Abstract: Many definitions have been formulated to characterize, detect, and measure active and non-active current and power for non-sinusoidal and non-periodic waveforms in electric systems. This paper presents definitions and compensation of non-active current from the compensation standpoint.

I. INTRODUCTION

Because of the widespread use of nonlinear loads and electronic power converters, non-sinusoidal and non-periodic loads and voltage distortion are becoming more common in today’s electrical systems. Many papers have dealt with the definition, identification, characterization, detection, measurement, and compensation of such non-sinusoidal and non-periodic current and power [1-11]. Tolbert (co-author of this paper) and Habetler have compiled a comprehensive technical survey of the published literature on the topic [12].

Instantaneous active power is defined as the time rate of energy generation, transfer, or utilization. It is a physical quantity and satisfies the principle of conservation of energy. For a single-phase circuit, it is defined as the instantaneous product of voltage and current:

\[ p(t) = v(t)i(t). \]  

(1)

Active power \( P \) is the time average of the instantaneous power over one period of the wave \( p(t) \). For a polyphase circuit with \( M \) phases, each phase’s instantaneous active power is still expressed as (1) and instantaneous total active power is the sum of the active powers of the individual phases:

\[ p(t) = \sum_{j=1}^{M} p_j(t) = \sum_{j=1}^{M} v_j(t)i_j(t) . \]  

(2)

Non-active power can be thought of as the useless power that causes increased line current and losses, greater generation requirements for utilities, and other effects/burdens to power systems and connected/related equipment.

II. COMPENSATION SYSTEM AND DEFINITION

A. Compensation Systems

For a single or polyphase power system, a shunt compensator to minimize the useless power/current can be configured as in Fig. 1. Assuming that the shunt compensator only consists of passive (inductor and/or capacitor) and/or switching devices without any external power source and neglecting the compensator’s power loss, then the active power of the com-
pensator should average zero according to the principle of conservation of energy. That is,

\[ P_S(t) = P_L(t), \quad P_C(t) = 0, \text{ for } t \to \infty. \]  \hfill (3)

\[ P_X(t) = \frac{1}{T_C} \int_{t-T_C}^{t} P_X(\tau) d\tau \quad \text{and} \quad X = S, L, \text{ or } C. \]  \hfill (4)

In (4), \( T_C \) is the averaging interval that can be zero, one fundamental cycle, one-half cycle, or multiple cycles, depending on compensation objectives and the passive components' energy storage capacity. Equations (3) and (4) must hold true regardless of single-phase or polyphase, passive compensation or active compensation. Based on these physical and practical limitations, non-active power and non-active current can be defined and formulated.

B. Definitions of Non-Active Power and Current

We extend Fryze's idea of non-active current/power [1] as follows:

\[ i_p(t) = \frac{P(t)}{V_p(t)} \cdot v_p(t), \quad i_q(t) = i(t) - i_p(t), \]  \hfill (5)

where

\[ V_p(t) = \sqrt{\frac{1}{T_C} \int_{t-T_C}^{t} v_p(\tau)^2 d\tau}. \]  \hfill (6)

\( i_p(t) \) is the active current and \( i_q(t) \) is the non-active current. \( P(t) \) is the average active power over the interval \([t-T_C, t]\), which can be calculated from (4). \( V_p(t) \) is the rms value of the voltage, \( v_p(t) \) over the interval \([t-T_C, t]\), which is expressed by (6). \( v_p(t) \) is the reference voltage that can be the voltage itself, \( v(t) \) or the fundamental component of \( v(t) \), where \( v(t) = v(t)+v_h(t) \) and \( v_p(t) = v(t) \), or something else, depending on compensation objectives. These definitions, (5) and (6) are valid for single- and polyphase circuits. However, for polyphase circuits, voltages and currents are expressed as a vector, e.g., for a three-phase system, \( v = [v_a, v_b, v_c]^T, \quad i = [i_a, i_b, i_c]^T \), and \( v^2 = [v_a^2, v_b^2, v_c^2] = [v_a, v_b, v_c]^T \).

III. DISCUSSION, DEDUCTION, AND COMPENSATION

Equation (5) provides the basic definitions of active and non-active current, from which most of the existing non-active power theories and definitions based on time-domain can be extended and deduced. The following discusses some deductions and compensation examples.

A. Sinusoidal Single-Phase Circuits

For a single-phase circuit with sinusoidal waveforms, e.g., \( v_S = V_S \sin(\omega t) \) and \( i_S = I_S \sin(\omega t + \alpha) \), the active and non-active currents are consistent with the traditional active and reactive powers and can be derived from (4), (5) and (6) by the following steps: (i) choose \( T_C \) to be one or half fundamental cycle, \( T_C = 2\pi/\omega \) or \( T_C = \pi/\omega \); (ii) calculate the average active power, \( P_L \) according to (4); (iii) calculate the rms value of the voltage \( v_S \), according to (6); (iv) calculate the active current and non-active current, \( i_{lp} \) and \( i_{lq} \) according to (5). The result is

\[ i_{lp}(t) = I_L \cos \alpha \sin(\omega t), \quad i_{lq}(t) = -I_L \sin \alpha \cos(\omega t). \]  \hfill (7)

Equation (7) clearly shows the consistence with the traditional reactive power theory. An algorithm for active compensators can be easily implemented because the definitions as formulated in (4), (5) and (6) are all real-time based. In addition, it is easy to show that the compensator needs zero average power when it injects the non-active current because \( P_C = 0 \) when \( i_C = i_{lq} \). After compensation, the source current, \( i_S \) will only contain the load active current, \( i_{lp} \).

B. Non-Sinusoidal Single-Phase Circuits

For a single-phase circuit with non-sinusoidal waveforms, the active and non-active currents, which can be derived from (4), (5) and (6) using the steps described in section III.A, are consistent with the traditional Fryze's active and non-reactive currents when choosing \( v_p = v_S \) in (5) and (6). For example, if \( v_S = v(t)+v_h(t) = V_S \sin(\omega t) + V_h \sin(\omega t + \beta_h) \) and \( i_S = i_S \sin(\omega t + \alpha) + I_{lh} \sin(\omega t + \beta_h) \), then

\[ i_{lp}(t) = V_S^2 I_{lf} \cos \alpha + V_{lh} I_{lh} \cos \alpha, \quad \frac{V_S^2 + V_{lh}}{V_S^2} \]  \hfill (8)

\[ i_{lq}(t) = -I_{lf} \sin \alpha \cos(\omega t) - I_{lh} \sin \alpha \cos(\omega t + \beta_h) \]

\[ + \frac{V_S^2 I_{lf} \cos \alpha - V_{lh} I_{lh} \cos \alpha \cdot \sin(\omega t)}{V_S^2 + V_{lh}} \]

\[ + \frac{V_S^2 I_{lh} \cos \alpha - V_{lh} I_{lh} \cos \alpha \cdot \sin(\omega t + \beta_h)}{V_S^2 + V_{lh}} \]  \hfill (9)

Again, the average power of \( i_{lq} \) is zero, which satisfies the requirements in (3) for a compensator. However, it is observed from (8) that the active current contains harmonics because of
the voltage distortion, which means that the source current will not become sinusoidal after the compensator injects (compensates) the non-active current expressed in (9). In most cases, it is desirable that compensation of non-active current results in a pure sine wave source current. In order to achieve that, one should choose \( v_p = v_{sf} \) in (5) and (6). By doing so, one has

\[
i_{lp}(t) = I_{lf} \cos \alpha + \frac{V_{sa} I_{lf} \cos \alpha_h}{V_{sf}} \sin(\alpha t), \tag{10}
\]

\[
i_{ls}(t) = -I_{lf} \sin \alpha \cos(\alpha t) - \frac{V_{sa} I_{lf} \cos \alpha_h}{V_{sf}} \cdot \sin(\alpha t) + I_{lh} \sin(\alpha_h t + \beta_h - \alpha_h). \tag{11}
\]

Equation (10) shows that the active current is a sine wave, and (11) shows that the non-active current contains all harmonic current and fundamental reactive current. After compensation, the source current will become sinusoidal and active. In addition, it is noticeable that the active and non-active currents expressed in (10) and (11) still meet the compensation energy conservation requirements in (3). This is a very important property of the definitions given in (4), (5), and (6), which is also necessary in order to implement compensation in Fig. 1.

C. Single-Phase Circuits with Non-Sinusoidal and Non-Periodic Current

It is more convenient using simulations to study a load current with non-sinusoidal and non-periodic waveform. Fig. 2 shows a case study, where the load generates a non-periodic pulse current. The calculation is based on (4), (5), and (6) with \( T_C = \) one-half fundamental cycle. The simulation results clearly demonstrate the following points: (1) the active load current, \( I_{lp} \), is sinusoidal and in phase with the voltage although the load current, \( I_L \), is highly distorted and non-periodic; (2) the calculated load non-active current, \( I_{Lq} \), is highly distorted and out of phase with the source voltage; (3) the average load power, \( P_{Lq} \), generated from the load non-active current is zero. Therefore, when a compensator is used to compensate for the load non-active current it will consume average zero power and maintain the requirement in (3). In addition, the source current will become sinusoidal and in phase with the voltage.

D. Three-Phase Circuits

The definitions described in (5) and (6) are valid for a three-phase system as well regardless of whether the voltage and current waveforms are sinusoidal or non-sinusoidal, periodic or non-periodic, and balanced or unbalanced. Results are similar to those in the previous subsections. For polyphase (\( M \)-phase) systems, there is one interesting concept, instantaneous reactive (or non-active) power and current, which do not exist in single phase situations. This instantaneous non-active power theory \([2, 3, 10]\) can be deduced from (5) and (6) as well.

Fig. 2. Simulation results for non-periodic current compensation.
In (4) and (6), choosing \( T_C \to 0 \) yields the instantaneous non-active current (power) theory. That is,
\[
i_p(t) = \frac{p(t)}{V_p^2(t)} v_p(t), \quad i_p(t) = i(t) - i_p(t),
\]
(12)
where
\[
v_p(t) = \begin{bmatrix} v_{1p} & v_{2p} & \cdots & v_{Mp} \end{bmatrix}^T, \quad i(t) = \begin{bmatrix} i_1 & i_2 & \cdots & i_M \end{bmatrix}^T, \quad \text{and}
\]
(13)
\[
V_p^2(t) = v_p^2(t) = v_{1p}^2(t) + v_{2p}^2(t) + \cdots + v_{Mp}^2(t).
\]
(14)
Again, the definitions given in (4), (5) and (6) have great flexibility to meet all compensation objectives. Regardless of whether the voltage and current are sinusoidal or non-sinusoidal, balanced or unbalanced, and periodic or non-periodic, the definitions give means to calculate any non-active current component that requires compensation. Consider a three phase four wire system (Fig. 3):
\[
v_S = [v_{Sa}, v_{Sb}, v_{Sc}]^T, \quad i_L = [i_{La}, i_{Lb}, i_{Lc}]^T, \quad \text{and} \quad i_C = [i_{Ca}, i_{Cb}, i_{Cc}]^T.
\]
If the compensation objective is to make the source current sinusoidal and balanced, one can calculate compensation current as follows:

- Separate the voltage into four components: fundamental positive-sequence, \( v_{Sp} \); fundamental negative-sequence, \( v_{Sn} \); zero sequence, \( v_{So} \); and harmonic component, \( v_{Sh} \), i.e.,
\[
v_S = v_{Sp} + v_{Sn} + v_{So} + v_{Sh}, \quad \text{and} \quad v_{So} = \frac{1}{3} (v_{Sa} + v_{Sb} + v_{Sc}) \cdot [1, 1, 1]^T.
\]
(15)
- Choose \( T_C = \text{one-half or one fundamental cycle} \) and \( v_p = v_{Sp} \) in (4), (5) and (6), i.e.,
\[
T_C = T/2, \quad \text{or} \quad T_C = T \quad \text{and} \quad v_p = v_{Sp}.
\]
(16)
- Calculate the load non-active current, \( i_{Lq} \), as (5) and let \( i_C = i_{Lq} \).

From (4), (5), (6), (15), and (16), it is easy to show that the compensator consumes zero average power, \( P_c = 0 \), and satisfies the requirements in (3). Fig. 4 shows simulation results of a case in which both the source voltage and load current are distorted and unbalanced and contain zero-sequence. The results clearly show that: (1) the source current, Isa, Isb, and Isc, becomes sinusoidal; (2) the source current becomes balanced as indicated by zero source neutral current, Isn; and (3) the average power of the compensator, \( P_c \), equals zero.

Fig. 3. A compensation system for three-phase four-wire system.

Fig. 4. Simulation results for a distorted and unbalanced three-phase four-wire system.
IV. CONCLUSIONS

In this paper, definitions of active and non-active power and current have been given from the compensation standpoint. Their definitions are consistent with the traditional reactive and non-active concept for single phase circuits. In addition, the instantaneous reactive power theories can be deduced from the proposed definitions for polyphase systems. The definitions also have the flexibility that any compensation objective can be achieved by choosing an appropriate averaging interval \( (T_C) \) and reference voltage.

REFERENCES


V. BIOGRAPHIES

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