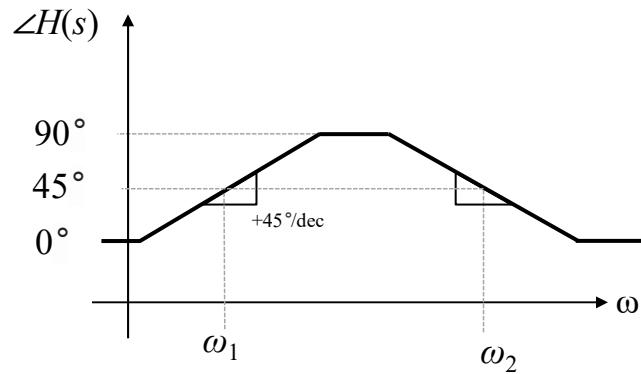
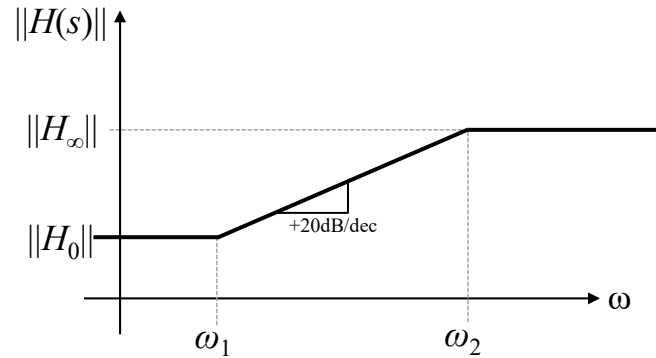
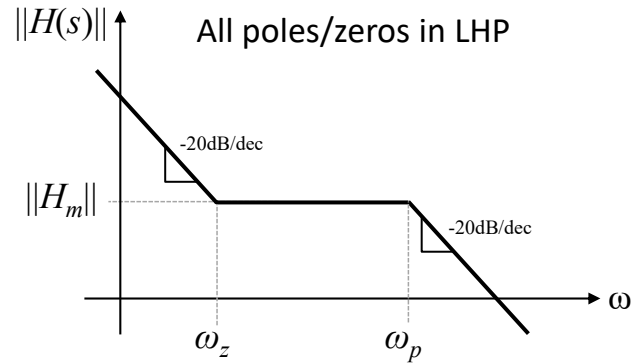


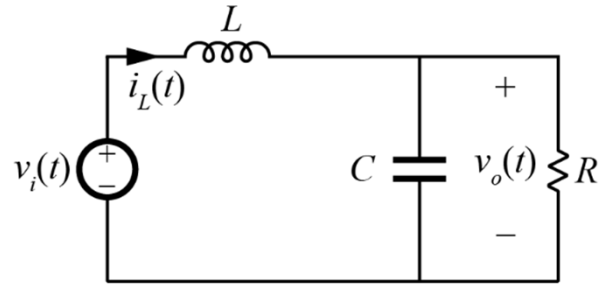
Example 2



Example 3



Complex Poles



Standard Form for Complex Poles

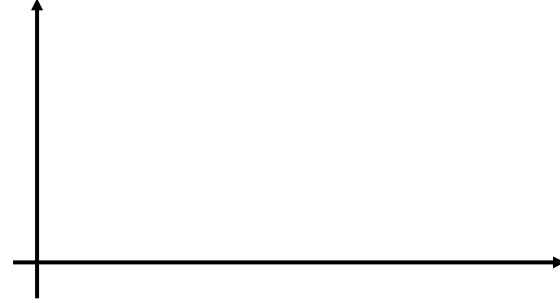
$$H(s) = \frac{1}{\left(\frac{s}{\omega_0}\right)^2 + \frac{s}{Q\omega_0} + 1}$$

$$H(s) = \frac{1}{\left(\frac{s}{\omega_0}\right)^2 + 2\zeta\frac{s}{\omega_0} + 1}$$

- ω_0 , Q , and ζ are real for real circuits
 - Found by coefficient matching to solved transfer function
- ω_0 is the angular/resonant frequency
 - Q is the “quality factor”
 - Roots are complex when $Q > \frac{1}{2}$
 - If $Q < \frac{1}{2}$, the polynomial can be factored into two real poles
 - ζ is the “damping factor”
 - Roots are complex when $\zeta > 1$
 - If $\zeta > 1$, the polynomial can be factored into two real poles

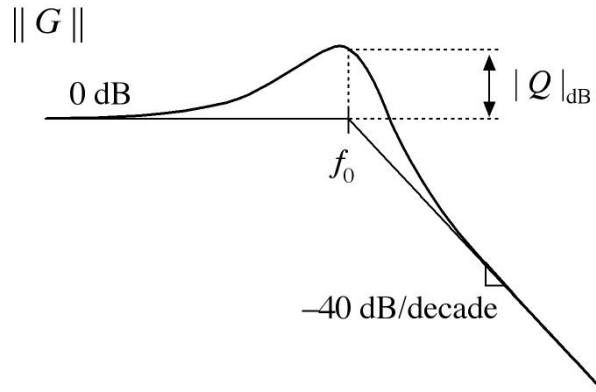
$$\zeta = \frac{1}{2Q}, \quad Q = \frac{1}{2\zeta}$$

Magnitude Asymptotes

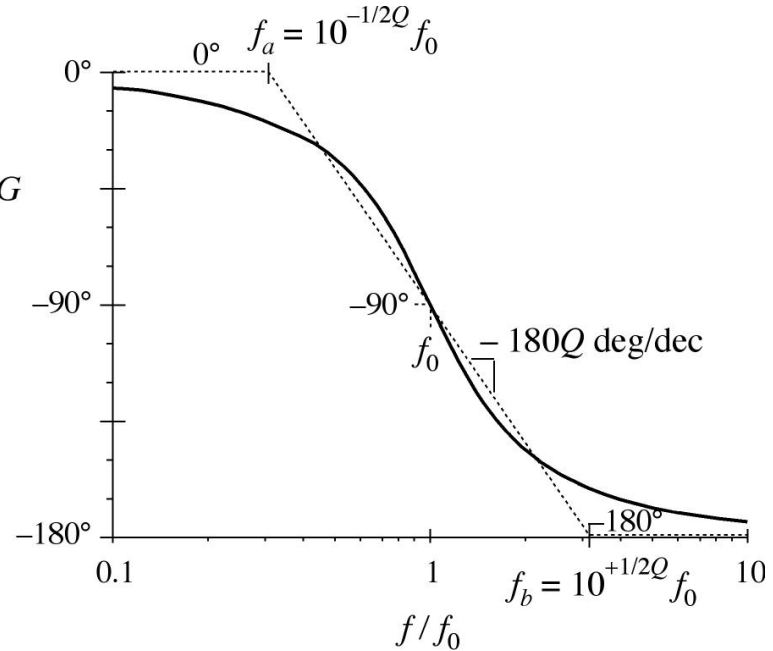


Asymptotes for Complex Poles, $Q > 0.5$

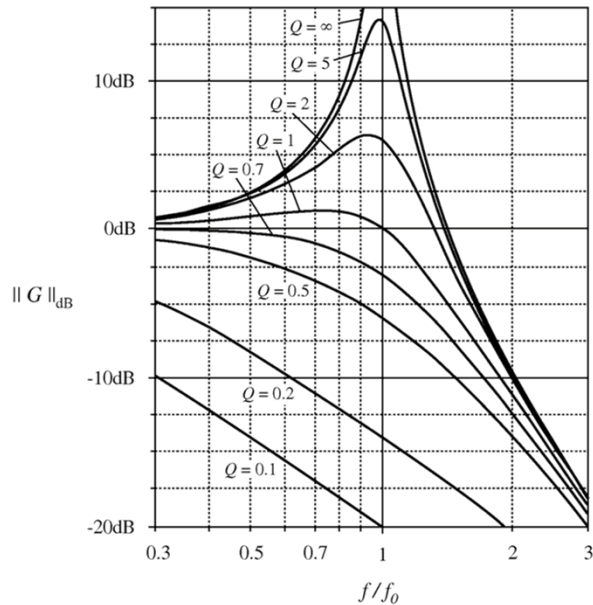
Magnitude



Phase

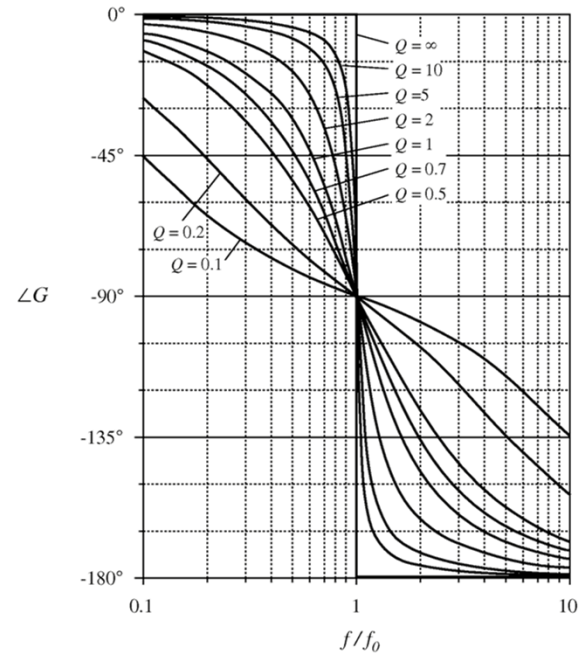


Curves for Varying Q



Fundamentals of Power Electronics

44

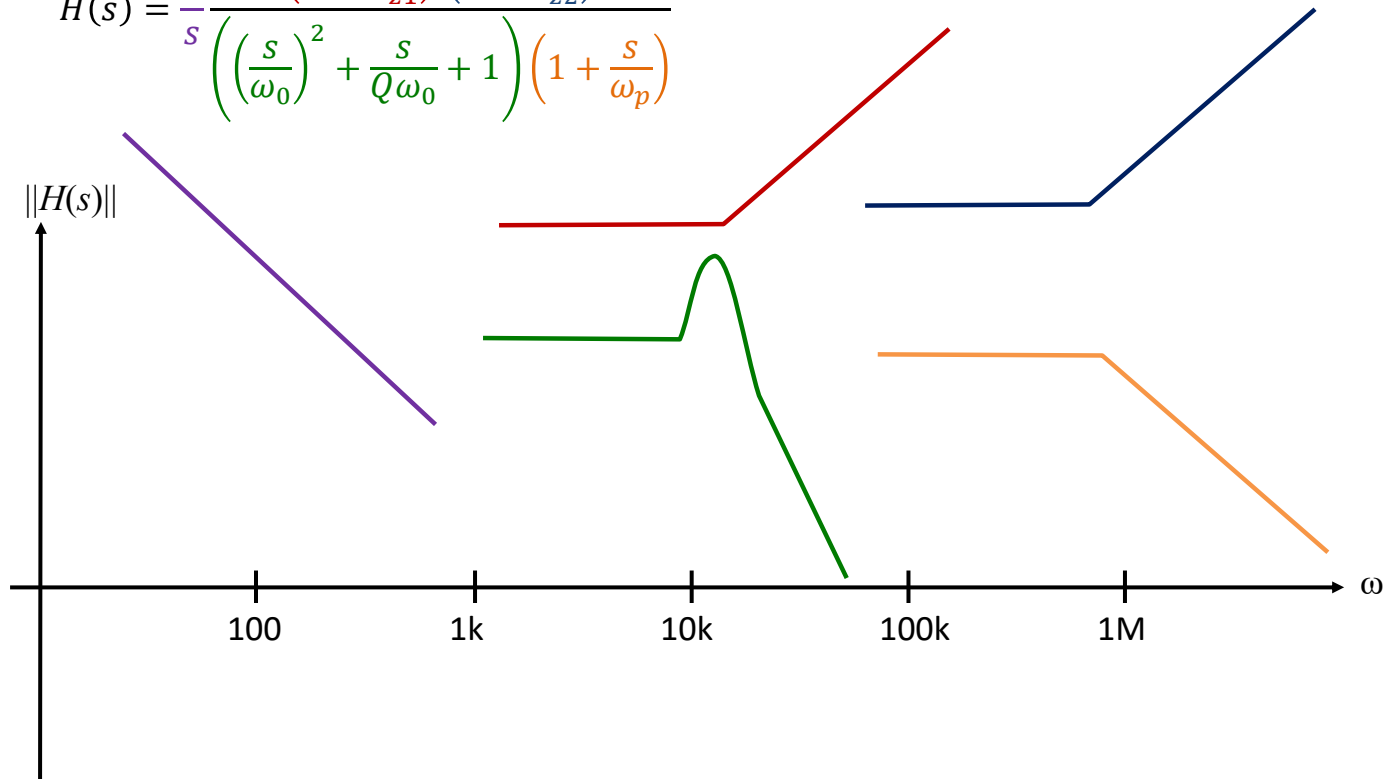


Chapter 8: Converter Transfer Functions

Example

A	w_{z1}	w_{z2}	ω_0	Q	ω_p
1000	100	10k	1k	10	100k

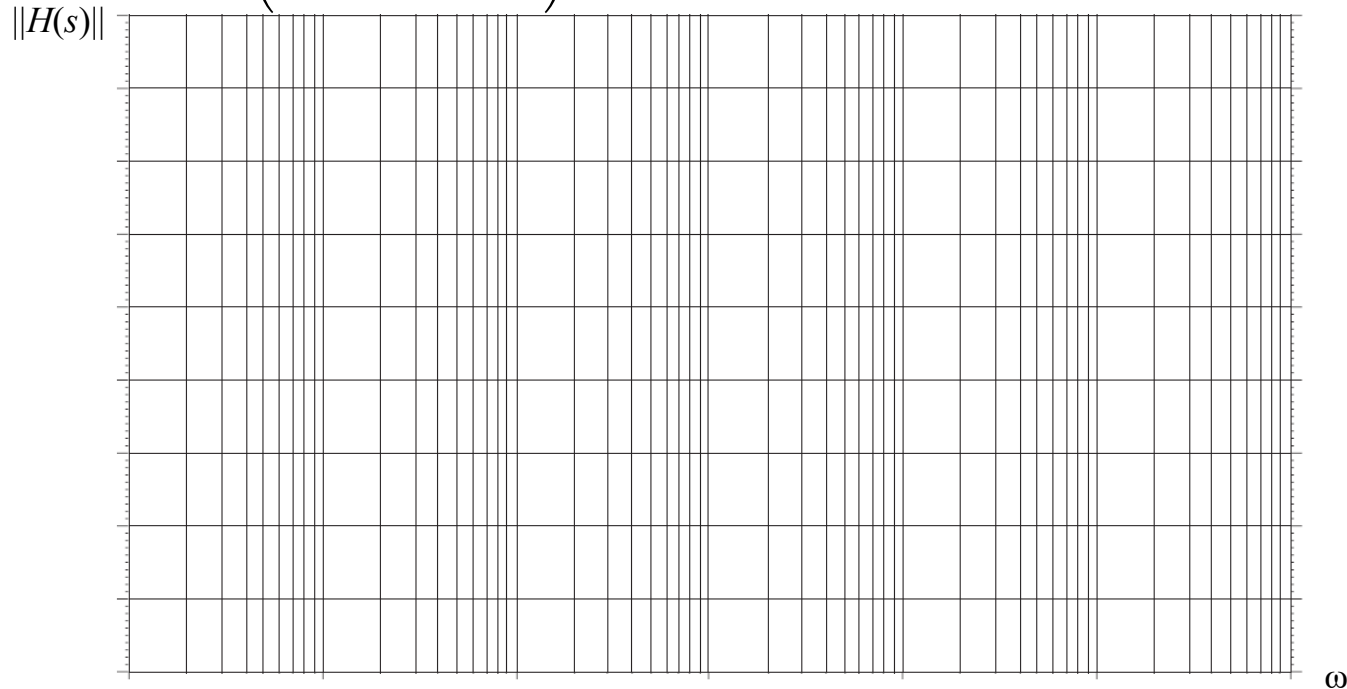
$$H(s) = \frac{A}{s} \frac{\left(1 + \frac{s}{w_{z1}}\right) \left(1 + \frac{s}{w_{z2}}\right)}{\left(\left(\frac{s}{\omega_0}\right)^2 + \frac{s}{Q\omega_0} + 1\right) \left(1 + \frac{s}{\omega_p}\right)}$$



Example

A	w_{z1}	w_{z2}	ω_0	Q	ω_p
1000	100	10k	1k	10	100k

$$H(s) = A \frac{\left(1 + \frac{s}{w_{z1}}\right)\left(1 + \frac{s}{w_{z2}}\right)}{s \left(\left(\frac{s}{\omega_0}\right)^2 + \frac{s}{Q\omega_0} + 1\right)\left(1 + \frac{s}{\omega_p}\right)}$$



ω

MATLAB

```
A = 1000;  
wz1 = 100;  
wz2 = 10e3;  
w0 = 1e3;  
Q = 10;  
wp = 100e3;  
  
s = tf('s');  
  
H = A*(1+s/wz1)*(1+s/wz2)/...  
      (s*(1+s/wp)*((s/w0)^2+s/Q/w0+1));  
  
bode(H)
```

