

**$K_{crit}$  and  $R_{crit}$**  ( $v=Vg$ )

Buck converter:

$$I_L = \frac{V}{R} > \frac{DVg}{R}$$

$$\Delta i_L = \frac{Vg-v}{2L} DT_s \approx \frac{Vg D'}{2L} DT_s$$

is in CCM when  $I_L > \Delta i_L$   
go into DCM when  $I_L < \Delta i_L$

for CCM:

$$\frac{DVg}{R} > \frac{Vg D'}{2L} DT_s$$

$$K = \left\lfloor \frac{2L}{RT_s} > D' \right\rfloor = K_{crit}(D)$$

$$K > K_{crit}(D) \quad \text{for CCM}$$

## DCM Mode Boundary: Summary

$$\begin{array}{lll} K > K_{crit}(D) & \text{or} & R < R_{crit}(D) \quad \text{for CCM} \\ K < K_{crit}(D) & \text{or} & R > R_{crit}(D) \quad \text{for DCM} \end{array}$$

Table 5.1. CCM-DCM mode boundaries for the buck, boost, and buck-boost converters

Converter	$K_{crit}(D)$	$\max_{0 \leq D \leq 1} (K_{crit})$	$R_{crit}(D)$	$\min_{0 \leq D \leq 1} (R_{crit})$
Buck	$(1-D)$	1	$\frac{2L}{(1-D)T_s}$	$2 \frac{L}{T_s}$
Boost	$D(1-D)^2$	$\frac{4}{27}$	$\frac{2L}{D(1-D)^2 T_s}$	$\frac{27}{2} \frac{L}{T_s}$
Buck-boost	$(1-D)^2$	1	$\frac{2L}{(1-D)^2 T_s}$	$2 \frac{L}{T_s}$

## Finding the Conversion Ratio $M(D,K)$

Analysis techniques for the discontinuous conduction mode:

Inductor volt-second balance

$$\langle v_L \rangle = \frac{1}{T_s} \int_0^{T_s} v_L(t) dt = 0$$

Capacitor charge balance

$$\langle i_C \rangle = \frac{1}{T_s} \int_0^{T_s} i_C(t) dt = 0$$

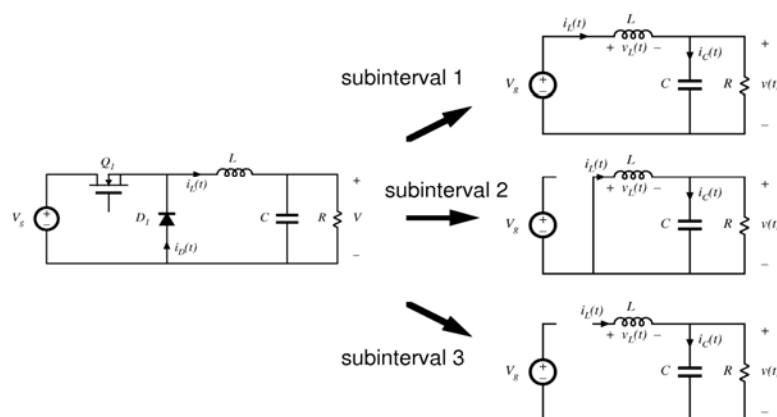
Small ripple approximation sometimes applies:

$$v(t) \approx V \quad \text{because} \quad \Delta v \ll V$$

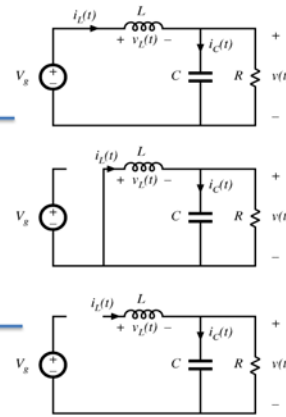
$$i(t) \approx I \quad \text{is a poor approximation when} \quad \Delta i > I$$

Converter steady-state equations obtained via charge balance on each capacitor and volt-second balance on each inductor. Use care in applying small ripple approximation.

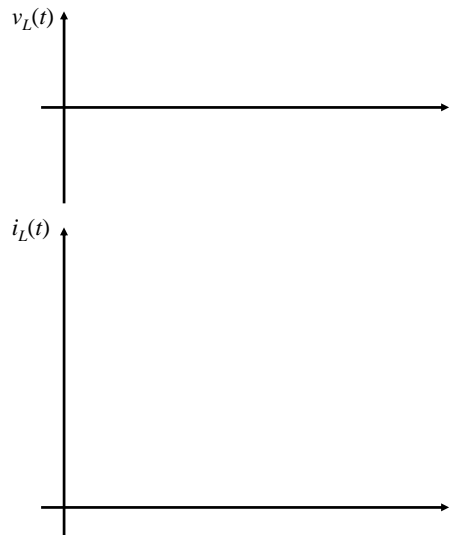
## Buck Converter in DCM



## Subinterval Analysis



## Waveforms in DCM



## Solving $M(D,K)$

Two equations and two unknowns ( $V$  and  $D_2$ ):

$$V = V_g \frac{D_1}{D_1 + D_2} \quad (\text{from inductor volt-second balance})$$

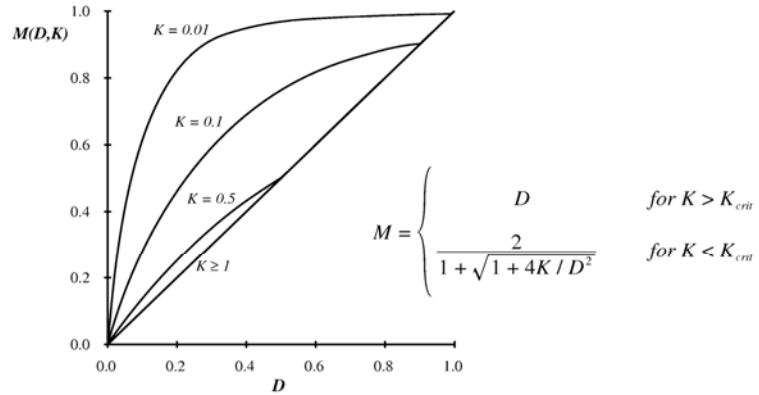
$$\frac{V}{R} = \frac{D_1 T_s}{2L} (D_1 + D_2) (V_g - V) \quad (\text{from capacitor charge balance})$$

Eliminate  $D_2$ , solve for  $V$ :

$$\frac{V}{V_g} = \frac{2}{1 + \sqrt{1 + 4K / D_1^2}}$$

where  $K = 2L / RT_s$   
 valid for  $K < K_{crit}$

## Buck Converter $M(D,K)$

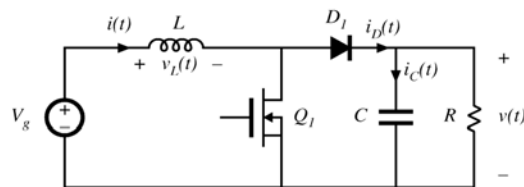


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Chapter 5: Discontinuous conduction mode

## Boost Converter in DCM



Mode boundary:

$$\begin{aligned} I &> \Delta i_L \text{ for CCM} \\ I &< \Delta i_L \text{ for DCM} \end{aligned}$$

Previous CCM soln:

$$I = \frac{V_g}{D^2 R} \quad \Delta i_L = \frac{V_g}{2L} DT_s$$

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Chapter 5: Discontinuous conduction mode

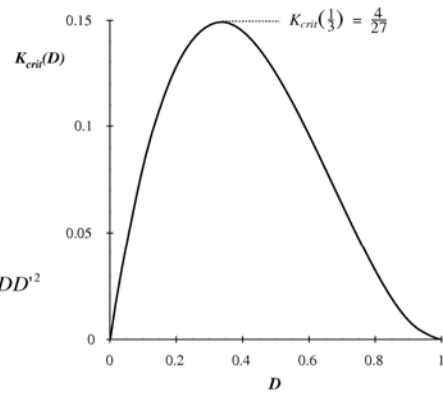
## Boost DCM Boundary

$$\frac{V_g}{D^2 R} > \frac{DT_s V_g}{2L} \quad \text{for CCM}$$

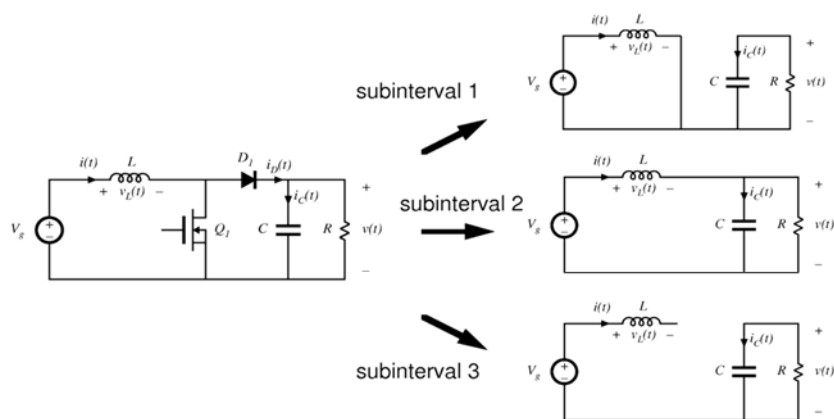
$$\frac{2L}{RT_s} > DD^2 \quad \text{for CCM}$$

$$\begin{aligned} K &> K_{crit}(D) && \text{for CCM} \\ K &< K_{crit}(D) && \text{for DCM} \end{aligned}$$

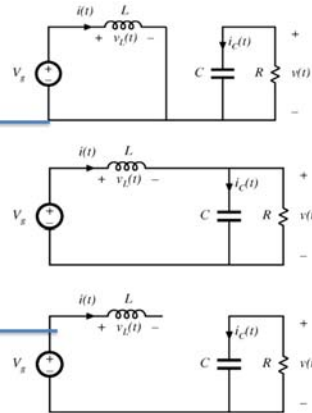
where  $K = \frac{2L}{RT_s}$  and  $K_{crit}(D) = DD^2$



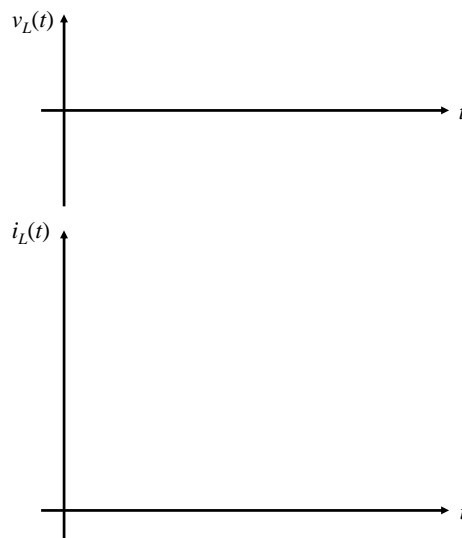
## Boost Converter Subintervals



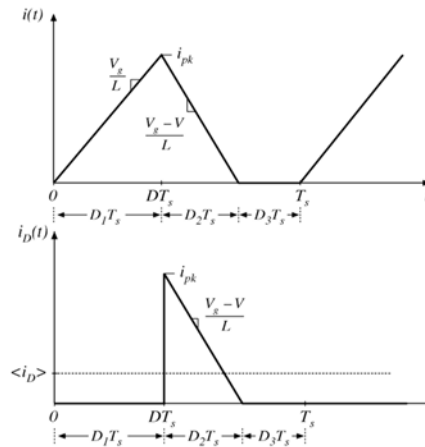
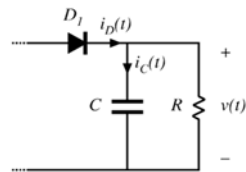
## Boost Conversion Ratio



## Boost Waveforms in DCM



## Boost Cap-Charge Balance



## Boost DCM Conversion Ratio

$$V^2 - VV_g - \frac{V_g^2 D_1^2}{K} = 0$$

Use quadratic formula:

$$\frac{V}{V_g} = \frac{1 \pm \sqrt{1 + 4D_1^2 / K}}{2}$$

Note that one root leads to positive  $V$ , while other leads to negative  $V$ . Select positive root:

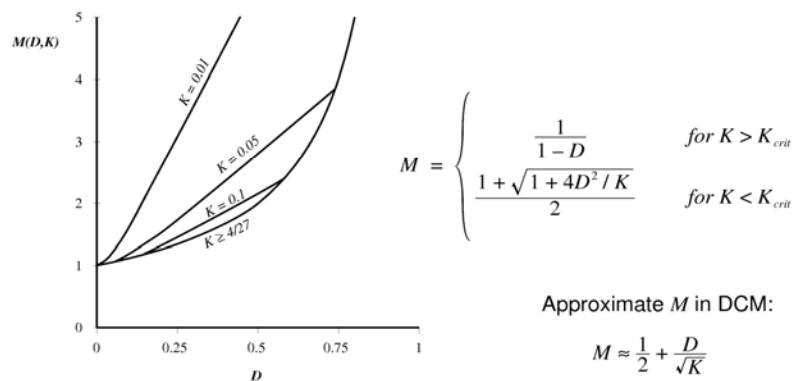
$$\frac{V}{V_g} = M(D_1, K) = \frac{1 + \sqrt{1 + 4D_1^2 / K}}{2}$$

where  $K = 2L / RT_s$   
valid for  $K < K_{crit}(D)$

Transistor duty cycle  $D$  = interval 1 duty cycle  $D_1$



## Boost Conversion Ratio



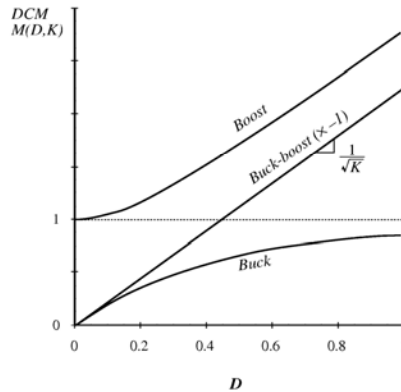
## Summary of DCM Characteristics

Table 5.2. Summary of CCM-DCM characteristics for the buck, boost, and buck-boost converters

Converter	$K_{crit}(D)$	DCM $M(D,K)$	DCM $D_2(D,K)$	CCM $M(D)$
Buck	$(1-D)$	$\frac{2}{1 + \sqrt{1 + 4K/D^2}}$	$\frac{K}{D} M(D,K)$	$D$
Boost	$D(1-D)^2$	$\frac{1 + \sqrt{1 + 4D^2/K}}{2}$	$\frac{K}{D} M(D,K)$	$\frac{1}{1-D}$
Buck-boost	$(1-D)^2$	$-\frac{D}{\sqrt{K}}$	$\sqrt{K}$	$-\frac{D}{1-D}$

with  $K = 2L / RT_s$ , DCM occurs for  $K < K_{crit}$ .

## Summary of DCM Characteristics



- DCM buck and boost characteristics are asymptotic to  $M = 1$  and to the DCM buck-boost characteristic
- DCM buck-boost characteristic is linear
- CCM and DCM characteristics intersect at mode boundary. Actual  $M$  follows characteristic having larger magnitude
- DCM boost characteristic is nearly linear

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Chapter 5: Discontinuous conduction mode

## Chapter 5 Summary

1. The discontinuous conduction mode occurs in converters containing current- or voltage-unidirectional switches, when the inductor current or capacitor voltage ripple is large enough to cause the switch current or voltage to reverse polarity.
2. Conditions for operation in the discontinuous conduction mode can be found by determining when the inductor current or capacitor voltage ripples and dc components cause the switch on-state current or off-state voltage to reverse polarity.
3. The dc conversion ratio  $M$  of converters operating in the discontinuous conduction mode can be found by application of the principles of inductor volt-second and capacitor charge balance.
4. Extra care is required when applying the small-ripple approximation. Some waveforms, such as the output voltage, should have small ripple which can be neglected. Other waveforms, such as one or more inductor currents, may have large ripple that cannot be ignored.
5. The characteristics of a converter changes significantly when the converter enters DCM. The output voltage becomes load-dependent, resulting in an increase in the converter output impedance.