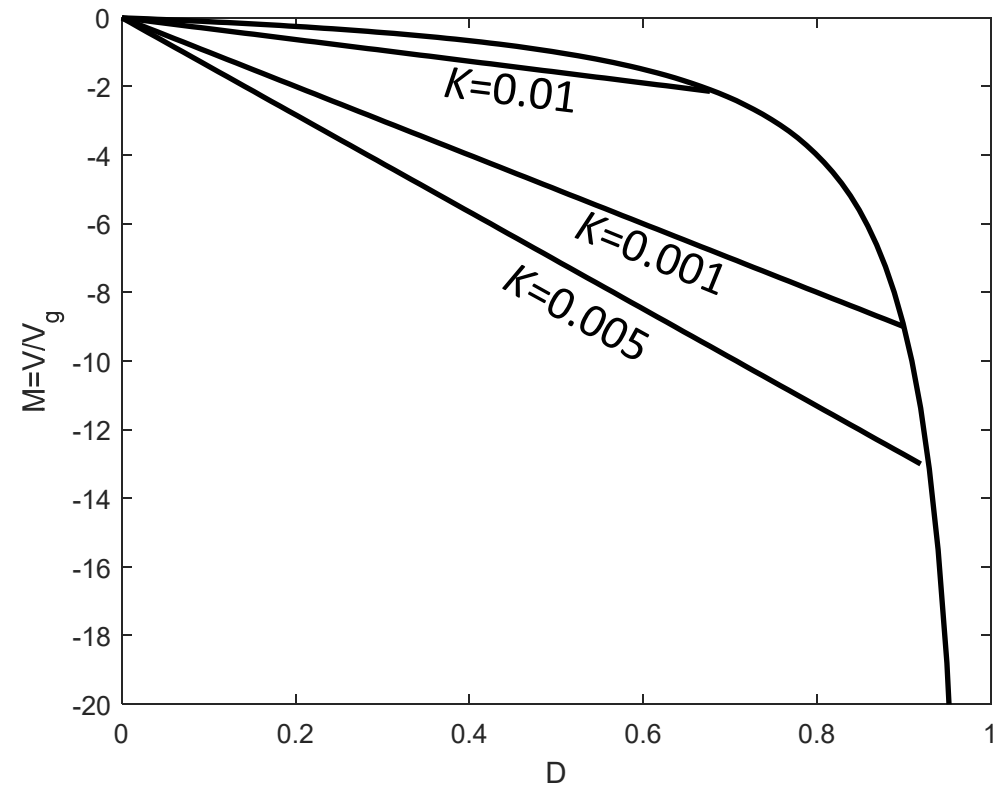


Cuk DCM M(D,K)



Summary of DCM Characteristics

Table 5.2. Summary of CCM-DCM characteristics for the buck, boost, and buck-boost converters

Converter	$K_{crit}(D)$	DCM $M(D,K)$	DCM $D_2(D,K)$	CCM $M(D)$
Buck	$(1 - D)$	$\frac{2}{1 + \sqrt{1 + 4K / D^2}}$	$\frac{K}{D} M(D,K)$	D
Boost	$D (1 - D)^2$	$\frac{1 + \sqrt{1 + 4D^2 / K}}{2}$	$\frac{K}{D} M(D,K)$	$\frac{1}{1 - D}$
Buck-boost	$(1 - D)^2$	$-\frac{D}{\sqrt{K}}$	\sqrt{K}	$-\frac{D}{1 - D}$

with $K = 2L / RT_s$. DCM occurs for $K < K_{crit}$.

Chapter 6

CONVERTER TOPOLOGIES

6.2 - A Short List of Converters

An infinite number of converters are possible, which contain switches embedded in a network of inductors and capacitors

Two simple classes of converters are listed here:

- Single-input single-output converters containing a single inductor. The switching period is divided into two subintervals. This class contains eight converters.
- Single-input single-output converters containing two inductors. The switching period is divided into two subintervals. Several of the more interesting members of this class are listed.

Single Input/Output/Inductor Converters

- Use switches to connect inductor between source and load, in one manner during first subinterval and in another during second subinterval
- There are a limited number of ways to do this, so all possible combinations can be found
- After elimination of degenerate and redundant cases, eight converters are found:

dc-dc converters

buck boost buck-boost noninverting buck-boost

dc-ac converters

bridge Watkins-Johnson

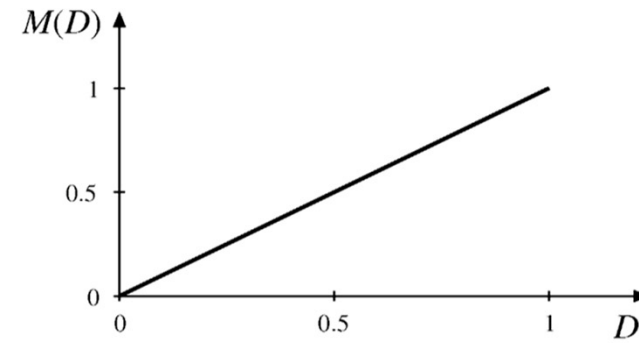
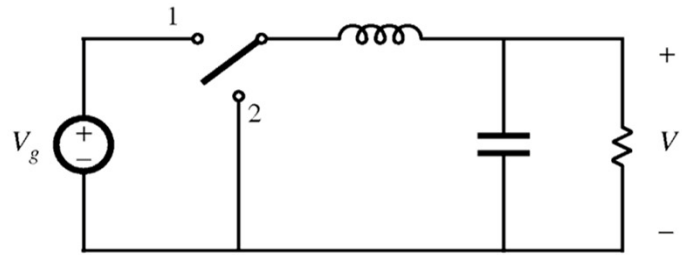
ac-dc converters

current-fed bridge inverse of Watkins-Johnson

Unipolar Output Converters

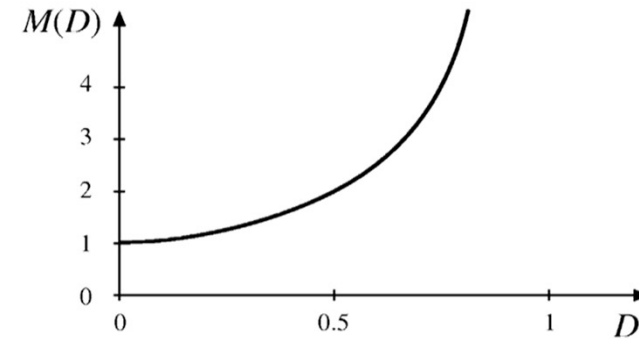
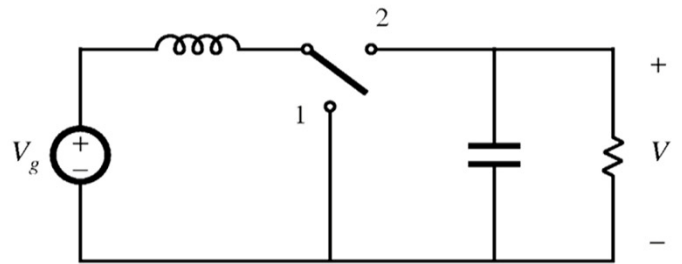
1. Buck

$$M(D) = D$$



2. Boost

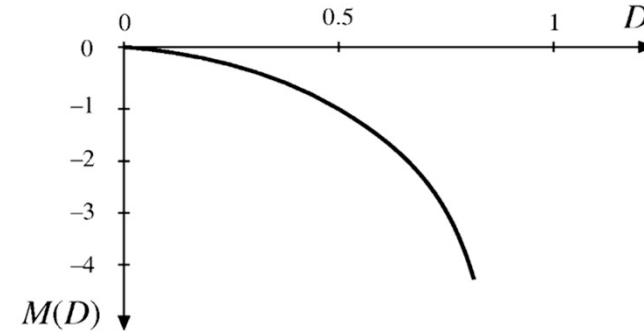
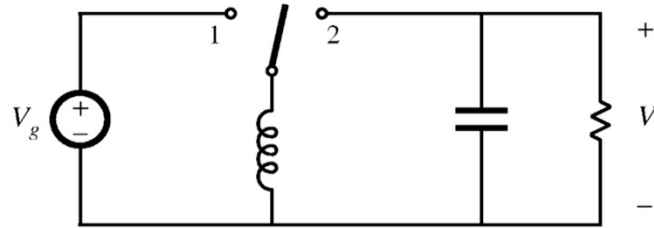
$$M(D) = \frac{1}{1-D}$$



Unipolar Output Converters (cont.)

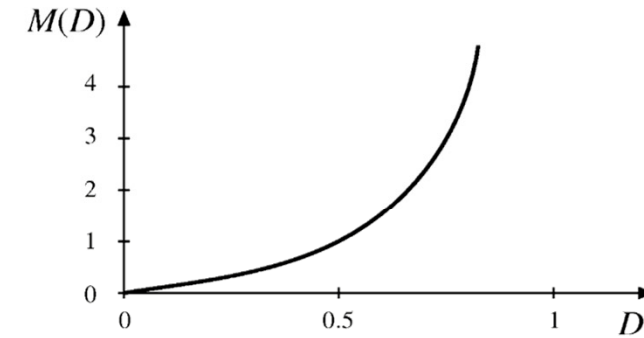
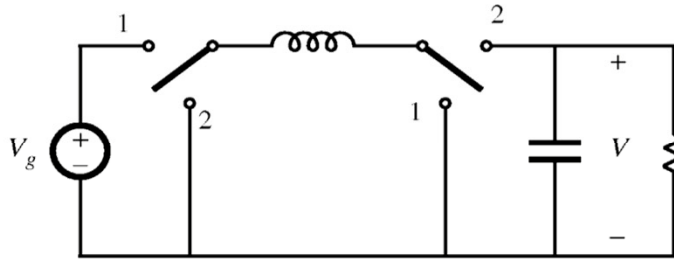
3. Buck-boost

$$M(D) = -\frac{D}{1-D}$$



4. Noninverting buck-boost

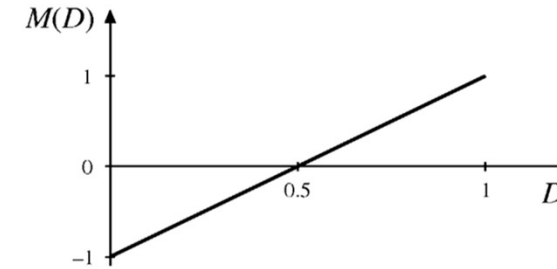
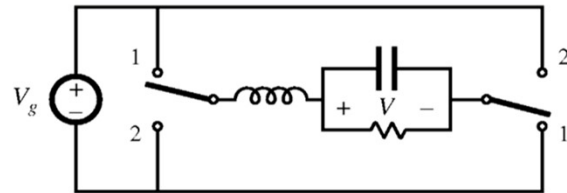
$$M(D) = \frac{D}{1-D}$$



Bipolar Output Converters

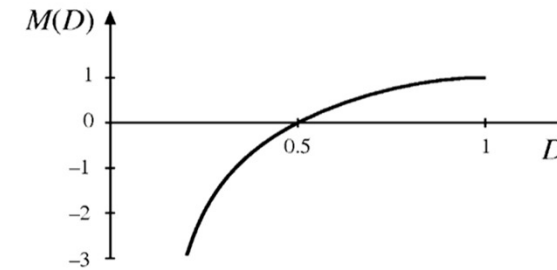
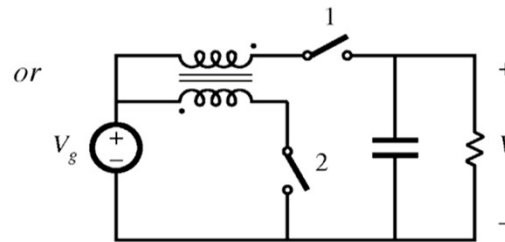
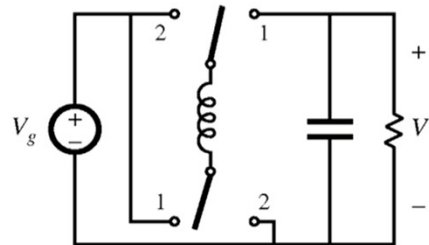
5. Bridge

$$M(D) = 2D - 1$$



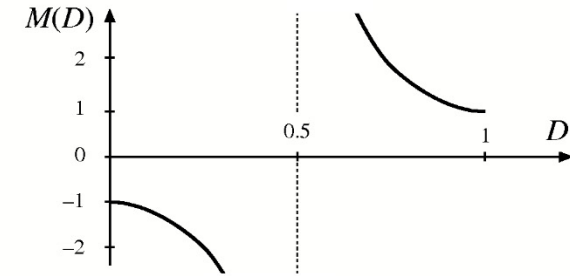
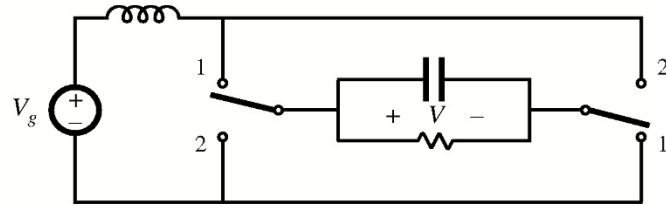
6. Watkins-Johnson

$$M(D) = \frac{2D - 1}{D}$$

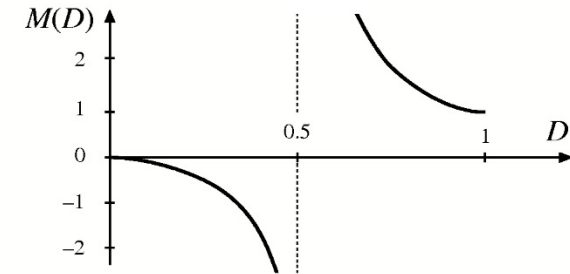
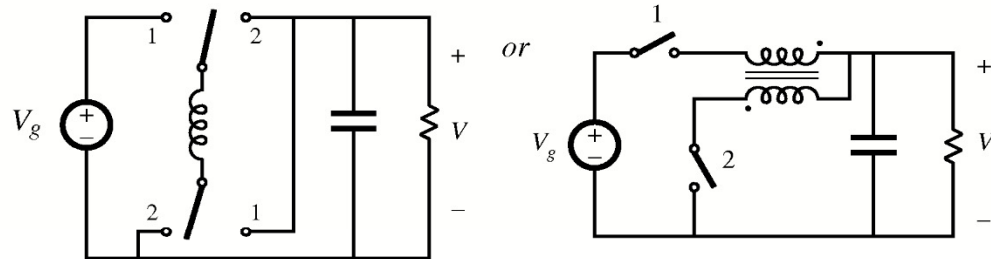


Bipolar Output Converters (cont.)

7. Current-fed bridge $M(D) = \frac{1}{2D-1}$



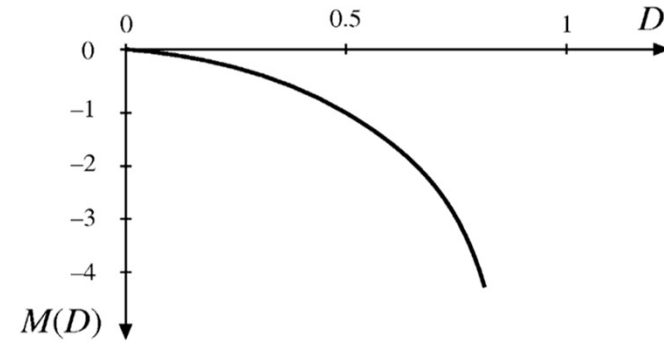
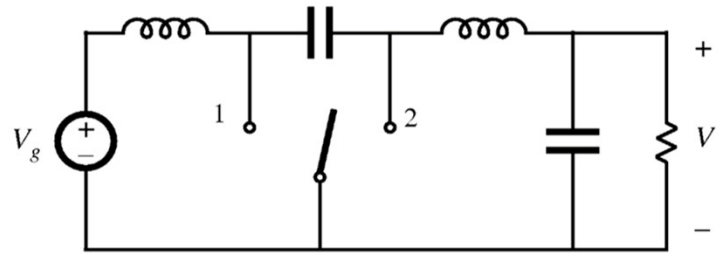
8. Inverse of Watkins-Johnson $M(D) = \frac{D}{2D-1}$



Example Two-Inductor Converters

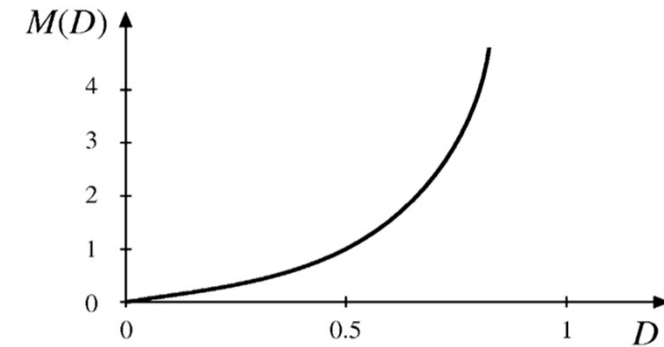
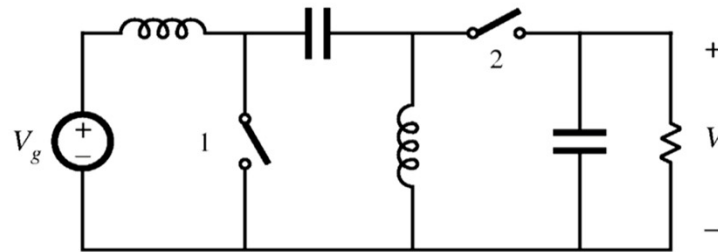
1. *Ćuk*

$$M(D) = -\frac{D}{1-D}$$



2. *SEPIC*

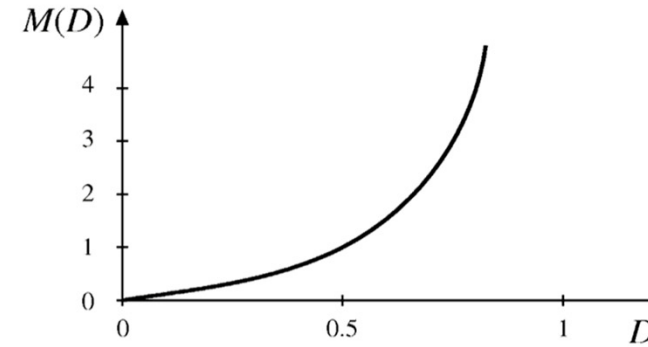
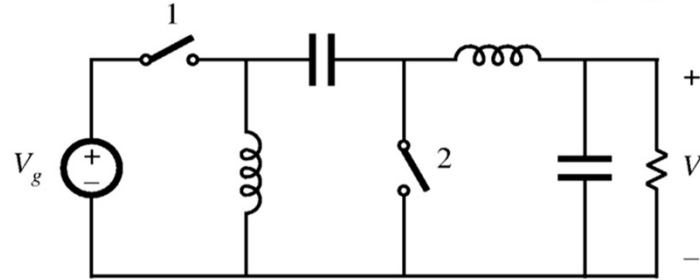
$$M(D) = \frac{D}{1-D}$$



Example Two-Inductor Converters (cont.)

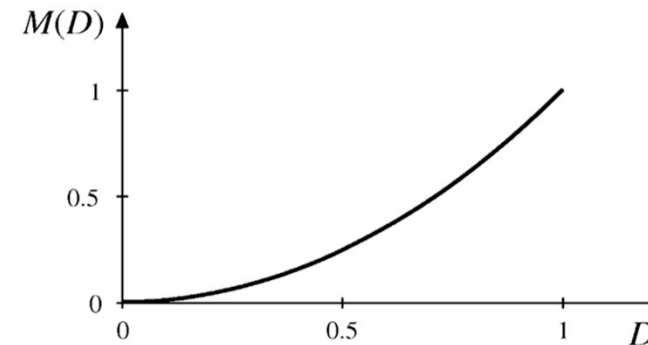
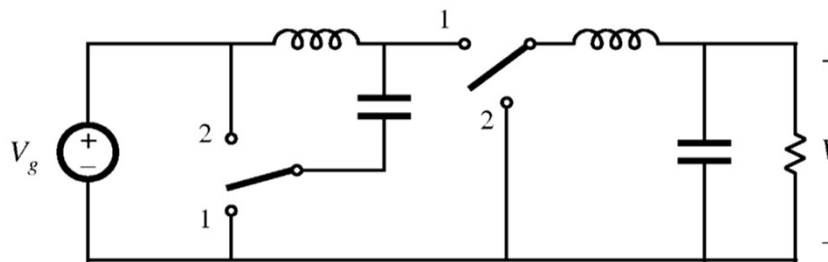
3. Inverse of SEPIC

$$M(D) = \frac{D}{1-D}$$



4. Buck²

$$M(D) = D^2$$

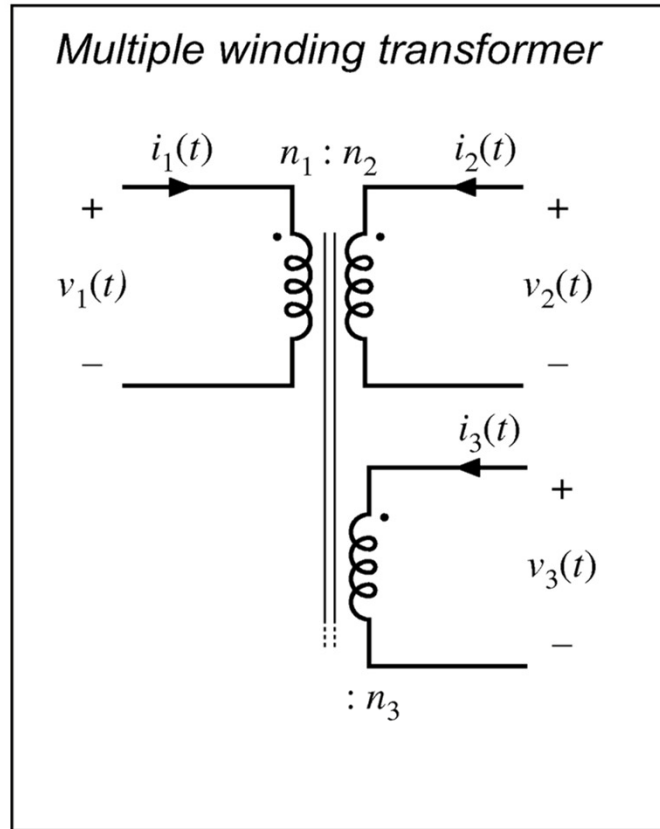


6.3 - Transformer Isolation

Objectives:

- Isolation of input and output ground connections, to meet safety requirements
- Reduction of transformer size by incorporating high frequency isolation transformer inside converter
- Minimization of current and voltage stresses when a large step-up or step-down conversion ratio is needed —use transformer turns ratio
- Obtain multiple output voltages via multiple transformer secondary windings and multiple converter secondary circuits

Ideal Transformer Model



$$\frac{v_1(t)}{n_1} = \frac{v_2(t)}{n_2} = \frac{v_3(t)}{n_3} = \dots$$

$$0 = n_1 i_1'(t) + n_2 i_2(t) + n_3 i_3(t) + \dots$$

Transformer Saturation

