbf{U} \\ -\omega_r\psi_q \\ -\omega_r\psi_q \end{vmatrix} = \begin{vmatrix} 0 & (-\omega_rL_q) \times (-i_q) - \omega_rkM_Qi_Q \\ (-\omega_rL_q) \times (-i_q) + \omega_rkM_Qi_Q \end{vmatrix}$

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Voltage Equations (cont'd)



If the neutral line-to-ground voltage is not zero

$$\begin{bmatrix} \mathbf{e}_{0dq} \\ \mathbf{e}_{FDQ} \end{bmatrix} = \begin{bmatrix} \mathbf{R}_{0dq} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{FDQ} \end{bmatrix} \begin{bmatrix} -\mathbf{i}_{0dq} \\ \mathbf{i}_{FDQ} \end{bmatrix} + \begin{bmatrix} \mathbf{P} \frac{d}{dt} \mathbf{P}^{-1} \times \mathbf{\Psi}_{0dp} \\ \mathbf{\Psi}_{FDQ} \end{bmatrix} + \frac{d}{dt} \begin{bmatrix} \mathbf{\Psi}_{0dq} \\ \mathbf{\Psi}_{FDQ} \end{bmatrix} + \begin{bmatrix} \mathbf{P} \times \mathbf{e}_{\mathbf{n}} \\ \mathbf{\Psi}_{FDQ} \end{bmatrix} \\ \mathbf{e}_{\mathbf{n}} \stackrel{\Delta}{=} \begin{bmatrix} e_{n} \\ e_{n} \\ e_{n} \end{bmatrix} = -(R_{n}i_{n} + L_{n} \frac{di_{n}}{dt}) \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} = -\begin{bmatrix} R_{n} & R_{n} & R_{n} \\ R_{n} & R_{n} & R_{n} \\ R_{n} & R_{n} & R_{n} \end{bmatrix} \begin{bmatrix} i_{a} \\ i_{b} \\ i_{c} \end{bmatrix} - \begin{bmatrix} L_{n} & L_{n} & L_{n} \\ L_{n} & L_{n} & L_{n} \\ L_{n} & L_{n} & L_{n} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} i_{a} \\ i_{b} \\ i_{c} \end{bmatrix} \\ = -\mathbf{R}_{\mathbf{n}}\mathbf{i}_{\mathbf{abc}} - \mathbf{L}_{\mathbf{n}} \frac{d}{dt} \mathbf{i}_{\mathbf{abc}}$$

$$i_b$$
 i_a
 i_a
 i_a
 i_n
 e_n
 $L_n R_n =$
 i_c

$$\mathbf{P} \times \mathbf{e_n} = \mathbf{P} \left(-\mathbf{R_n} \, \mathbf{i_{abc}} - \mathbf{L_n} \, d\mathbf{i_{abc}} / dt \right)$$

$$= -\mathbf{P} \, \mathbf{R_n} \, \mathbf{P}^{-1} \, \mathbf{P} \, \mathbf{i_{abc}} - \mathbf{P} \, \mathbf{L_n} \, \mathbf{P}^{-1} \, \mathbf{P} \, d\mathbf{i_{abc}} / dt$$

$$= -\mathbf{P} \, \mathbf{R_n} \, \mathbf{P}^{-1} \, \mathbf{i_{0dq}} - \mathbf{P} \, \mathbf{L_n} \, \mathbf{P}^{-1} \, (d\mathbf{i_{0dq}} / dt - d\mathbf{P} / dt \times \mathbf{i_{abc}})$$

$$\mathbf{P} \, \mathbf{L_n} \, \mathbf{P}^{-1} = \begin{bmatrix} 3L_n & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}$$

$$= -\begin{bmatrix} 3R_n i_0 \\ 0 \\ 0 \end{bmatrix} -\begin{bmatrix} 3L_n \, di_0 / dt \\ 0 \\ 0 \end{bmatrix}$$

$$= \begin{bmatrix} \mathbf{R}_{0dq} & \mathbf{0} \\ \mathbf{0} & \mathbf{R}_{FDQ} \end{bmatrix} \begin{bmatrix} -\mathbf{i}_{0dq} \\ \mathbf{i}_{FDQ} \end{bmatrix} + \frac{d}{dt} \begin{bmatrix} \Psi_{0dq} \\ \Psi_{FDQ} \end{bmatrix} + \begin{bmatrix} -3R_n i_0 - 3L_n \, \frac{d}{dt} \, i_0 \\ -\omega_r \psi_q \\ -\frac{\omega_r \psi_q}{\mathbf{0}_{3\times 1}} \end{bmatrix}$$

$$\mathbf{Note:} \, \mathbf{P} \, \mathbf{L_n} \, \mathbf{P}^{-1} \times d\mathbf{P} / dt \times \mathbf{i_{abc}} = \mathbf{0}$$

Summary: Flux and Voltage Equations



Winding Circuits (Equivalent Transformers) after Park's Transformation

$\int e_0$]	$\int R_a$	0	0	0	0	0	$\begin{bmatrix} -i_0 \end{bmatrix}$]	$\int L_0$	0	0	0	0	0 -		$\begin{bmatrix} -i_0 \end{bmatrix}$		[0]
e_d		0	R_a	0	0	0	0	$-i_d$		0	L_{d}	kM _F	kM _D	0	0		$-i_d$		$-\omega_r \psi_q$
e_{F}		0	0	R_{F}	0	0	0	i_F		0	kM _F	L_{F}	M_{R}	0	0	d	i _F		0
0	=	0	0	0	R_{D}	0	0	i_D	+	0	kM _D	M_{R}	L_{D}	0	0	dt	i _D	+	0
e_q		0	0	0	0	R_a	0	$-i_q$		0	0	0	0	L_q	kM_{Q}		$-i_q$		$\omega_r \psi_d$
0		0	0	0	0	0	R_Q	i_Q		0	0	0	0	kM_Q	L_Q		i_Q		0



 $\omega_r \psi_d$ is a speed voltage caused by d-axis flux in the q-axis winding $-\omega_r \psi_d$ is a speed voltage caused by q-axis flux in the d-axis winding

Voltage equations follow Faraday's law

$$\mathcal{E} = \frac{1}{q} \oint_{\text{wire}} \mathbf{F} \cdot d\boldsymbol{\ell} = \oint_{\text{wire}} (\mathbf{E} + \mathbf{v} \times \mathbf{B}) \cdot d\boldsymbol{\ell}$$

$$\begin{array}{c}
R_{a} & X_{s} \\
& & & \\
& & & \\
& & & \\
\end{array} \\
E_{q} \angle \delta_{i} \\
& & & \\
& & & \\
& & & \\
\end{array} \\
E_{q} = X_{ad} i_{fd} \\
& & \\
& & & \\
& & & \\
& & & \\
\end{array} \\
= X_{d} = X_{g} = X_{s}
\end{array}$$

Naming Convention

My Notes	Kundur's book
i_F	i_{fd}
i_D	i_{kd}
i_Q	i_{kq}
L_s	L_{aa0}
M_{s}	L_{ab0}
L_m	L_{aa2}
M_F	L_{afd}
M_D	L_{akd}
M_Q	L_{akq}
M_R	L_{fkd}
L_F	L_{ffd}
L_D	L_{kkd}
L_Q	L_{kkq}

Alternative Park's Transformation

• If $k_d = k_q = 2/3$ and $k_0 = 1/3$, i_a , i_b , i_c , i_d , i_q all have the same amplitude I_m , or in other words, a unit-to-unit relationship holds

$$i_{a} = I_{m} \sin \omega_{s} t$$

$$i_{b} = I_{m} \sin(\omega_{s} t - 2\pi/3)$$

$$\tilde{\mathbf{P}} = \frac{2}{3} \begin{bmatrix} 1/2 & 1/2 & 1/2 \\ \cos \theta & \cos(\theta - \frac{2\pi}{3}) & \cos(\theta + \frac{2\pi}{3}) \\ -\sin \theta & -\sin(\theta - \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \end{bmatrix}$$

$$\tilde{\mathbf{P}}^{-1} = \begin{bmatrix} 1 & \cos \theta & -\sin \theta \\ 1 & \cos(\theta - \frac{2\pi}{3}) & -\sin(\theta - \frac{2\pi}{3}) \\ 1 & \cos(\theta + \frac{2\pi}{3}) & -\sin(\theta + \frac{2\pi}{3}) \end{bmatrix}$$

$$i_{d} = k_{d} I_{m} \sin(\omega_{s} t - \theta) \times 3/2 = I_{m} \sin(\omega_{s} t - \theta)$$

$$i_{q} = -k_{q} I_{m} \cos(\omega_{s} t - \theta) \times 3/2 = -I_{m} \cos(\omega_{s} t - \theta)$$

$$i_{0} = k_{0}(i_{a} + i_{b} + i_{c})$$

• However, $\tilde{\mathbf{P}}$ is not isometric. An isometric \mathbf{P} that preserves distances or inner products is orthogonal: $\mathbf{P}^T \mathbf{P} = \mathbf{I}$.

$$P_{3\phi} = e_a i_a + e_b i_b + e_c i_c = \mathbf{e}_{abc}^T \mathbf{i}_{abc} = \mathbf{e}_{0dq}^T \mathbf{\tilde{P}}^{-T} \mathbf{\tilde{P}}^{-1} \mathbf{i}_{0dq} = 3e_0 i_0 + \frac{3}{2}e_d i_d + \frac{3}{2}e_q i_q \qquad \neq e_0 i_0 + e_d i_d + e_q i_q$$

• L_{ad} - L_{aq} based per unit system: all per unit mutual inductances between the stator and rotor circuits are equal to \overline{L}_{ad} in d-axis or \overline{L}_{aq} in q-axis, where $\overline{L}_{ad} = \overline{L}_d - \overline{L}_l$ and $\overline{L}_{aq} = \overline{L}_q - \overline{L}_l$

• This reduces the number of constant parameters in model without losing its accuracy.

$$\begin{bmatrix} \psi_{0} \\ \psi_{d} \\ \psi_{q} \\ \psi_{F} \\ \psi_{D} \\ \psi_{Q} \end{bmatrix} = \begin{bmatrix} L_{0} & 0 & 0 & 0 & 0 & 0 \\ 0 & L_{d} & 0 & M_{F} & M_{D} & 0 \\ 0 & 0 & L_{q} & 0 & 0 & M_{Q} \\ 0 & \frac{3}{2}M_{F} & 0 & L_{F} & M_{R} & 0 \\ 0 & \frac{3}{2}M_{D} & 0 & M_{R} & L_{D} & 0 \\ 0 & 0 & \frac{3}{2}M_{Q} & 0 & 0 & L_{Q} \end{bmatrix} \begin{bmatrix} -i_{0} \\ -i_{d} \\ -i_{q} \\ i_{F} \\ i_{D} \\ i_{Q} \end{bmatrix} = \begin{bmatrix} \overline{L}_{0} & 0 & 0 & 0 & 0 & 0 \\ 0 & \overline{L}_{I} + \overline{L}_{dd} & 0 & \overline{L}_{dd} & \overline{L}_{dd} & 0 \\ 0 & 0 & \overline{L}_{I} + \overline{L}_{dd} & 0 & \overline{L}_{dd} & 0 \\ 0 & \overline{L}_{dd} & 0 & \overline{L}_{R} & \overline{M}_{R} & 0 \\ 0 & \overline{L}_{dd} & 0 & \overline{M}_{R} & \overline{L}_{D} & 0 \\ 0 & 0 & \overline{L}_{dd} & 0 & \overline{M}_{R} & \overline{L}_{D} & 0 \\ 0 & 0 & \overline{L}_{dq} & 0 & 0 & \overline{L}_{Q} \end{bmatrix} \begin{bmatrix} -\overline{i}_{0} \\ -\overline{i}_{d} \\ -\overline{i}_{d} \\ -\overline{i}_{d} \\ -\overline{i}_{d} \\ \overline{i}_{F} \\ \overline{i}_{D} \\ \overline{i}_{Q} \end{bmatrix}$$

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Per Unit Representation

- Each of stator/rotor winding circuit involves quantities, e.g. e, i, S, Z and f, but only three of them are independent.
- Using the machine ratings (stator-side) as the base values

peak value of rated line-to-neutral voltage $e_{s \ base} (V)$ $i_{s base}$ (A) peak value of rated line current f_{base} (Hz) rated frequency $S_{3\phi base}$ (VA) = $3 \frac{e_{s base}}{\sqrt{2}} \times \frac{\dot{i}_{s base}}{\sqrt{2}} = \frac{3}{2} e_{s base} \times \dot{i}_{s base}$ ω_{base} (elec. rad/s)= $2\pi f_{base} = \omega_{mbase} \times (\frac{p}{2})$ e.g. ≈ 377 (rad/s) for 60Hz $t_{base}(s) = 1/\omega_{base} = 1/(2\pi f_{base})$ ≈0.0027 sec. $Z_{s base}(\Omega) = e_{s base}/i_{s base}$ $L_{s base}$ (H) = $Z_{s base}/\omega_{base}$ $\psi_{s base}$ (Wb·turns)= $L_{s base} \times i_{s base} = e_{s base} / \omega_{base}$ $T_{base} (N \cdot m) = \frac{S_{3\phi \ base}}{\omega_{mbase}} = \frac{3}{2} \left(\frac{p}{2}\right) \psi_{s \ base} \times i_{s \ base}$

$$\bar{X}(p.u.) = \frac{X}{Z_{base}} = \frac{2\pi f}{2\pi f_{base}} \times \frac{L}{L_{base}}$$
If $f=f_{base}$ $\overline{X} = \overline{L}$

Bases	d	q	0	F	D	Q				
1				f_{base}	f_{base}					
2	e _{s base}			$S_{3\phi base}$						
3	$i_{s \ base}$			i _{F base}	i _{Q base}					

 i_{Fbase} , i_{Dbase} and i_{Qbase} should enable a symmetric per-unit inductance matrix.

$$\begin{aligned} \mathbf{L}_{ad} - \mathbf{L}_{aq} \text{ based Per Unit System} \\ \begin{bmatrix} w_{a} \\ w_{b} \\ w_{b}$$

 \overline{M}_{F} \overline{M}_{D}

$$\begin{bmatrix} \overline{\psi}_{a} \\ \overline{\psi}_{q} \\ \overline{\psi}_{q$$

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Formal Simplification of the Model

21 inductances (many are θ -dependent)



14 constant inductances



8 constant inductances

$$\begin{array}{ll} L_d - L_l, \ M_F, \ M_D, \ \dots \ \rightarrow \ L_{ad} \\ L_q - L_l, \ M_R, \ \dots \ & \rightarrow \ L_{aq} \end{array}$$

All mutual inductances in p.u. are equal to L_{ad} or L_{aq} , linked to the same element in an equivalent circuit.

e_a e_b e_c e_F 0	$ \begin{array}{c} R_a \\ 0 \\ 0 \\ 0 \\ 0 \end{array} $	0 R_a 0 0 0 0	0 0 R_a 0 0 0	0 0 0 R_{1} 0	Ŧ	0 0 0 0 8 ₀	0 0 0 0 0	$\begin{vmatrix} -i_a \\ -i_b \\ -i_c \\ i_F \\ i_D \end{vmatrix}$	+	$\begin{vmatrix} l_{aa} \\ l_{ba} \\ l_{ca} \\ l_{Fa} \\ l_{Da} \end{vmatrix}$	$egin{array}{llllllllllllllllllllllllllllllllllll$	$l_{ac} l_{aF}$ $l_{bc} l_{bF}$ $l_{cc} l_{cF}$ $l_{Fc} l_{FI}$ $l_{Dc} l_{DI}$	$ \begin{array}{c} & l_{aD} \\ & l_{bD} \\ & l_{bD} \\ & l_{cD} \\ & l_{cD} \\ & & l_{FD} \\ & & & l_{DD} \end{array} $	l_{aQ} l_{bQ} l_{cQ} l_{FQ} l_{DO}	$\frac{d}{dt}$	$\begin{vmatrix} -i_a \\ -i_b \\ -i_c \\ i_F \\ i_D \end{vmatrix}$	-		
0	0	0	0	0		0	R_Q	i_Q		l_{Qa}	l_{Qb}	l_{Qc} l_{QI}	l_{QD}	l_{QQ}		i_Q]		
$\begin{bmatrix} e_0 \\ e_d \\ e_F \\ 0 \\ e_q \\ 0 \end{bmatrix}$	<i>R_a</i> 0 0 0 0 0	0 <i>R_a</i> 0 0 0 0	$\begin{array}{c} 0\\ 0\\ R_F\\ 0\\ 0\\ 0\\ 0\\ \end{array}$	$\begin{array}{c} 0\\ 0\\ 0\\ R_D\\ 0\\ 0\\ 0 \end{array}$	$ \begin{array}{c} 0 \\ 0 \\ 0 \\ R_a \\ 0 \end{array} $	$\begin{array}{c} 0 \\ 0 \\ 0 \\ 0 \\ 0 \\ R_{\varrho} \end{array}$	$\left \begin{array}{c} -i_0\\ -i_d\\ i_F\\ i_D\\ -i_q\\ i_Q \end{array}\right $	$\left \begin{array}{c} L\\ 0\\ 0\\ 0\\ 0\\ 0\\ 0\\ 0\\ 0\end{array}\right $	″0)))	0 L_d kM_F kM_D 0 0	0 kM _F L _F M _R 0 0	0 kM_{D} M_{R} L_{D} 0 0	$0\\0\\0\\L_q\\kM_Q$	$\begin{array}{c} 0\\ 0\\ 0\\ kM_{\varrho}\\ L_{\varrho} \end{array}$	$\left \frac{d}{dt} \right $	$egin{bmatrix} -i_0\ -i_d\ i_F\ i_D\ -i_q\ i_Q\ \end{bmatrix}$	+	$ \begin{array}{c} 0\\ -\omega_r\psi_q\\ 0\\ 0\\ \omega_r\psi_d\\ 0 \end{array} $	
$\begin{bmatrix} e_0 \\ e_d \\ e_F \end{bmatrix}$	<i>R_a</i> 0 0	0 <i>R_a</i> 0	0 0 <i>R</i> _F	0 0 0	0 0 0	0 0 0	$\begin{bmatrix} -i_0 \\ -i_d \\ i_F \end{bmatrix}$	$\begin{bmatrix} L_{0} \\ 0 \\ 0 \end{bmatrix}$	0)	$egin{array}{c} 0 \ L_l + L_{ad} \ L_{ad} \end{array}$	$egin{array}{ccc} 0 \ L_a \ L_1 \end{array}$	$\begin{array}{c} 0 \\ 0 \\ L_{ad} \\ F \end{array} M_{R} \end{array}$	0 0 0	0 0 0	$\Big] d$	$\begin{bmatrix} -i_0 \\ -i_d \\ i_F \end{bmatrix}$		$ \begin{array}{c} 0\\ -\omega_r\psi\\ 0 \end{array} $	q

Equivalent Circuits

Define differential operator p = d/dt

$$\begin{bmatrix} e_{0} \\ e_{d} \\ e_{r} \\ e_{p} = 0 \end{bmatrix} = \begin{bmatrix} R_{a} & 0 & 0 & 0 & 0 & 0 \\ 0 & R_{r} & 0 & 0 & 0 & 0 \\ 0 & 0 & R_{r} & 0 & 0 & 0 \\ 0 & 0 & 0 & R_{p} & 0 & 0 \\ 0 & 0 & 0 & 0 & R_{a} & 0 \\ 0 & 0 & 0 & 0 & 0 & R_{a} \end{bmatrix} \begin{bmatrix} -i_{0} \\ -i_{d} \\ i_{r} \\ i_{p} \\ i_{q} \end{bmatrix} + \begin{bmatrix} L_{0} & 0 & 0 & 0 & 0 & 0 \\ 0 & L_{ad} & L_{r} & M_{R} & 0 & 0 \\ 0 & L_{ad} & M_{R} & L_{D} & 0 & 0 \\ 0 & 0 & 0 & 0 & L_{r+L_{ad}} & L_{ad} \\ 0 & 0 & 0 & 0 & L_{aq} & L_{0} \end{bmatrix} \begin{bmatrix} -i_{0} \\ -i_{d} \\ i_{p} \\ i_{p} \\ i_{q} \end{bmatrix} + \begin{bmatrix} 0 \\ -\omega_{r}\psi_{q} \\ 0 \\ 0 \\ \omega_{r}\psi_{d} \end{bmatrix}$$

$$e_{d} = (L_{l}+L_{ad}) \times (-pi_{d}) + L_{ad} \times pi_{F} + L_{ad} \times pi_{D} - \omega_{r}\psi_{q} - R_{a}i_{d}$$

$$= L_{ad} \times p(-i_{d}+i_{F}+i_{D}) - L_{i} \times pi_{d} - \omega_{r}\psi_{q} - R_{a}i_{d}$$

$$p \psi_{d} = (L_{l}+L_{ad}) \times (-pi_{d}) + L_{ad} \times pi_{F} + L_{ad} \times pi_{D} = e_{d} + \omega_{r}\psi_{q} + R_{a}i_{d}$$

$$e_{F} = L_{ad} \times (-pi_{d}) + L_{F} \times pi_{F} + M_{R} \times pi_{D} + R_{F}i_{F}$$

$$= L_{ad} \times p(-i_{d}+i_{F}+i_{D}) - L_{ad} \times p(i_{D}+i_{F}) + M_{R} \times pi_{D} + L_{F} \times pi_{F} + R_{F}i_{F}$$

$$= L_{ad} \times p(-i_{d}+i_{F}+i_{D}) + (M_{R}-L_{ad}) \times p(i_{D}+i_{F}) + (L_{F}-M_{R}) \times pi_{F} + R_{F}i_{F}$$

$$e_{D} = 0 = L_{ad} \times (-pi_{d}) + M_{R} \times pi_{F} + L_{D} \times pi_{D} + R_{D}i_{D}$$

$$= L_{ad} \times p(-i_{d}+i_{F}+i_{D}) + (M_{R}-L_{ad}) \times p(i_{D}+i_{F}) + (L_{D}-M_{R}) \times pi_{D} + R_{D}i_{D}$$

$$e_{q} = (L_{l}+L_{aq}) \times (-pi_{q}) + L_{aq} \times pi_{Q} + \omega_{r}\psi_{d} - R_{a}i_{q}$$

$$= L_{aq} \times p(-i_{q}+i_{Q}) - L_{l} \times pi_{q} + \omega_{r}\psi_{d} - R_{a}i_{q}$$

$$p\psi_{q} = (L_{l}+L_{aq}) \times (-pi_{q}) + L_{aq} \times pi_{Q} = e_{q} - \omega_{r}\psi_{d} + R_{a}i_{q}$$

 $e_{Q} = 0 = L_{aq} \times (-pi_{q}) + L_{Q} \times pi_{Q} + R_{Q}i_{Q}$ $= L_{aq} \times p(-i_{q} + i_{Q}) + (L_{Q} - L_{aq}) \times pi_{Q} + R_{Q}i_{Q}$

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Equivalent Circuits with Multiple Damper Windings (e.g. Round-rotor Machines)

• d-axis:

$$L_{fd} \triangleq L_F - M_R \qquad R_{fd} \triangleq R_F$$

$$\psi_{fd} \triangleq \psi_F \qquad e_{fd} \triangleq e_F$$

$$L_{1d} \triangleq L_D - M_R \qquad R_{1d} \triangleq R_D$$

$$\psi_{1d} \triangleq \psi_D$$

 M_R - $L_{ad} \approx 0$ is named L_{fkd1} in some literature to model rotor mutual flux leakage, i.e. the flux linking the rotor's field and damper windings but not stator windings)

• q-axis:

$$L_{1q} \triangleq L_Q - L_{aq} \qquad R_{1q} \triangleq R_Q$$

$$\psi_{1q} \triangleq \psi_Q$$

$$L_{2q} \triangleq L_G - L_{aq} \qquad R_{2q} \triangleq R_G$$

$$\psi_{2q} \triangleq \psi_G$$

Subscript Notations: ()_{fd} ~ field winding quantities ()_{kd} ~ k-th d-axis damper winding quantities ()_{kq} ~ k-th q-axis damper winding quantities



Example: a model with 3 rotor windings in each of d- and q-axis equivalent circuits

- EPRI Report EL-1424-V2, "Determination of Synchronous Machine Stability Study Constants, Volume 2", 1980
 - The proposed equivalent circuits are expected to contain sufficient details to model all machines
 - Parameters are estimated by frequency response tests



Figure S-2. Simulation of Nanticoke Generator Power and Field Current During a Transient Caused By Line Switching.

Example 3.1 (Kundur's book)

A 555 MVA, 24kV, 0.9p.f., 60Hz, 3 phase, 2 pole synchronous generator has the following inductances an resistances associated with the stator and field windings:

l_{aa} =3.2758+0.0458 cos(2 θ)	mН	l_{ab} =-1.6379-0.0458 cos(20+ $\pi/3$)	mН
l_{aF} =40.0 cos θ	mH	$L_F = 576.92$	mН
$R_a = 0.0031$	Ω	$R_F = 0.0715$	Ω

- a. Determine L_d and L_q in H
- b. If the stator leakage inductance L_l is 0.4129mH, determine L_{ad} and L_{aq} in H
- c. Using the machine rated values as the base values for the stator quantities, determine the per unit values of the following in the L_{ad} base reciprocal per unit system (assuming $L_{ad}=M_F=M_R$ in per unit): L_l , L_{ad} , L_{aq} , L_d , L_q , M_F , L_F , L_{fd} , R_a and R_F

Solution:

a.
$$l_{aa} = L_s + L_m \cos 2\theta$$
 =3.2758+0.0458 cos2 θ mH
 $l_{ab} = -M_s - L_m \cos(2\theta + \pi/3)$ = -1.6379-0.0458 cos(2 $\theta + \pi/3$) mH
 $l_{aF} = M_F \cos \theta$ =40.0 cos θ mH
 $L_d = L_s + M_s + 3L_m/2$ =3.2758+1.6379+ $\frac{3}{2} \times 0.0458$ =4.9825 mH
 $L_q = L_s + M_s - 3L_m/2$ =3.2758-1.6379+ $\frac{3}{2} \times 0.0458$ =4.8451 mH
b. $L_{ad} = L_d - L_l$ =4.9825 - 0.4129 =4.5696 mH
 $L_{aq} = L_q - L_l$ =4.4851 - 0.4129 =4.432 mH

3-phase VA _{base}	$_{e} = 555 \text{ MVA}$				
$E_{RMS\ base}$	$=24/\sqrt{3}$		=13.856 k	xV	
$e_{s base}$ (peak)	$=\sqrt{2} \times E_{RMS \ base} = \sqrt{2} \times 1$	3.856	=19.596 k	xV	
I _{RMS base}	=3-phaseVAbase / $(3 \times$	$E_{RMS \ base}$)=555×10 ⁶ /(3	×13.856×10) ³)=13.3512	×10 ³ A
<i>i_{s base}</i> (peak)	$=\sqrt{2} \times I_{RMS \ base} = \sqrt{2} \times 13$.3512×10 ³	=18.8815	×10 ³ A	
$Z_{s \ base}$	$=e_{s\ base}/i_{s\ base}$	$=19.596 \times 10^{3}/(18.881)$	5×10 ³)	=1.03784	Ω
ω_{base}	$=2\pi\times60$	=377 elec. rad/s			
$L_{s \ base}$	$=Z_{s\ base}/\omega_{base}$	$=1.03784/377 \times 10^{3}$		=2.753	mΗ
$i_{F \ base}$	$=L_{ad}/M_F \times i_{s\ base}$	=4.5696/40×18.8815	×10 ³	=2158.0	А
$e_{F base}$	=3-phaseVAbase/ <i>i</i> _{Fba}	$s_{ase} = 555 \times 10^{6} / 2158$		=257.183	kV
$Z_{F \ base}$	$= e_{F base} / i_{F base}$	=257.183×10 ³ /2158		=119.18	Ω
$L_{F \ base}$	$= Z_{F base} / \omega_{base}$	=119.18×10 ³ /377		=316.12	mΗ
$M_{F \ base}$	$= L_{S base} imes i_{S base} / i_{F base}$	use =2.753×188815/215	8	=241 mH	

Then per unit values are:

$L_l = L_l / L_{s \ base} = 0.4129 / 2.753$	=0.15 pu
L _{ad} =4.5696/2.753	=1.66 pu
L _{aq} =4.432/2.753	=1.61pu
$L_d = L_l + L_{ad} = 0.15 + 1.66$	=1.81 pu
$L_q = L_l + L_{aq} = 0.15 + 1.61$	=1.76 pu
$R_a = 0.0031/1.03784$	=0.003 pu

$R_F = R_F / Z_{F \ base} = 0.0715 / 119.18$	=0.0006 pu
$M_F = M_F / M_{F base} = 400/241$	=1.66 pu
$L_F = L_F / L_{F base} = 576.92 / 316.12$	=1.825 pu
$L_{fd} = L_F - M_R = L_F - L_{ad} = 1.825 - L_{ad}$	1.66 <i>=</i> 0.165 pu

c.

Steady-state Analysis

• All derivatives (*pX*) are zero:

 $p\omega_r=0 \rightarrow \omega_r=1$ and L=X in p.u. $p \psi_{fd} = 0 \rightarrow e_{fd} = R_{fd} i_{fd}$ $p \psi_{1d} = 0 \rightarrow i_{1d} = 0 \qquad \psi_d = -L_d i_d + L_{ad} i_{fd}$ $p \psi_{1q} = 0 \rightarrow i_{1q} = 0 \qquad \psi_q = -L_q i_q$ $p \psi_d = 0 \rightarrow e_d = -\psi_q - R_a i_d \implies e_d = L_q i_q - R_a i_d$ $p \psi_q = 0 \rightarrow e_q = \psi_d - R_a i_q$ $e_a = -L_d i_d + L_{ad} i_{fd} - R_a i_q$

e

Voltage and flux equations:	$: 6+6 \rightarrow 3+3$
$e_{fd} = R_{fd} i_{fd}$	$\psi_{fd} = (L_{ad} + L_{fd})i_{fd} - L_{ad}i_d$
$e_d = X_q i_q - R_a i_d$	$\psi_{1d} = L_{ad} (i_{fd} - i_d)$
$e_q = -X_d i_d + X_{ad} i_{fd} - R_a i_q$	$\psi_{1q} = \psi_{2q} = -L_{aq}i_q$

• Single equivalent circuit for both d and q axes:

 $\tilde{I}_t = i_d + ji_q$ $\tilde{E}_t = e_d + je_q = X_q i_q - R_a i_d - jX_d i_d + jX_{ad} i_{fd} - jR_a i_q$ $= -R_a(i_d + ji_q) - jX_q \times ji_q - jX_di_d + jX_{ad}i_{fd}$ $\widetilde{E}_t = \widetilde{E}_q - (R_a + jX_a)\widetilde{I}_t$ where $\widetilde{E}_q = j[X_{ad}i_{fd} - (X_d - X_a)i_d]$



Computing per-unit steady-state values

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 $\omega_r = 1 \text{p.u.}$

- **Active and Reactive Powers** • $S = \tilde{E}_t \tilde{I}_t^*$ $=(e_d+je_a)(i_d-ji_a)$ $= (e_d i_d + e_q i_q) + j(e_q i_d - e_d i_q)$ where $e_d = -\omega_r \psi_a - R_a i_d$ $e_a = \omega_r \psi_d - R_a i_a$ $P_t = e_d i_d + e_q i_q$ $=\omega_r(\psi_d i_q - \psi_q i_d) - R_a(i_d^2 + i_q^2)$ $= P_e - R_a (i_d^2 + i_a^2)$ $Q_t = e_a i_d - e_d i_a$
- Air-gap torque (or electric torque) • D

$$T_e = \frac{P_e}{\omega_r} = \psi_d i_q - \psi_q i_d$$

 $=P_{t}+R_{a}(i_{d}^{2}+i_{a}^{2})$ © 2021 Kai Sun



Representation of Magnetic Saturation

- Assumptions for stability studies
 - The leakage fluxes are not significantly affected by saturation of the iron portion, so L_l is constant and only L_{ad} and L_{ag} saturate in equivalent circuits
 - There is no magnetic coupling between d and q axes, so saturations on L_{ad} and L_{aq} can be modeled individually. Let L_{adu} and L_{aqu} denote their unsaturated values:

$$L_{ad}(i_{fd}) = L_{adu}, \quad L_{aq}(i_{fd}) = L_{aqu}$$
 only if i_{fd} or MMF is not large

- The leakage fluxes do not contribute to the iron saturation. Thus, saturation is determined only by the air-gap flux linkage
 - $\psi_{d} = -(L_{l} + L_{ad})i_{d} + L_{ad}i_{fd}$ $\psi_{q} = -(L_{l} + L_{aq})i_{q}$ $\psi_{at} = \psi_{ad} + j\psi_{aq}$ $= \psi_{d} + L_{l}i_{d} + j\psi_{q} + jL_{l}i_{q}$ $= L_{ad}(i_{fd} i_{d}) jL_{aq}i_{q}$ $\psi_{at} = \sqrt{\psi_{ad}^{2} + \psi_{aq}^{2}}$ First find ψ_{at} vs. i_{fd} and then estimate the function $L_{ad}(i_{fd})$ or $L_{aq}(i_{fd})$
- Saturation relationship ψ_{at} vs. i_{fd} (or *MMF*) under loaded conditions is assumed to be the same as under no-load conditions ($i_d=i_q=0$), so only the open-circuit characteristic (OCC) is considered.

$$\tilde{\psi}_{at} = \psi_{at} = L_{ad} i_{fd}$$



Estimating Saturation Factors K_{sd} and K_{sd}

$$L_{ad} = K_{sd} L_{adu}$$
 $L_{aq} = K_{sq} L_{aqu}$

- Salient pole machines
 - The path for *q*-axis flux is largely in air, so L_{aq} does not vary significantly with saturation of the iron portion of the path
 - Assume K_{sq} =1.0 for all loading conditions.
- Round rotor machines
 - There is a magnetic saturation in both axes, but the saturation data in *q* axis is usually not available
 - Assume $K_{sq} = K_{sd}$
- Thus, we focus on estimating K_{sd}

$$K_{sd} = L_{ad}/L_{adu}$$

= $\psi_{at}/\psi_{at0} = \psi_{at}/(\psi_{at} + \psi_{I})$
= I_0 / I



(See Kundur's Example 3.3 on Estimating K_{sd} for different loading conditions)

Modeling of the Saturation Characteristic

- ψ_I is modeled by 3 approximate functions
 - Segment I ($\psi_{at} < \psi_{T1}$): $\psi_I = \psi_{at0} \psi_{at} = 0$

- **Segment II**
$$(\psi_{T1} < \psi_{at} < \psi_{T2})$$
: $\psi_I = A_{sat} e^{B_{sat}} (\psi_{at} - \psi_{T1})$

- Segment III (
$$\psi_{at} > \psi_{T2}$$
): $\psi_I = \psi_{G2} + L_{ratio}(\psi_{at} - \psi_{T2}) - \psi_{at}$

- Note: segments I and II are not connected since when $\psi_{at} = \psi_{T1}$, $\psi_I = A_{sat} \neq 0$ (usually small)
- Segments II and III are assumed to be connected at $\psi_{at} = \psi_{T2}$ to solve ψ_{G2}

 $A_{sat}e^{B_{sat}(\psi_{T2}-\psi_{T1})} = \psi_{G2} + L_{ratio}(\psi_{T2}-\psi_{T2}) - \psi_{T2}$ $\psi_{G2} = \psi_{T2} + A_{sat}e^{B_{sat}(\psi_{T2}-\psi_{T1})}$



Five independent parameters: A_{sat} , B_{sat} , ψ_{T1} , ψ_{T2} and L_{ratio}
Synchronous Machine Model DG1S1

Model Descriptions

This model uses parameters in basic form and type 1 saturation model.

Data Format

IBUS, 'DG1S1', I, MVA, X_{ad}, X_{aq}, X_l, R_a, X_{fd}, R_{fd}, X_{kq1}, R_{kq1}, X_{kd1}, R_{kd1}, X_{kq2}, R_{kq2}, X_{kd2}, R_{kd2}, X_{kq3}, R_{kq3}, H, K_D, α , A_{sat}, B_{sat}, ψ _L, ψ _M, RS /

Parameter Descriptions

- IBUS Bus number, name, or generator equipment name of the machine.
- ID of the machine (may or may not be enclosed in single quotes).
- MVA MVA base of the machine. If not specified (i.e., no value or zero is entered), the MVA base of the matched generator in powerflow data will be used.
- Xad Unsaturated direct axis mutual reactance in per unit on machine MVA base.
- Unsaturated quadrature axis mutual reactance in per unit on machine MVA base.
- X₁ Leakage reactance in per unit on machine MVA base.
- Ra Armature resistance in per unit on machine MVA base.
- Xfd Field winding leakage reactance in per unit on machine MVA base.
- Rfd Field winding resistance in per unit on machines MVA base.
- Xkq1 First quadrature axis damper winding leakage reactance in per unit on machine MVA base.
- Rkq1 First quadrature axis damper winding resistance in per unit on machine MVA base.
- Xkd1 First direct axis damper winding leakage reactance in per unit on machine MVA base.
- Rkd1 First direct axis damper winding resistance in per unit on machine MVA base.
- Xkq2 Second quadrature axis damper winding leakage reactance in per unit on machine MVA base.
- Rkq2 Second quadrature axis damper winding resistance in per unit on machine MVA base.
- Xkd2 Second direct axis damper winding leakage reactance in per unit on machine MVA base.
- Rkd2 Second direct axis damper winding resistance in per unit on machine MVA base.
- Xkq3 Third quadrature axis damper winding leakage reactance in per unit on machine MVA base.
- Rkq3 Third quadrature axis damper winding resistance in per unit on machine MVA base.
- Inertia time constant of the machine in MW-second/MVA.
- K_D Damping coefficient in (p.u. torque)/(p.u. speed deviation).

 A_{sat}
 - Coefficient in saturation characteristic.

 B_{sat}
 - Coefficient in saturation characteristic.

 ψ_L
 - Flux linkage on the saturation curve where the saturation curv

- Flux linkage on the saturation curve where the Region II characteristic starts.
- Flux linkage on the saturation curve where the Region III characteristic starts.
- RS Ratio of the slopes of air-gap line and the Region III characteristic.







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5.25 Estimating saturation as an exponential function.

$$S(1.0) = S_{G1} = \frac{i_{F1} - i_{F0}}{i_{F0}}$$

$$S(1.2) = S_{G2} = \frac{i_{F3} - i_{F2}}{i_{F2}} = \frac{i_{F3} - 1.2i_{F0}}{1.2i_{F0}}$$

$$S_G V_t = A_G e^{B_G (V_t - 0.8)} \qquad \psi_I = A_{sat} e^{B_{sat} (\Psi_{at} - \Psi_{T1})}$$

$$A_G = S_{G1}^2 / 1.2S_{G2} \qquad A_{sat} = A_G, B_{sat} = B_G, \ \psi_{T1} = 0.8, \ \psi_{T2} = 1.2$$

$$B_G = 5 \ln (1.2S_{G2} / S_{G1}) \qquad L_{ratio} = L_{ad} (i_{F3}) / L_{adu}$$

CONs	#	Value	Description
J			T' _{do} (>0) (sec)
J+1			T" _{do} (>0) (sec)
J+2			T" _{qo} (>0) (sec)
J+3			H, Inertia
J+4			D, Speed damping
J+5			X _d
J+6			Xq
J+7			X'd
J+8			X" _d = X" _q
J+9			X
J+10			S(1.0)
J+11			S(1.2)

Note: X_d, X_q, X'_d, X"_d, X"_q, X_l, H, and D are in pu, machine MVA base.

X"_q must be equal to X"_d.

STATEs	#	Description
K		E′q
K+1		Ψ″q
K+2		ψkd
K+3		∆ speed (pu)
K+4		Angle (radians)

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IBUS, 'GENSAE', ID, CON(J) to CON(J+11) /

Example 3.2 in Kundur's Book

The following are the parameters in per unit on machine rating of a 555 MVA, 24 kV, 0.9 p.f., 60 Hz, 3600 RPM turbine-generator¹:

L _{ad} = 1.66	$L_{aq} = 1.61$	$L_l = 0.15$	$R_a = 0.003$
$L_{fd} = 0.165$	$R_{fd} = 0.0006$	$L_{1d} = 0.1713$	$R_{1d} = 0.0284$
$L_{1q} = 0.7252$	$R_{1q} = 0.00619$	$L_{2q} = 0.125$	$R_{2q} = 0.02368$
M_R is assumed to	be equal to L_{ad} .		

- (a) When the generator is delivering rated MVA at 0.9 p.f. (lag) and rated terminal voltage, compute the following:
- (i) Internal angle δ_i in electrical degrees
- (ii) Per unit values of e_d , e_q , i_d , i_q , i_{1d} , i_{1q} , i_{2q} , i_{fd} , e_{fd} , ψ_{fd} , ψ_{1d} , ψ_{1q} , ψ_{2q}
- (iii) Air-gap torque T_e in per unit and in newton-meters

Assume that the effect of magnetic saturation at the given operating condition is to reduce L_{ad} and L_{ag} to 83.5% of the values given above.

(b) Compute the internal angle δ_i and field current i_{fd} for the above operating condition, using the approximate equivalent circuit of Figure 3.22. Neglect R_a .

Solution

(a) With the given operating condition, the per unit values of terminal quantities are

P = 0.9, Q = 0.436, $E_t = 1.0$, $I_t = 1.0$, $\phi = 25.84^\circ$

 $R_{1d} >> R_{fd}$ $L_{fd}/R_{fd} >> L_{1d}/R_{1d}$ $L_{1q}/R_{1q} >> L_{2q}/R_{2q}$

 $\frac{X_q I_t \cos \phi - R_a I_t \sin \phi}{E_t + R_a I_t \cos \phi + X_a I_t \sin \phi}$ $\delta_i = \tan^{-1}$



The saturated values of the inductances are

$$\begin{split} L_{ad} &= 0.835 \times 1.66 = 1.386 \\ L_{aq} &= 0.835 \times 1.61 = 1.344 \\ L_{d} &= L_{ad} + L_{l} = 1.386 + 0.15 = 1.536 \\ L_{q} &= L_{aq} + L_{l} = 1.344 + 0.15 = 1.494 \end{split}$$

Following the procedure outlined in Section 3.6.5,

(i)
$$\delta_i = \tan^{-1} \left(\frac{1.494 \times 1.0 \times 0.9 - 0.003 \times 1.0 \times 0.436}{1.0 + 0.003 \times 1.0 \times 0.9 + 1.494 \times 1.0 \times 0.436} \right)$$

 $= \tan^{-1} (0.812) = 39.1$ electrical degrees
(ii) $e_d = E_t \sin \delta_i = 1.0 \sin 39.1 = 0.631$ pu
 $e_q = E_t \cos \delta_i = 1.0 \cos 39.1 = 0.776$ pu
 $i_d = I_t \sin(\delta_i + \phi) = 1.0 \sin(39.1 + 25.84) = 0.906$ pu
 $i_q = I_t \cos(\delta_i + \phi) = 1.0 \cos(39.1 + 25.84) = 0.423$ pu

$$i_{fd} = \frac{e_q + R_a i_q + X_d i_d}{X_{ad}}$$

= $\frac{0.776 + 0.003 \times 0.423 + 1.536 \times 0.906}{1.386}$
= 1.565 pu



 $e_{fd} = R_{fd}i_{fd} = 0.0006 \times 1.565$ = 0.000939 pu $\psi_{fd} = (L_{ad} + L_{fd})i_{fd} - L_{ad}i_d$ = (1.386+0.165)×1.565-1.386×0.907 = 1.17 pu $\psi_{1d} = L_{ad}(i_{fd} - i_d)$ = 1.386×(1.565-0.906) = 0.913 pu $\psi_{1q} = \psi_{2q} = -L_{aq}i_q = -1.344 \times 0.423$ = -0.569 pu

Under steady state,

$$i_{1d} = i_{1q} = i_{2q} = 0$$
 40

(iii) Air-gap torque

 $T_{e} = P_{t} + I_{t}^{2} R_{a}$ = 0.9 + 1.0²×0.003 = 0.903 pu $T_{base} = \frac{MVA_{base} \times 10^{6}}{\omega_{m \, base}}$ = $\frac{555 \times 10^{6}}{2\pi 60} = 1.472 \times 10^{6}$ N·m

Therefore,

$$T_e = 0.903 \times 1.472 \times 10^6$$

= 1.329×10⁶ N·m

(b) Using the saturated value of X_{ad} ,

 $E_q = X_{ad}i_{fd} = 1.386i_{fd}$

 $X_s = X_{ad} + X_l = 1.386 + 0.15 = 1.536$

From the equivalent circuit of Figure 3.22, with \tilde{E}_t as reference phasor,



Sub-transient and Transient Analysis

- Following a disturbance, currents are induced in rotor circuits. Some of these induced rotor currents decay more rapidly than others.
 - <u>Sub-transient parameters</u>: influencing rapidly decaying (cycles) components
 - <u>Transient parameters</u>: influencing the slowly decaying (seconds) components
 - <u>Synchronous parameters</u>: influencing sustained (steady state) components



Figure 3.27 Fundamental frequency component of armature current

Transient Phenomena

- Study transient behavior of a simple RL circuit

$$Ri(t) + L\frac{di(t)}{dt} = v(t)$$

• Apply Laplace Transform

Steady-state

(sinusoidal component)

dc offset

(transient component)

• Apply Inverse Laplace Transform to *I*(*s*)

$$i(t) = I_m \sin(\omega t + \alpha - \gamma) - I_m e^{-t/\tau} \sin(\alpha - \gamma)$$

where $I_m = V_m / \sqrt{R^2 + X^2}$, $\gamma = \tan^{-1}(\omega L/R)$







(time to decay to $1/e \approx 37\%$)

Saadat's Example 8.2: Balanced 3-Phase Short Circuit

A 500-MVA, 30-kV, 60-Hz synchronous generator operates at no-load with a constant excitation voltage of 400V. A 3-phase short circuit occurs at its armature terminals. Use its voltage equations to find transient waveforms for the current in each phase and the field current. Assume the short circuit to be applied at the instant when $\theta=0$, i.e. the daxis is along the magnetic axis of phase a. Also assume the rotor speed $\omega_r = \omega_s$.

Solution:

Solution: The voltage equations are: $\mathbf{e} = \mathbf{R} \times \mathbf{i} + \mathbf{L} \frac{d}{dt} \mathbf{i}$

$egin{array}{c} e_0 \ e_d \ e_F \ 0 \ e_q \end{array}$	=	$\begin{bmatrix} R_a + 3R_n \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}$	$egin{array}{c} 0 \ R_a \ 0 \ 0 \ \omega_r L_d \end{array}$	0 0 R_{F} 0 $\omega_{r}kM_{F}$	0 0 0 R_D $\omega_r kM_D$	$0\\-\omega_r L_q\\0\\0\\R_a$	$ \begin{array}{c} 0\\ -\omega_r kM_Q\\ 0\\ 0\\ 0\\ 0\\ 0 \end{array} $	$\begin{vmatrix} -i_0 \\ -i_d \\ i_F \\ i_D \\ -i_q \end{vmatrix} + \begin{vmatrix} -i_q \\ -i_q \end{vmatrix}$	$L_0 + 3L_n$ 0 0 0 0 0 0	0 L_d kM_F kM_D 0	$0 \\ kM_F \\ L_F \\ M_R \\ 0$	$0 \\ kM_D \\ M_R \\ L_D \\ 0$	$egin{array}{c} 0 \ 0 \ 0 \ 0 \ L_q \end{array}$	0 0 0 <i>kM_Q</i>	$\frac{d}{dt}$	$-i_0$ $-i_d$ i_F i_D $-i_q$
e_q 0		0	$\omega_r L_d$	$\omega_r kM_F$	$\omega_r kM_D$	$R_a = 0$	0 R_Q	$\left[\begin{array}{c} -i_q \\ i_Q \end{array} \right] \left[\begin{array}{c} -i_q \\ i_Q \end{array} \right]$	0 0	0 0	0 0	0 0	$L_q \ kM_Q$	kM_Q L_Q		$-i_q$ i_Q

 $i_a(0^+)=i_b(0^+)=i_c(0^+)=0 \rightarrow i_0(0^+)=i_d(0^+)=i_q(0^+)=0$ No-load condition: $e_a \equiv e_b \equiv e_c \equiv 0, \qquad \rightarrow e_0 \equiv e_d \equiv e_a \equiv 0$ Balanced 3-phase fault: Initial rotor winding currents: $i_F(0^+) = e_F/R_F = 400/0.4 = 1000 \text{ A}, \quad i_D(0^+) = i_O(0^+) = 0$

Then, solve the differential equations $\mathbf{i}(t)$ with these initial values $\mathbf{i}(0^+)$: $\frac{d}{dt}\mathbf{i} = \mathbf{L}^{-1}(\mathbf{e} - \mathbf{R} \times \mathbf{i})$

Generator Parameters for Ex	ample 8.2
$L_d = 0.0072 \text{ H}$ $L_q = 0.0070 \text{ H}$	$L_F = 2.500 \text{ H}$
$L_D = 0.0068 \text{ H}$ $L_Q = 0.0016 \text{ H}$	$M_F = 0.100 \text{ H}$
$M_D = 0.0054 \text{ H}$ $M_Q = 0.0026 \text{ H}$	$M_R = 0.125 \text{ H}$
$r = 0.0020 \ \Omega$ $r_F = 0.4000 \ \Omega$	$r_D=0.015~\Omega$
$r_Q = 0.0150 \ \Omega$ $L_0 = 0.0010 \ \mathrm{H}$	in the month of h





Saadat's Example 8.4: Line-to-Ground Short Circuit

Same as 8.2 except for a line-to-ground short circuit of phase *a*. Solution: No-load condition: $i_b \equiv i_c \equiv 0$ Line-to-Ground fault: $e_a \equiv 0$ $e_a = 0 = -R_a i_a + \frac{d}{dt} (-l_{aa} i_a - l_{ab} i_b - l_{ac} i_c + l_{aF} i_F + l_{aD} i_D + l_{aQ} i_Q)$ $= -R_a i_a - l_{aa} \frac{di_a}{dt} + l_{aF} \frac{di_F}{dt} + l_{aD} \frac{di_D}{dt} + l_{aQ} \frac{di_Q}{dt} - i_a \frac{dl_{aa}}{dt} + i_F \frac{dl_{aF}}{dt} + i_D \frac{dl_{aD}}{dt} + i_Q \frac{dl_{aQ}}{dt}$ $= -R_a i_a - (L_s + L_m \cos 2\theta) \frac{di_a}{dt} + M_F \cos \theta \frac{di_F}{dt} + M_D \cos \theta \frac{di_D}{dt} - M_Q \sin \theta \frac{di_Q}{dt} - (-2\omega_r L_m \sin 2\theta) \cdot i_a - \omega_r M_F \sin \theta \cdot i_F - \omega_r M_D \sin \theta \cdot i_D - \omega_r M_Q \cos \theta \cdot i_D$ Solve these voltage equations:

$$\begin{bmatrix} e_{a}=0\\ e_{F}\\ e_{D}=0\\ e_{Q}=0 \end{bmatrix} = \begin{bmatrix} R_{a}-2\omega_{r}L_{m}\sin 2\theta & -\omega_{r}M_{F}\sin \theta & -\omega_{r}M_{D}\sin \theta & -\omega_{r}M_{Q}\cos \theta\\ -\omega_{r}M_{F}\sin \theta & R_{F} & 0 & 0\\ -\omega_{r}M_{D}\sin \theta & 0 & R_{D} & 0\\ -\omega_{r}M_{Q}\cos \theta & 0 & 0 & R_{Q} \end{bmatrix} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+L_{m}\cos 2\theta & M_{F}\cos \theta & -M_{Q}\sin \theta\\ M_{F}\cos \theta & L_{F} & M_{R} & 0\\ M_{D}\cos \theta & M_{R} & L_{D} & 0\\ -M_{Q}\sin \theta & 0 & 0 & L_{Q} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+M_{s}+\frac{2}{2}L_{m}, \\ L_{q}=L_{s}+M_{s}-\frac{3}{2}L_{m} \\ -M_{Q}\sin \theta & 0 & 0 & L_{Q} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+M_{s}+\frac{2}{2}L_{m}, \\ L_{q}=L_{s}+M_{s}-\frac{3}{2}L_{m} \\ -M_{Q}\sin \theta & 0 & 0 & L_{Q} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+M_{s}+\frac{2}{2}L_{m}, \\ L_{q}=L_{s}+M_{s}-\frac{3}{2}L_{m} \\ -M_{Q}\sin \theta & 0 & 0 & L_{Q} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+M_{s}+\frac{2}{2}L_{m}, \\ L_{q}=L_{s}+M_{s}-\frac{3}{2}L_{m} \\ -M_{Q}\sin \theta & 0 & 0 & L_{Q} \end{bmatrix} \frac{d}{dt} \begin{bmatrix} -i_{a}\\ i_{F}\\ i_{D}\\ i_{Q} \end{bmatrix} + \begin{bmatrix} L_{s}+M_{s}+\frac{2}{2}L_{m}, \\ L_{q}=L_{s}+M_{s}-\frac{3}{2}L_{m} \\ L_{$$



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Finding short-circuit and open circuit time constants

Consider the d-axis network

- Short-circuit time constant τ
 - Instantaneous change on ψ_d
 - Delayed change on i_d (through $\frac{1}{1+s\tau}$)
- Open-circuit time constant τ_0
 - Instantaneous change on i_d
 - Delayed change on ψ_d (through $\frac{1}{1+s\tau_0}$)



t=0

• Time constant τ or τ_0 is equal to the total inductance *L* divided by the total resistance *R* of the effective circuit

v(t)

Transient and sub-transient parameters



	d a	xis circuit	q axis	s circuit
Considered rotor windings	Only field Winding	Add the damper winding (ignoring R _{fd})	Only 1 st damper winding	Add the 2 nd damper winding (ignoring <i>R</i> _{1q})
Time constant (open circuit)	$T'_{d0} = \frac{L_{ad} + L_{fd}}{R_{fd}}$ 8.07(s)	$T''_{d0} = \frac{L_{ad} / / L_{fd} + L_{1d}}{R_{1d}}$	$T'_{q0} = \frac{L_{aq} + L_{1q}}{R_{1q}}$	$T''_{q0} = \frac{L_{aq}}{R_{2q}} + \frac{L_{2q}}{R_{2q}}$
Time constant (short circuit)	$T'_{d} = \frac{L_{ad}//L_{l} + L_{fd}}{R_{fd}}$	$T''_{d} = \frac{L_{ad}//L_{fd}//L_{l} + L_{1d}}{R_{1d}}$	$T'_{q} = \frac{L_{aq}//L_{l} + L_{1q}}{P}$	$T''_{q} = \frac{L_{aq}//L_{1q}//L_{l} + L_{2q}}{R}$
Inductance (Reactance) $L_d(s)$ or $L_q(s)$	1.34(s) $L'_{d} = L_{l} + L_{ad} / L_{fd}$	0.023(s) $L''_{d} = L_{l} + L_{ad} / L_{fd} / L_{1d}$	0.37(s) R_{1q} $L'_{q} = L_{l} + L_{aq} / L_{1q}$	0.03(s) $L''_q = L_l + L_{aq} / L_{1q} / L_{2q}$
terminal	0.30(pu)	0.23(pu)	0.65(pu)	0.25(pu)

Based on Parameters in Kundur's Example 3.2

 Note: time constants are all in p.u. To be converted to seconds, they have to be multiplied by t_{base}=1/ω_{base} (i.e. 1/377 for 60Hz).

Table 4.1

Expressions for Standard Parameters of Synchronous Machine

Parameter	Classical Expression	Accurate Expression
T_{d0}^{\prime}		$T_1 + T_2$
T_d'	T_4	$T_4 + T_5$
$T_{d0}^{\prime\prime}$	T ₃	$T_3[T_1/(T_1+T_2)]$
T_d''	T ₆	$T_6[T_4/(T_4+T_5)]$
L_d'	$L_d(T_4/T_1)$	$L_d(T_4 + T_5)/(T_1 + T_2)$
L_d''	$L_d(T_4T_6)/(T_1T_3)$	$L_{d}(T_{4}T_{6})/(T_{1}T_{3})$
with T _ L _{ad} +L	fd Laa	L^+L_{1d}
with $T_1 = \frac{L_{ad} + L_{ad}}{R_{fd}}$	$T_2 = \frac{L_{aa}}{1}$	$\frac{L_{1d}}{R_{1d}}$
with $T_{1} = \frac{L_{ad} + L}{R_{fd}}$ $T_{3} = \frac{1}{R_{1d}} \left(L_{fd} \right)$	$T_{2} = \frac{L_{ad}}{L_{ad}}$ $T_{4} = \frac{1}{R_{fd}}$	$ \frac{L_{1d} + L_{1d}}{\left(L_{fd} + \frac{L_{ad}L_{l}}{L_{ad} + L_{l}}\right)} $

Notes:

- 1. Similar expressions apply to q-axis parameters.
- 2. All parameters are in per unit.
- 3. Time constants in seconds are obtained by dividing the per unit values given in the table by $\omega_0 = 2\pi f$.
- 4. All mutual inductances in *d*-axis assumed equal.

Deriving sub-transient and transient inductances (Classical)

• Subtransient inductance L"_d

$$\psi_d = -L_d i_d + L_{ad} i_F + L_{ad} i_D = -L''_d i_d$$

$$\psi_F = 0 = -L_{ad} i_d + L_F i_F + M_R i_D$$

$$\psi_D = 0 = -L_{ad} i_d + M_R i_F + L_D i_D$$





$$L''_d = L_d - L^2_{ad}/L_D$$

• Transient inductance L'_d : open circuit of the damping winding branch $i_D=0$

$$\psi_d = -L_d i_d + L_{ad} i_F = -L'_d i_d$$
$$\psi_F = 0 = -L_{ad} i_d + L_F i_F$$
$$L'_d = L_d - \frac{L^2_{ad}}{L_F} = L_l + \frac{L_{ad} L_{fd}}{L_{ad} + L_{fd}}$$

Synchronous, Transient and Sub-transient Inductances

$$L_d(s) = L_d \frac{(1 + sT'_d)(1 + sT''_d)}{(1 + sT'_{d0})(1 + sT''_{d0})}$$

- Under steady-state condition: s=0 (t $\rightarrow\infty$) $L_d(0)=L_d$ (d-axis synchronous inductance)
- During a rapid transient: $s \rightarrow \infty$

$$L_{d}'' = L_{d}(\infty) = L_{d} \frac{T_{d}'T_{d}''}{T_{d0}'} = L_{l} + \frac{L_{ad}L_{fd}L_{ld}}{L_{ad}L_{fd} + L_{ad}L_{ld} + L_{fd}L_{ld}}$$
(d-axis sub-transient inductance)

• Without the damper winding: $s >> 1/T'_{d}$ and $1/T'_{d0}$ but $<< 1/T''_{d}$ and $1/T''_{d0}$

$$L'_{d} = L_{d}(\infty) = L_{d} \frac{T'_{d}}{T'_{d0}} = L_{l} + \frac{L_{ad}L_{fd}}{L_{ad} + L_{fd}}$$
 (d-axis transient inductance)



Figure 4.3 Equivalent circuit for incremental values, immediately following a disturbance

Ordering of Time and Inductance Constants

In general, there are:

 $R_{1d} \gg R_{fd}, \quad L_{ad} \gg L_{fd} \text{ and } L_{1d}$ $R_{2q} \gg R_{1q}, \quad L_{aq} \gg L_{1q} \text{ and } L_{2q}$

The following are the parameters in per unit on machine rating of a 555 MVA, 24 kV, 0.9 p.f., 60 Hz, 3600 RPM turbine-generator ¹:

L _{ad} = 1.66	<i>L_{aq}</i> = 1.61	$L_l = 0.15$	R _a =0.003
$L_{fd} = 0.165$	$R_{fd} = 0.0006$	$L_{1d} = 0.1713$	$R_{1d} = 0.0284$
$L_{1q} = 0.7252$	$R_{1q} = 0.00619$	$L_{2q} = 0.125$	R _{2q} = 0.02368
M_R is assumed to	be equal to L_{ad} .		

 $L_{1d} = L_D - L_{ad}$ $L_{fd} = L_F - L_{ad}$ $L_{1q} = L_Q - L_{aq}$ $L_{2q} = L_G - L_{aq}$

$$T'_{d0} = \frac{L_{ad} + L_{fd}}{R_{fd}} > T'_{d} = \frac{L_{ad} / / L_{l} + L_{fd}}{R_{fd}} >> T''_{d0} = \frac{L_{ad} / / L_{fd} + L_{1d}}{R_{1d}} > T''_{d} = \frac{L_{ad} / / L_{fd} / / L_{l} + L_{1d}}{R_{1d}}$$

$$T'_{q0} = \frac{L_{aq} + L_{1q}}{R_{1q}} > T'_{q} = \frac{L_{aq} / / L_{l} + L_{1q}}{R_{1q}} >> T''_{q0} = \frac{L_{ad} / / L_{1q} + L_{2q}}{R_{2q}} > T''_{q} = \frac{L_{ad} / / L_{1q} / L_{l} + L_{2q}}{R_{2q}}$$

$$L_{d} = L_{l} + L_{ad} > L'_{d} = L_{l} + L_{ad} / / L_{fd} = L_{d} - L_{ad}^{2} / L_{F} > L''_{d} = L_{l} + L_{ad} / / L_{1d} (\text{if assuming } M_{R} = L_{ad})$$

$$= L_{d} - L_{ad}^{2} / L_{D} (\text{in some books, if assuming } M_{R} M_{D} = M_{F} L_{D})$$

$$L_{q} = L_{l} + L_{aq} > L'_{q} = L_{l} + \frac{L_{aq}}{L_{1q}} + \frac{L_{aq}}{L_{1q}} + \frac{L_{aq}}{L_{1q}} + \frac{L_{aq}}{L_{1q}} + \frac{L_{aq}}{L_{2q}} = L_{l} + \frac{L_{aq}}{L_{1q}} + \frac{L_{aq}}{L_{2q}} + \frac{L_{aq}}{L_{2$$

	Table 4.	2		

Parameter		Hydraulic Units	Thermal Units		
Synchronous	X _d	0.6 - 1.5	1.0 - 2.3		
Reactance	Xq	0.4 - 1.0	1.0 - 2.3		
Transient	X'd	0.2 - 0.5	0.15 - 0.4		
Reactance	X'_q	dente and the second	0.3 - 1.0		
Subtransient	X''_d	0.15 - 0.35	0.12 - 0.25		
Reactance	X_q''	0.2 - 0.45	0.12 - 0.25		
Transient OC	T'd0	1.5 - 9.0 s	3.0 - 10.0 s		
Time Constant	T_{q0}'	ne statistica part supprise. Presidente a la superiore de la	0.5 - 2.0 s		
Subtransient OC	$T_{d0}^{\prime\prime}$	0.01 - 0.05 s	0.02 - 0.05 s		
Time Constant	$T_{q0}^{\prime\prime}$	0.01 - 0.09 s	0.02 - 0.05 s		
Stator Leakage Inductance	X _l	0.1 - 0.2	0.1 - 0.2		
Stator Resistance	R _a	0.002 - 0.02	0.0015 - 0.005		

1. Reactance values are in per unit with stator base values equal to the 2. Time constants are in seconds.

$$X_d \ge X_q \ge X'_q \ge X'_d \ge X''_q \ge X''_d > X_l$$

$$T'_{d0} > T'_{d} > > T''_{d0} > T''_{d}$$
$$T'_{q0} > T''_{q} > > T''_{q0} > T''_{q}$$

Parameter Estimation by Frequency Response Tests

$$L_d(s) = L_d \frac{(1 + sT'_d)(1 + sT''_d)}{(1 + sT'_{d0})(1 + sT''_{d0})}$$

$$T'_{d0} > T'_{d} > > T''_{d0} > T''_{d}$$

$$1/T'_{d0} < 1/T'_{d} << 1/T''_{d0} < 1/T''_{d0}$$

$$T'_{d0}=5, T'_{d}=1, T''_{d0}=0.05, T''_{d}=0.01 (s)$$

Swing Equations

$$J \frac{d\omega_m}{dt} = T_a = T_m - T_e$$

J

 ω_m

 T_a T_m

 T_e

t

combined moment of inertia of generator
and turbine, $kg \cdot m^2$
angular velocity of the rotor, mech. rad/s
time, s
accelerating torque in N.m
mechanical torque in N.m
electromagnetic torque in N.m

• Define per unit inertia constant

$H = \frac{1}{2} \frac{J \omega_{0m}}{\omega_{0m}} (s)$	Type of generating unit	H (s)
$2 \text{ VA}_{base} $	Thermal unit (a) 3600 r/min (2-pole) (b) 1800 r/min (4-pole)	2.5 to 6.0 4.0 to 10.0
$J = \frac{2\pi}{2} VA_{base}$	Hydraulic unit	2.0 to 4.0

Some references define T_M or M=2H, called the mechanical starting time, i.e. the time required for rated torque to accelerate the rotor from standstill to rated speed

$$J \frac{d\omega_m}{dt} = T_a = T_m - T_e$$
Angular position of the rotor in electrical radian
with respect to a synchronously rotating reference

$$\frac{2H}{\omega_{o_{m}}^2} \nabla A_{bace} \frac{d\omega_m}{dt} = T_m - T_e$$

$$2H \frac{d}{dt} \left(\frac{\omega_m}{\omega_{o_{m}}}\right) = \frac{T_m - T_e}{\nabla A_{bace} / \omega_{o_m}}$$

$$2H \frac{d}{dt} \left(\frac{\omega_m}{\omega_{o_{m}}}\right) = \frac{T_m - T_e}{T_{base}}$$

$$2H \frac{d\overline{\omega}_r}{dt} = \overline{T_m} - \overline{T_e} \quad \text{(in per unit)}$$

$$\frac{d^2\delta}{dt^2} = \frac{d\omega_r}{dt} = \frac{d(\Delta\omega_r)}{dt} \quad \text{in rad/s}^2$$

$$\frac{d^2\delta}{dt^2} = \omega_0 \frac{d(\overline{\omega}_r)}{dt} = \omega_0 \frac{d(\Delta\overline{\omega}_r)}{dt} \quad \text{in rad/s}^2$$
where

$$\omega_r \text{ (in per unit)} = \frac{\omega_m}{\omega_{0m}} = \frac{\omega_r / p_f}{\omega_0 / p_f} = \frac{\omega_r}{\omega_0}$$

$$\frac{2H}{\omega_0} \frac{d^2\delta}{dt^2} = \overline{T_m} - \overline{T_e} \quad \overline{T_e} - K_D \Delta\overline{\omega}_r = \overline{T_m} - \overline{T_e} - \frac{K_D}{\omega_0} \frac{d\delta}{dt}$$

State Space and Block diagram Representations

$$\frac{2H}{\omega_0}\frac{d^2\delta}{dt^2} = \overline{T}_m - \overline{T}_e - K_D \Delta \overline{\omega}_r = \overline{T}_m - \overline{T}_e - \frac{K_D}{\omega_0}\frac{d\delta}{dt}$$

$$2H \frac{d(\Delta \overline{\omega}_r)}{dt} = \overline{T}_m - \overline{T}_e - K_D \Delta \overline{\omega}_r$$
$$\frac{1}{\omega_0} \frac{d\delta}{dt} = \Delta \overline{\omega}_r$$

$$\overline{T}_{e}$$

$$\overline{T}_{m} \xrightarrow{+} \Sigma \xrightarrow{+} \boxed{\frac{1}{2Hs + K_{D}}} \Delta \overline{\omega}_{r} \xrightarrow{-} \frac{\omega_{0}}{s} \xrightarrow{-} \delta$$

Figure 3.34 Block diagram representation of swing equations

State-Space Representation of a Synchronous Machine

So far, we modeled all critical dynamics about a synchronous machine:

- State variables (pX):
 - stator and rotor voltages, currents or flux linkages
 - swing equations (rotor angle and speed)
- Time constants:
 - Inertia: 2H
 - Sub-transient and transient time constants, e.g. T'_{d0} and T''_{d0}
- Other parameters
 - Stator and rotor self- or mutual-inductances and resistances
 - Rotor mechanical torque T_m and stator electromagnetic torque T_e

State Space Model on a Salient-pole Machine

- e_{fd} and T_m are usually known but e_d and e_q are related to its loading conditions (the grid), so algebraic power-flow equations should be introduced.
- The grid model is a set of Differential-Algebraic Equations (DAEs)

Simplified Models

• Salient: $[\psi_d \ \psi_{fd} \ \psi_{1d} \ \psi_q \ \psi_{1q} \ \omega_r \ \delta]^T$ Round: $[\psi_d \ \psi_{fd} \ \psi_{1d} \ \psi_q \ \psi_{1q} \ \psi_{2q} \ \omega_r \ \delta]^T$

Let $p \psi_d = p \psi_q = 0$ and $\omega_r \psi = \psi$

- Salient: $[\psi_{fd} \ \psi_{1d} \ \psi_{1q} \ \omega_r \ \delta]^T$ (2.1 model) Round: $[\psi_{fd} \ \psi_{1d} \ \psi_{1q} \ \psi_{2q} \ \omega_r \ \delta]^T$ (2.2 model)
 - Inertia $2H p\omega_r$
 - Transient $T'_{d0} p \psi_{fd} \quad T'_{q0} p \psi_{1q}$
 - Sub-transient $T''_{d0} p \psi_{1d} T''_{q0} p \psi_{2q}$

Let $p \psi_{1d} = p \psi_{2q} = 0$ (neglect one damper winding in each axis)

- Salient: $[\psi_{fd} \ \omega_r \ \delta]^T$ (**1.0 model** / **1-axis model**) Round: $[\psi_{fd} \ \psi_{1q} \ \omega_r \ \delta]^T$ (**1.1 model** / **2-axis model**)
 - Inertia $2H p\omega_r$
 - Transient $T'_{d0} p \psi_{fd} T'_{q0} p \psi_{1q}$

$$p\Psi = -(\mathbf{R} \times \mathbf{L}^{-1} - \mathbf{\Omega}) \times \Psi + \mathbf{e}$$
$$2H \cdot p\omega_r = T_m - T_e$$
$$p\delta = \omega_r - 1$$

Neglecting $p\psi_d$ and $p\psi_q$

$$e_{d} = -\omega_{r}\psi_{q} - R_{a}i_{d} + p\psi_{d} = -\omega_{r}\psi_{q} - R_{a}i_{d}$$

$$e_{q} = \omega_{r}\psi_{d} - R_{a}i_{q} + p\psi_{q} = \omega_{r}\psi_{d} - R_{a}i_{q}$$

Figure 5.3 Effect of neglecting stator transients on speed deviation

Figure 5.4 Effect of neglecting stator transients on rotor angle swings

Generator models neglecting $p\psi_d$ and $p\psi_a$

Flux equations (AEs):	Voltag
$\psi_d = -L_d i_d + L_{ad} i_{fd} + L_{ad} i_{1d}$	$e_d = -R$
$\psi_q = -L_q i_q + L_{aq} i_{1q} + L_{aq} i_{2q}$	$e_q = -1$
$\psi_{fd} = -L_{ad}i_d + L_F i_{fd} + M_R^d i_{1d}$	$e_{fd} = R$
$\psi_{1d} = -L_{ad}i_d + M_R^d i_{fd} + L_D i_{1d}$	0 = R
$\psi_{1q} = -L_{aq}i_q + L_Qi_{1q} + M_R^q i_{2q}$	0 = K
$\psi_{2q} = -L_{aq}i_q + M_R^q i_{1q} + L_G i_{2q}$	0 = K
where $M_R^d \approx L_D \qquad M_R^q \approx L_G$	

Voltage equations (ODEs): $e_d = -R_a i_d - \omega_r \psi_q + p \psi_d$ $e_q = -R_a i_q + \omega_r \psi_d + p \psi_q$ $e_{fd} = R_{fd} i_{fd} + p \psi_{fd}$ $0 = R_{1d} i_{1d} + p \psi_{1d}$ $0 = R_{1q} i_{1q} + p \psi_{1q}$ $0 = R_{2q} i_{2q} + p \psi_{2q}$

• There are four state variables ψ_{fd} , ψ_{1d} , ψ_{1q} , ψ_{2q} in voltage equations.

- Our goal is to find four ODEs using only X_d, X_q, X'_d, X'_q, X"_d, X"_q, X"_d, X"_q, T'_{d0}, T''_{q0}, T"_{d0}, T"_{q0} as parameters, which can be identified by frequency response tests.
- We prefer more meaningful state variables such as its internal voltages or emfs than flux linkages for a stability study on a power system having many generators.

 $e_d = -R_a i_d + X_q i_q + 0$ $e_q = -R_a i_q - X_d i_d + X_{ad} i_{fd}$

Steady-state voltage equations indicate an internal voltage in the q-axis

Internal-voltage-behind-reactance model appearing under different conditions

 E_0'

Ea

 E_0'' is the predisturbance value of internal voltage given by

$$\tilde{E}_0'' = \tilde{E}_{t0} + jX''\tilde{I}_{t0}$$

(a) Subtransient model

(c) Steady-state model

Voltage equations:

$$e_{d} = -R_{a}i_{d} - \omega_{r}\psi_{q}$$

$$e_{q} = -R_{a}i_{q} + \omega_{r}\psi_{d}$$

$$e_{fd} = R_{fd}i_{fd} + p\psi_{fd}$$

$$0 = R_{1d}i_{1d} + p\psi_{1d}$$

$$0 = R_{1q}i_{1q} + p\psi_{1q}$$

$$0 = R_{2q}i_{2q} + p\psi_{2q}$$

Flux equations:

$$\begin{split} \psi_{d} &= -L_{d}i_{d} + L_{ad}i_{fd} + L_{ad}i_{1d} \\ \psi_{q} &= -L_{q}i_{q} + L_{aq}i_{1q} + L_{aq}i_{2q} \\ \psi_{fd} &= -L_{ad}i_{d} + L_{F}i_{fd} + M_{R}^{d}i_{1d} \\ \psi_{1d} &= -L_{ad}i_{d} + M_{R}^{d}i_{fd} + L_{D}i_{1d} \\ \psi_{1q} &= -L_{aq}i_{q} + L_{Q}i_{1q} + M_{R}^{q}i_{2q} \\ \psi_{2q} &= -L_{aq}i_{q} + M_{R}^{q}i_{1q} + L_{G}i_{2q} \end{split}$$

$$M_R^d \approx L_D \qquad M_R^q \approx L_C$$

D. W. Olive, Digital Simulation of Synchronous Machine Transients, IEEE Trans. Power Apparatus and Systems, Vol. PAS-87, No. 8, Aug. 1968

Steady-state: $e_{d} = -R_{a}i_{d} - \omega_{r}\psi_{a} = -R_{a}i_{d} + X_{a}i_{a} - \omega_{r}L_{aa}(i_{1a} + i_{2a})$ $e_d = -R_a i_d + X_q i_q + 0$ $e_{a} = -R_{a}i_{a} + \omega_{r}\psi_{d} = -R_{a}i_{a} - X_{d}i_{d} + \omega_{r}L_{ad}(i_{fd} + i_{1d})$ $e_a = -R_a i_a - X_d i_d + X_{ad} i_{fd}$ • Find internal voltage $E'' = E''_d + jE''_q$ under sub-transient conditions: $e_d = -R_a i_d + X_a'' i_a + E_d''$ Assume: $E_d'' = k_{2q} \psi_{2q}$ $e_{q} = R_{a}i_{q} - X_{d}''i_{d} + E_{a}''$ $E_{a}'' = k_{1d}\psi_{1d}$ $e_{d} + R_{a}i_{d} = X_{a}''i_{a} + E_{d}'' = X_{q}''i_{q} + k_{2q}(-L_{aq}i_{q} + M_{R}^{q}i_{1q} + L_{G}i_{2q}) = X_{q}i_{q} - \omega_{r}L_{aq}(i_{1q} + i_{2q})$ $e_{a} + R_{a}i_{a} = -X_{d}''i_{d} + E_{a}'' = -X_{d}''i_{d} + k_{1d}(-L_{ad}i_{d} + M_{R}^{d}i_{fd} + L_{D}i_{1d}) = -X_{d}i_{d} + \omega_{r}L_{ad}(i_{fd} + i_{1d})$ $E_{d}'' = -\omega_{r} \frac{L_{ad}}{L_{G}} \psi_{2q} \qquad \psi_{q}'' = -\frac{E_{d}''}{\omega_{r}} = \frac{L_{aq}}{L_{G}} \psi_{2q}$ $E_{q}'' = \omega_{r} \frac{L_{ad}}{L_{D}} \psi_{1d} \qquad \psi_{d}'' = \frac{E_{q}''}{\omega_{r}} = \frac{L_{ad}}{L_{D}} \psi_{1d}$ $\psi_{q} = \psi_{q}'' - L_{q}''i_{q}$ $\psi_d = \psi_d'' - L_d'' i_d$

Voltage equations:

 $e_{d} = -R_{a}i_{d} - \omega_{r}\psi_{q}$ $e_{q} = -R_{a}i_{q} + \omega_{r}\psi_{d}$ $e_{fd} = R_{fd}i_{fd} + p\psi_{fd}$ $0 = R_{1d}i_{1d} + p\psi_{1d}$ $0 = R_{1q}i_{1q} + p\psi_{1q}$ $0 = R_{2q}i_{2q} + p\psi_{2q}$

Flux equations:

$$\begin{split} \psi_{d} &= -L_{d}i_{d} + L_{ad}i_{fd} + L_{ad}i_{1d} \\ \psi_{q} &= -L_{q}i_{q} + L_{aq}i_{1q} + L_{aq}i_{2q} \\ \psi_{fd} &= -L_{ad}i_{d} + L_{F}i_{fd} + M_{R}^{d}i_{1d} \\ \psi_{1d} &= -L_{ad}i_{d} + M_{R}^{d}i_{fd} + L_{D}i_{1d} \\ \psi_{1q} &= -L_{aq}i_{q} + L_{Q}i_{1q} + M_{R}^{q}i_{2q} \\ \psi_{2q} &= -L_{aq}i_{q} + M_{R}^{q}i_{1q} + L_{G}i_{2q} \end{split}$$

$$M_R^d \approx L_D \qquad M_R^q \approx L_G$$

D. W. Olive, Digital Simulation of Synchronous Machine Transients, IEEE Trans. Power Apparatus and Systems, Vol. PAS-87, No. 8, Aug. 1968

$$e_{d} = -R_{a}i_{d} - \omega_{r}\psi_{q} = -R_{a}i_{d} + X_{q}i_{q} - \omega_{r}L_{aq}(i_{1q} + i_{2q})$$

$$e_{q} = -R_{a}i_{q} + \omega_{r}\psi_{d} = -R_{a}i_{q} - X_{d}i_{d} + \omega_{r}L_{ad}(i_{fd} + i_{1d})$$
• Find internal voltage $E' = E'_{d} + jE'_{q}$ under transient conditions $(i_{1d}=0, i_{2q}=0)$:

$$e_{d} = -R_{a}i_{d} + X'_{q}i_{q} + E'_{d}$$

$$e_{q} = R_{a}i_{q} - X'_{d}i_{d} + E'_{q}$$
Assume: $E'_{d} = k_{1q}\psi_{1q}$

$$E'_{q} = k_{d}i_{q} - X'_{d}i_{d} + E'_{q}$$

$$e_{d} + R_{a}i_{q} = -X'_{d}i_{d} + k_{fd}(-L_{aq}i_{q} + L_{Q}i_{1q} + 0) = X_{d}i_{q} - \omega_{r}L_{aq}(i_{1q} + 0)$$

$$e_{q} + R_{a}i_{q} = -X'_{d}i_{d} + k_{fd}(-L_{ad}i_{d} + L_{F}i_{fd} + 0) = -X_{d}i_{d} + \omega_{r}L_{ad}(i_{fd} + 0)$$

$$E'_{d} = -\omega_{r}\frac{L_{aq}}{L_{Q}}\psi_{1q}$$

$$\psi'_{q} = -\frac{E'_{d}}{\omega_{r}} = \frac{L_{aq}}{L_{Q}}\psi_{1q}$$

$$\psi'_{d} = \frac{E'_{q}}{\omega_{r}} = \frac{L_{ad}}{L_{F}}\psi_{fd}$$

$$\psi_{d} = \psi'_{d} - L'_{d}i_{d}$$

$$\psi_{q} = \psi'_{q} - L'_{d}i_{d}$$
Machine stems,

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 Ψ_{1q}

 Ψ_{2q}

Finding a detailed generator model

Voltage equations: $e_{d} = -R_{a}i_{d} - \omega_{r}\psi_{q}$ $e_{q} = -R_{a}i_{q} + \omega_{r}\psi_{d}$ $e_{fd} = R_{fd}i_{fd} + p\psi_{fd}$ $0 = R_{1d}i_{1d} + p\psi_{1d}$ $0 = R_{1q}i_{1q} + p\psi_{1q}$ $0 = R_{2q}i_{2q} + p\psi_{2q}$

Flux equations: $\psi_d = -L_d i_d + L_{ad} i_{fd} + L_{ad} i_{1d}$ $\psi_q = -L_q i_q + L_{aq} i_{1q} + L_{aq} i_{2q}$ $\psi_{fd} = -L_{ad} i_d + L_F i_{fd} + M_R^d i_{1d}$ $\psi_{1d} = -L_{ad} i_d + M_R^d i_{fd} + L_D i_{1d}$ $\psi_{1q} = -L_{aq} i_q + L_Q i_{1q} + M_R^q i_{2q}$ $\psi_{2q} = -L_{aq} i_q + M_R^q i_{1q} + L_G i_{2q}$

 $M_R^d \approx L_D \qquad M_R^q \approx L_G$ © 2021 Kai Sun

$E'_{d} = -\omega_r \frac{L_{aq}}{L_Q} \psi_{1q}$	$T_{d0}' = \frac{L_F}{R_{fd}}$	X _d	$-X'_{d} = \omega_{r} \frac{L_{ad}^{2}}{L_{F}}$	$i_d = \frac{E'_q}{X'_d} - E'_q$	$\frac{-E''_q}{-X''_d}$	
$E_q' = \omega_r \frac{L_{ad}}{L_F} \psi_{fd}$	$T_{q0}^{\prime}=\frac{L_Q}{R_{1q}}$	X_q	$-X'_{q} = \omega_{r} \frac{L_{aq}^{2}}{L_{Q}}$	$i_q = \frac{E'_d}{X'_q} - E'_d$	$\frac{-E''_d}{-X''_q}$	
$E_d'' = -\omega_r \frac{L_{ad}}{L_G} \psi_{2q}$	$T_{d0}^{\prime\prime} = \frac{L_D^2 - \frac{L_{ad}^2}{L_F}}{D}$	X _d	$-X_d'' = \omega_r \frac{L_{ad}^2}{L_D}$			
$E_q'' = \omega_r \frac{L_{ad}}{L_D} \psi_{1d}$	R_{1d} $L_G - \frac{L_{aq}^2}{L_G}$	X_q	$-X_{q}'' = \omega_{r} \frac{L_{aq}^{2}}{L_{G}}$			
$E_{fd} \stackrel{\text{def}}{=} \frac{\omega_r L_{ad}}{R_{fd}} e_{fd}$	$T_{q0}^{\prime\prime} = \frac{-\underline{Q}}{R_{2q}}$		$T_{d0}^{\prime} \frac{d}{dt} E_{q}^{\prime} = -\frac{X_{d}}{X_{d}^{\prime}}$	$\frac{-X_d''}{-X_d''} \cdot E_q' +$	$+\frac{X_d - X'_d}{X'_d - X''_d}$	$E_q'' + E_{fd}$
Eliminate all but E	X, X, T, i_d, i_q		$T_{q0}^{\prime}\frac{d}{dt}E_{d}^{\prime} = -\frac{X_{q}}{X_{q}^{\prime}}$	$\frac{-X''_q}{-X''_q} \cdot E'_d +$	$-\frac{X_q-X_q'}{X_q'-X_q''}$	E_d''
(modeled as a cur) having internal vol	r ent source tages behind a		$T_{d0}'' \frac{d}{dt} E_q'' = E_q' - L$	$E_q'' - (X_d' -$	X_d'') $\cdot i_d$	
reactan	ce)	7	$T_{q0}''\frac{d}{dt}E_d''=E_d'-1$	$E_d'' + (X_q' -$	$(X_q'') \cdot i_q$	
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6th order (2.2) Detailed Model

$$\begin{cases} \frac{d\delta}{dt} = \Delta \omega_r \\ \frac{2H}{\omega_0} \frac{d\Delta \omega_r}{dt} = P_m - P_e - K_D \frac{\Delta \omega_r}{\omega_0} \\ T'_{d0} \frac{d}{dt} E'_q = -\frac{X_d - X''_d}{X'_d - X''_d} \cdot E'_q + \frac{X_d - X'_d}{X'_d - X''_d} \cdot E''_q + E_{fd} \\ T'_{q0} \frac{d}{dt} E'_d = -\frac{X_q - X''_q}{X'_q - X''_q} \cdot E'_d + \frac{X_q - X'_q}{X'_q - X''_q} \cdot E''_d \\ T''_{d0} \frac{d}{dt} E''_q = E'_q - E''_q - (X'_d - X''_d) \cdot i_d \\ T''_{q0} \frac{d}{dt} E''_d = E'_d - E''_d + (X'_q - X''_q) \cdot i_q \\ \tilde{E}_t = e_d + je_q = e^{-j(\delta - \pi/2)} \tilde{E}_g \\ \tilde{I}_t = i_d + ji_q = e^{-j(\delta - \pi/2)} \tilde{I}_g \\ e_d = -R_a i_d + X''_q i_q + E''_q \\ e_q = -R_a i_q - X''_d i_d + E''_q \\ P_e = e_d i_d + e_q i_q \end{cases}$$

$$\delta$$
 – rad
 ω_r – rad/s
 H, T', T'' – sec
 P, E, V, I, i, X, L, R – pu

 E_g and I_g are respectively E_t and I_t seen in the common reference of the system

4th order (1.1) Two-Axis Model

• Ignore sub-transient dynamics $(\text{let } i_{1d} = i_{2d} = 0 \text{ or } T''_{d0} = T''_{a0} = 0)$ $\int \frac{d\delta}{dt} = \Delta \omega_r$ $\frac{2H}{\omega_0} \frac{d\Delta\omega_r}{dt} = P_m - P_e - K_D \frac{\Delta\omega_r}{\omega_0}$ $T'_{d0} \frac{d}{dt} E'_{q} = -E'_{q} - (X_{d} - X'_{d}) \cdot i_{d} + E_{fd}$ $T'_{q0}\frac{d}{dt}E'_{d} = -E'_{d} + (X_q - X'_q) \cdot i_q$ $\tilde{E}_t = e_d + je_q = e^{-j(\delta - \pi/2)}\tilde{E}_g$ $\tilde{I}_{t} = i_{d} + ji_{a} = e^{-j(\delta - \pi/2)}\tilde{I}_{g}$ $e_d = -R_a i_d + X'_a i_q + E'_d$ $e_a = -R_a i_a - X'_d i_d + E'_a$ $P_e = e_d i_d + e_a i_a$

If T_{q0} '=0, a 3rd order (1.0) 1-axis model is obtained.

2nd order (0.0) Classical Model

- Only keep swing equations as differential equations
- Assume $X'_d = X'_q$

$$\begin{cases} \frac{d\delta}{dt} = \Delta \omega_r \\ \frac{2H}{\omega_0} \frac{d\Delta \omega_r}{dt} = P_m - P_e - K_D \frac{\Delta \omega_r}{\omega_0} \\ \tilde{E}_g = e^{j(\delta - \pi/2)} \tilde{E}_t \\ \tilde{I}_g = e^{j(\delta - \pi/2)} \tilde{I}_t \\ \tilde{E}_t = \tilde{E}' - (R_a + jX'_d) \tilde{I}_t \\ P_e = real(\tilde{E}_t \tilde{I}_t^*) \end{cases}$$

Figure 5.7 Simplified transient model

E' is constant and can be estimated by computing its pre-disturbance value

$$E' = |\tilde{E}_{t0} + (R_a + jX'_d)\tilde{I}_{t0}|$$

Phasor Diagram

$$\tilde{E}_{g} = \tilde{E}_{t} \times 1 \angle (\delta - 90^{\circ})$$
$$\tilde{I}_{g} = \tilde{I}_{t} \times 1 \angle (\delta - 90^{\circ})$$

Terminal voltage and current seen from the grid side

Steady state

$$e_{d} = -R_{a}i_{d} + X_{q}i_{q}$$

$$e_{q} = -R_{a}i_{q} - X_{d}i_{d} + X_{ad}i_{fd}$$

$$\tilde{E}_{t} = \tilde{E}_{q} - (R_{a} + jX_{q})\tilde{I}_{t}$$
where $\tilde{E}_{q} = jX_{ad}i_{fd}$ - $j(X_{d}-X_{q})i_{d} = jX_{ad}i_{fd}$ (if $X_{d}=X_{q}=X_{s}$)
Transient dynamic

$$e_{d} = -R_{a}i_{d} + X'_{q}i_{q} + E'_{d}$$

$$e_{q} = -R_{a}i_{q} - X'_{d}i_{d} + E'_{q}$$
If $X'_{d} = X'_{q}$

$$\tilde{E}_{t} = \tilde{E}' - (R_{a} + jX'_{d})\tilde{I}_{t}$$
Sub-transient dynamic

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Comparison of PSS/E Generator Models

Reactance and Time Constants Used	Model							
	GENSAL and GENSAE	GENROU and GENROE	GENDCO	GENTRA	GENCLS			
Xd	~	~	~	~	~			
Xq	~	~	~	~				
X'd	~	~	~	~	~			
X'q		~	~					
X″d	~	~	~					
X″q	*	*	*					
XI	~	~	~	~				
T'do	~	~	~	~				
T'qo		~	~					
T″ _{do}	~	~	~	~				
T″ _{qo}	~	~	~	~				
Saturation Factors	~	~	~	~				

Table 1: Summary of Generator Models in Terms of Data Used

 $*X''_{q}$ is assumed to be equal to X''_{d} .

2.1 Machine: GENSAL model with saturation.

2.3 Machine: GENROU model with saturation

The GENROU model including saturation is presented below. The Mechanical Swing Equations, Electric Torque, and Network Interface Equations are all the same as the GENROU and GENSAL models previously.

