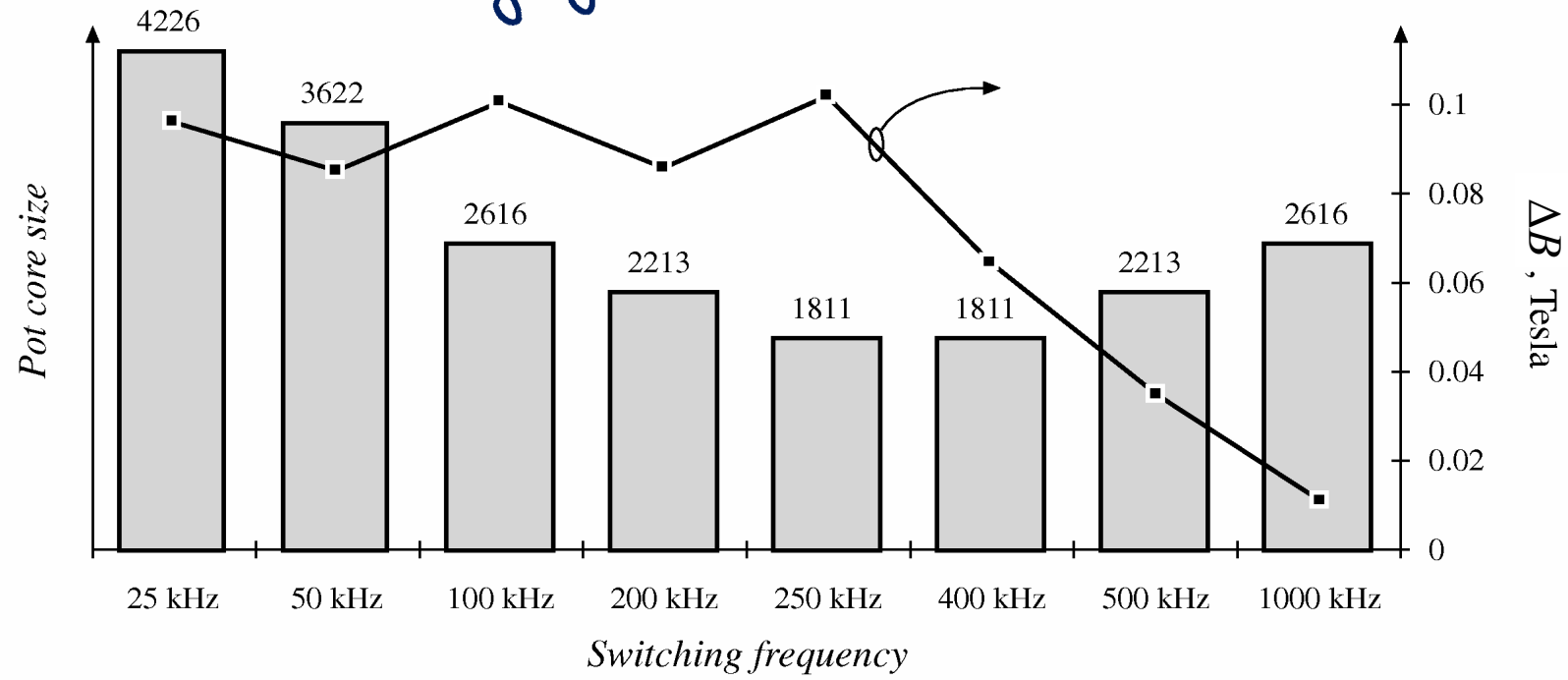
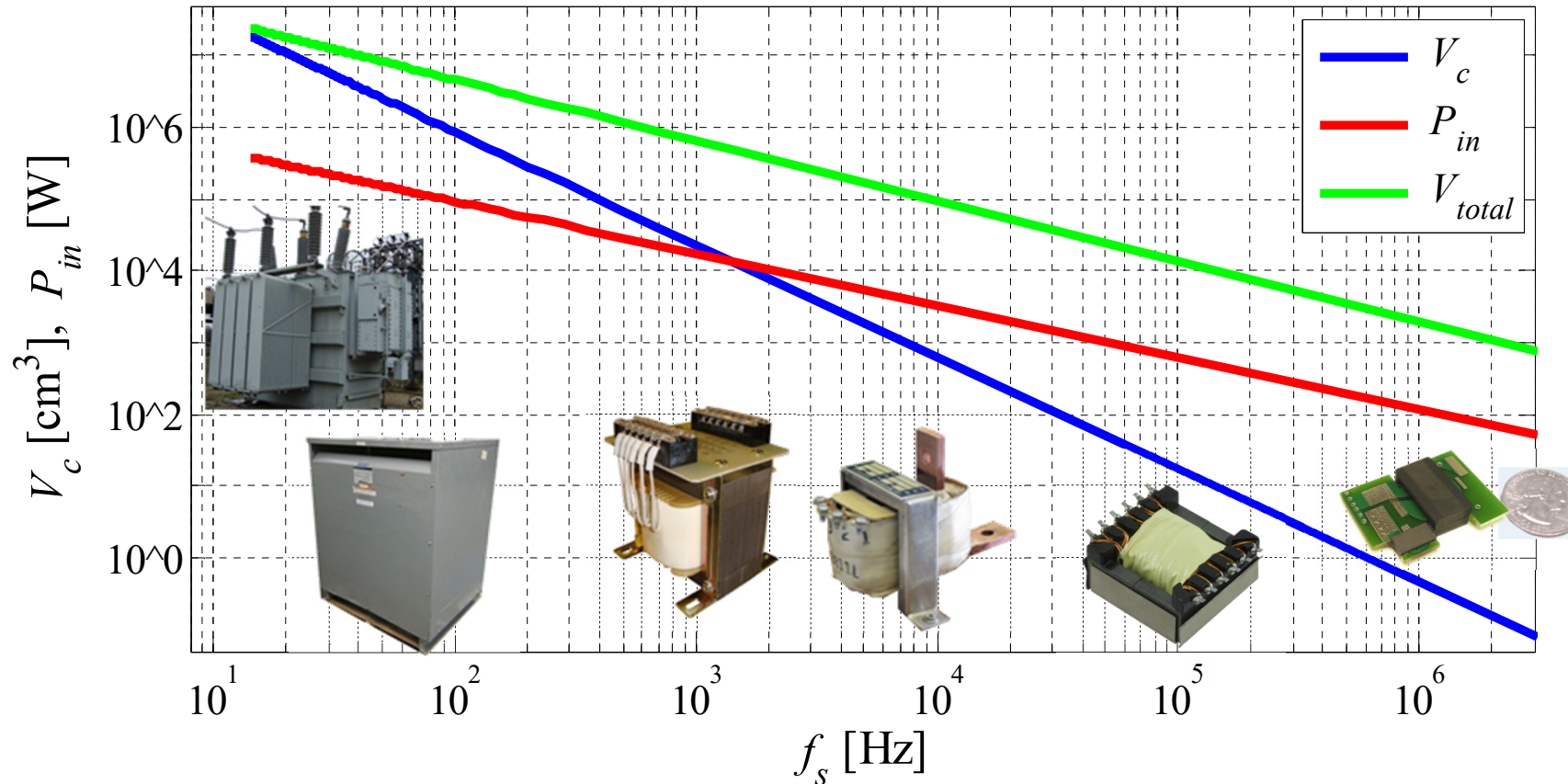


XF Size vs f_s

For a constant P_{XF}
using kyle :



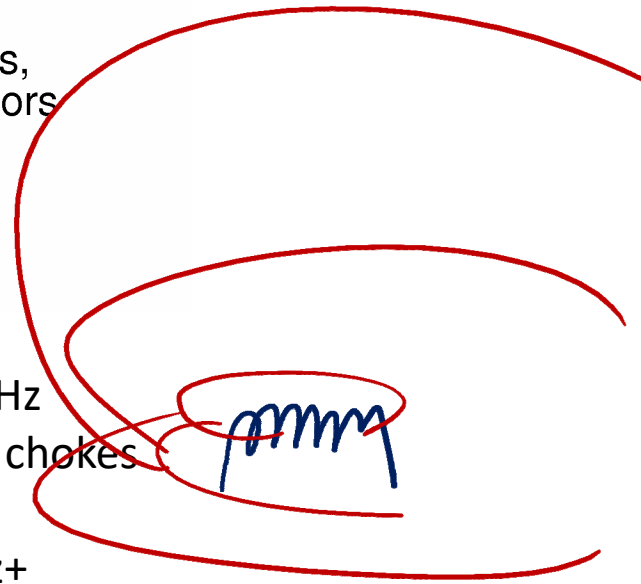
Transformer Size Comparison



- V_{total} is total magnetic volume required to process 500 kW
- $K_U = .5, B_{pk} = 200 \text{ mT}, n_l = 70, V_g = 400 \text{ V}, \eta_{XF} = 95\%$

Some Example Core Materials

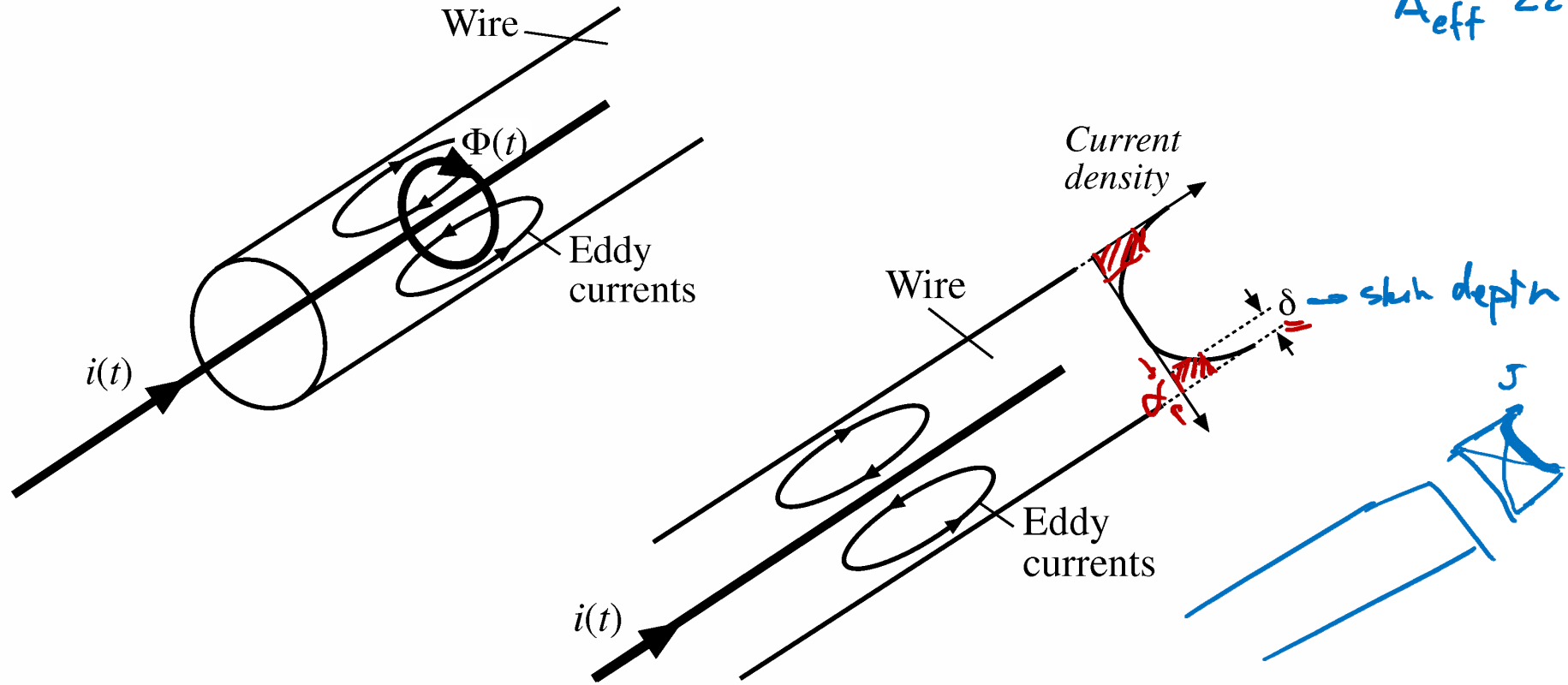
Core type	B_{sat}	Relative core loss	Applications
Laminations iron, silicon steel	1.5 - 2.0 T	high	50-60 Hz transformers, inductors
→ Powdered cores powdered iron, molypermalloy	0.6 - 0.8 T	medium	1 kHz transformers, 100 kHz filter inductors
→ Ferrite Manganese-zinc, Nickel-zinc	0.25 - 0.5 T	low	20 kHz - 1 MHz transformers, ac inductors
Laminated nanocrystalline	1.5 - 2.2T	low	5kHz – 500 kHz Transformers, CM chokes
Air	N/A	zero	1 – 100 MHz+ ac inductors
Low- μ powdered iron/carbonyl	0.5-1.2 T (soft)	very low	1 – 100 MHz+ ac inductors



$\mu = 2-50$

Skin Effect

$$R_{DC} = \rho \frac{l}{A_w}$$



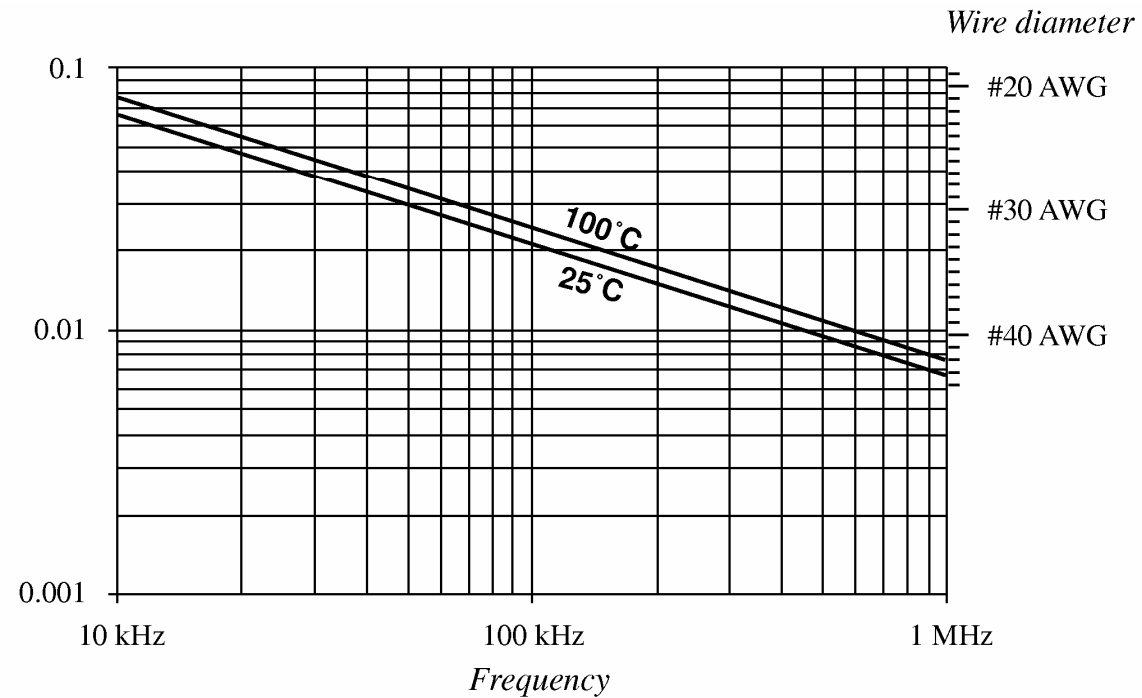
Skin Depth

Penetration
depth δ , cm

$$\delta = \sqrt{\frac{\rho}{\pi \mu f}}$$

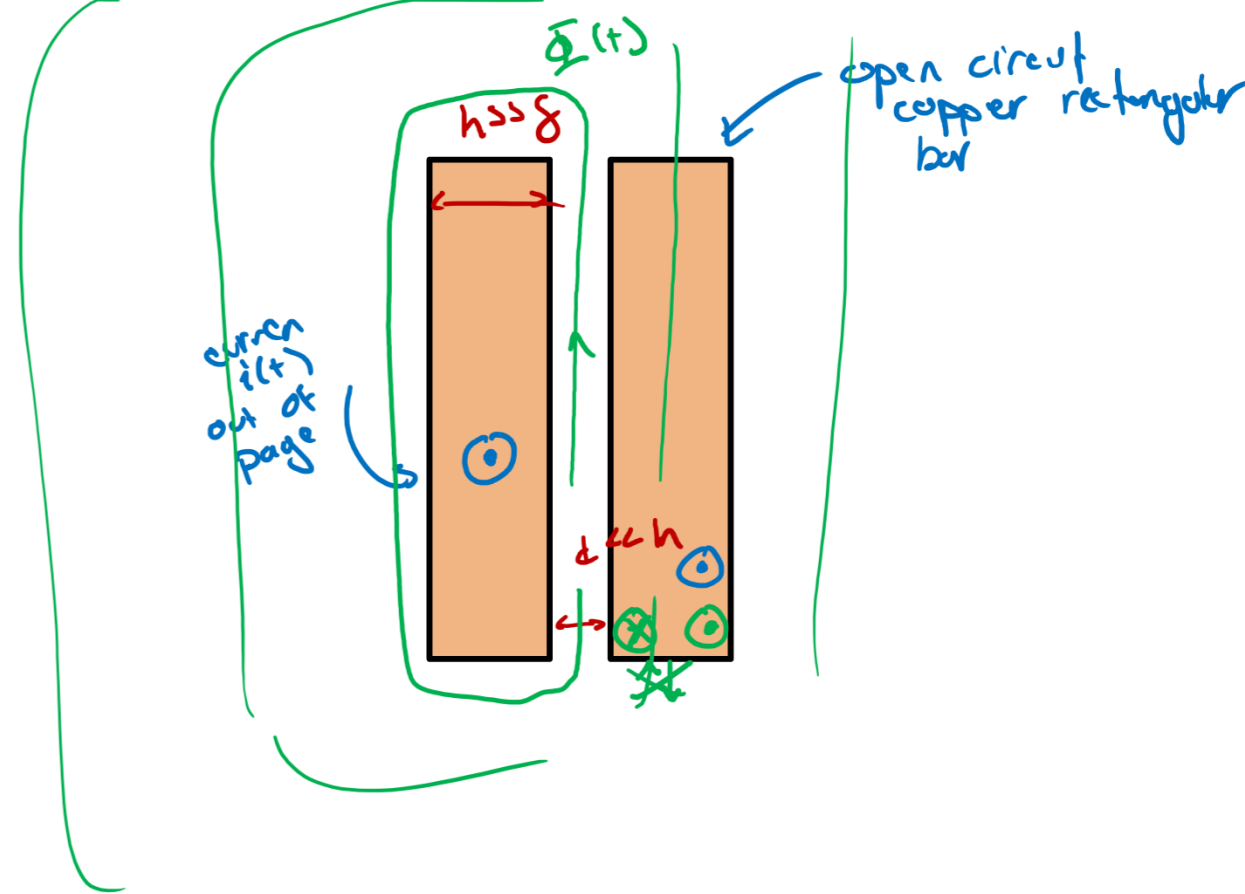
For copper at room
temperature:

$$\delta = \frac{7.5}{\sqrt{f}} \text{ cm}$$

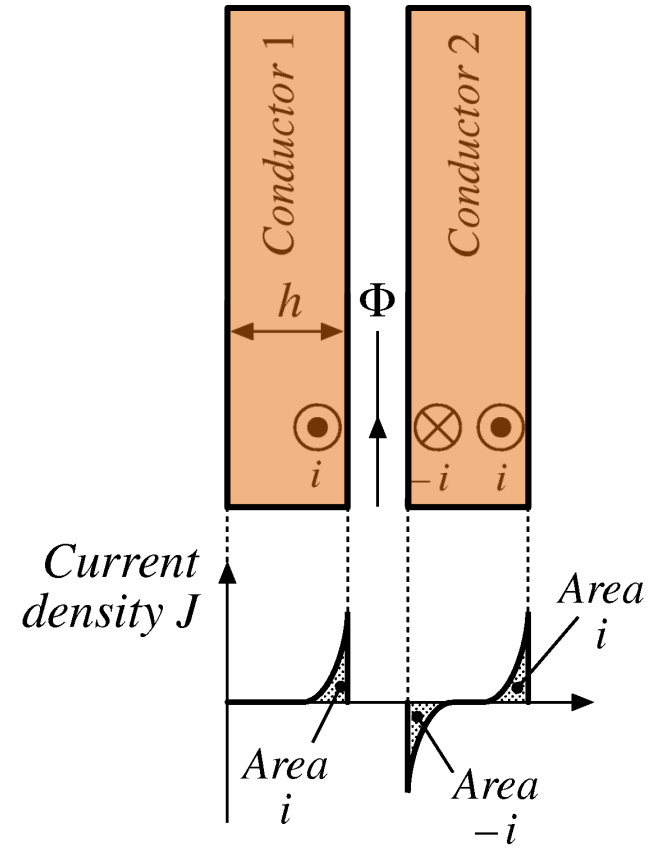


Proximity Effect

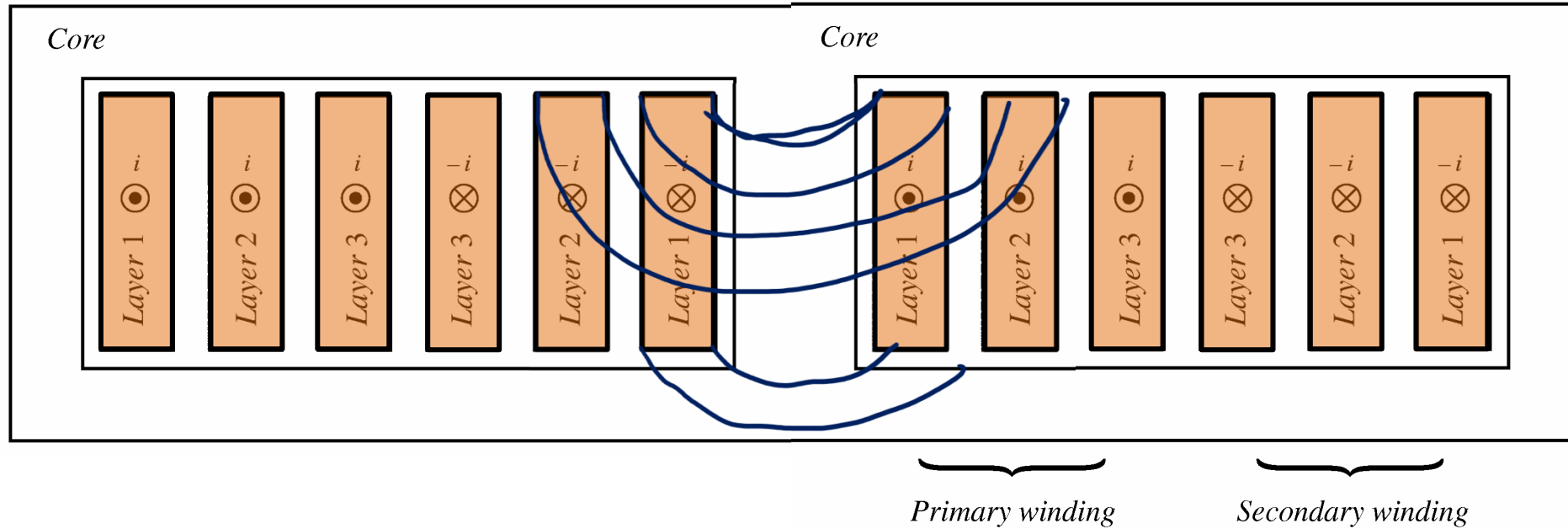
For closely spaced windings with fields from each winding in adjacent conductors generate eddy currents with $h \gg \delta$, eddy currents



Proximity Effect

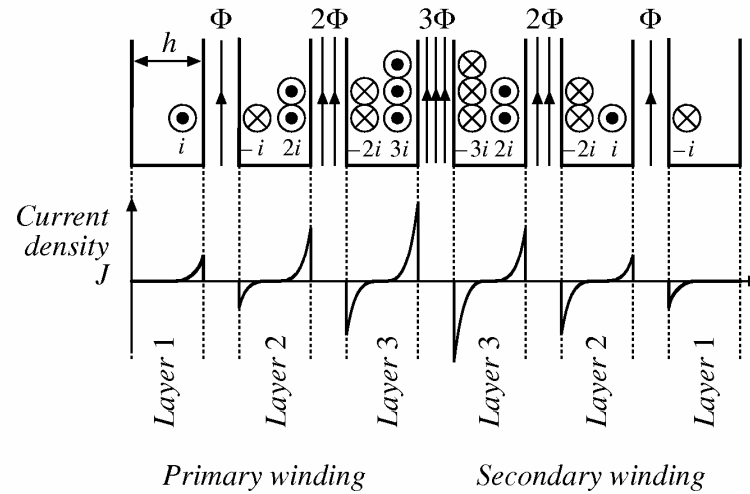
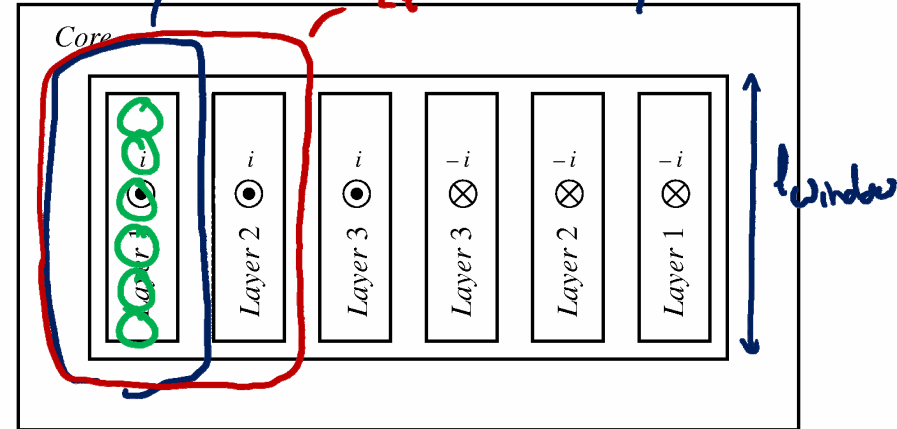


Two-Winding Transformer Example

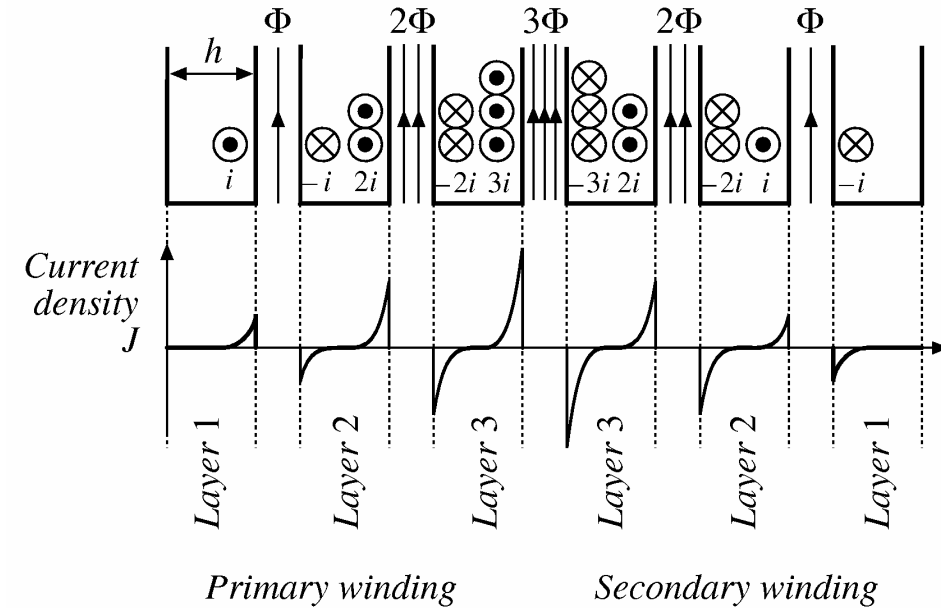


Current Distribution

Ampere: $I_{enc} = \int H \cdot dl$
 $i \approx H_{window} l_{window}$
 $2i \approx H_{window,2} l_{window,2}$ *l_{window,2} > l_{window}*



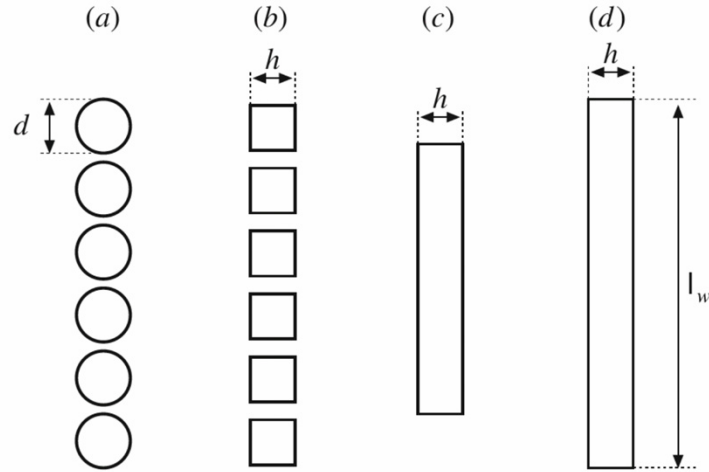
High Frequency Estimation



Power loss P_m in layer m is:

$$P_m = I^2 \left[(m-1)^2 + m^2 \right] \left(\frac{h}{\delta} R_{dc} \right)$$

Equivalent Foil Winding Loss



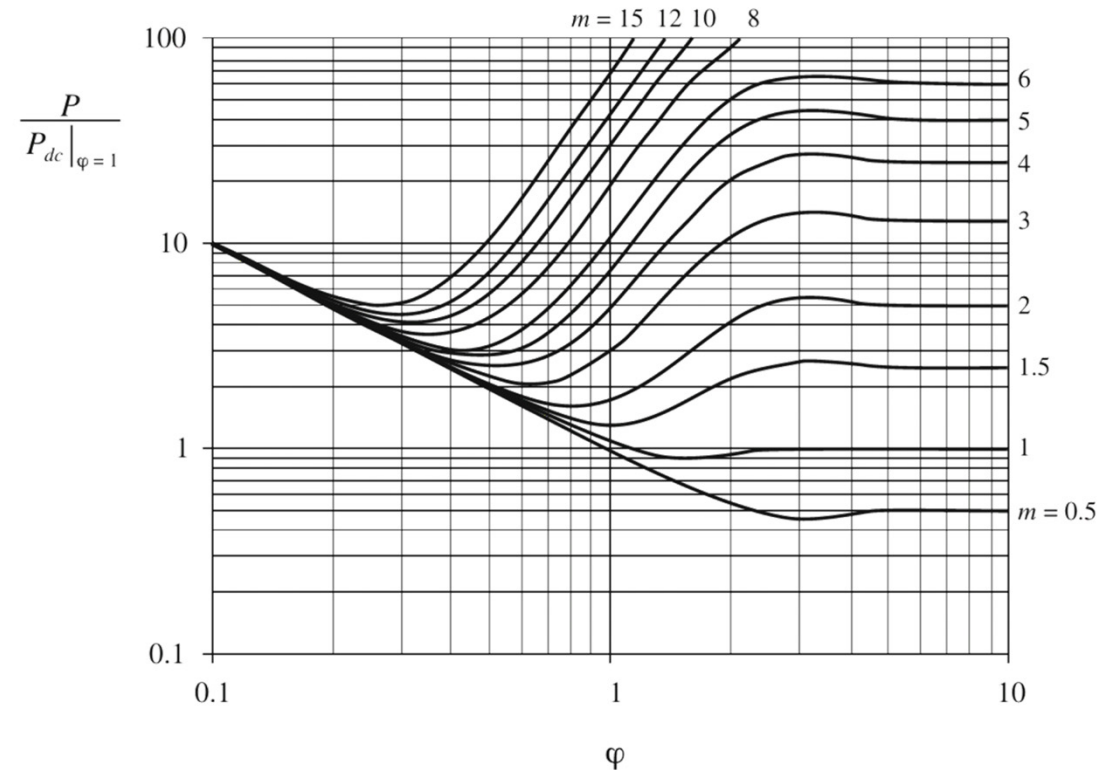
$$\eta = \sqrt{\frac{\pi}{4}} d \frac{n_\ell}{\ell_w} \quad \delta' = \frac{\delta}{\sqrt{\eta}}$$

$$\varphi = \frac{h}{\delta'} = \sqrt{\eta} \sqrt{\frac{\pi}{4}} \frac{d}{\delta}$$

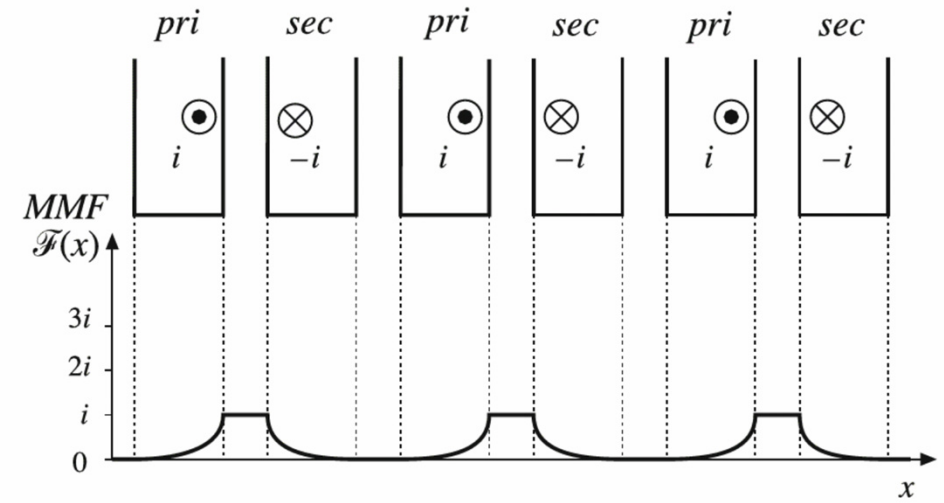
$$P = R_{dc} \frac{\varphi}{n_\ell^2} \left[(\mathcal{F}^2(h) + \mathcal{F}^2(0)) G_1(\varphi) - 4 \mathcal{F}(h) \mathcal{F}(0) G_2(\varphi) \right]$$

$$G_1(\varphi) = \frac{\sinh(2\varphi) + \sin(2\varphi)}{\cosh(2\varphi) - \cos(2\varphi)}$$

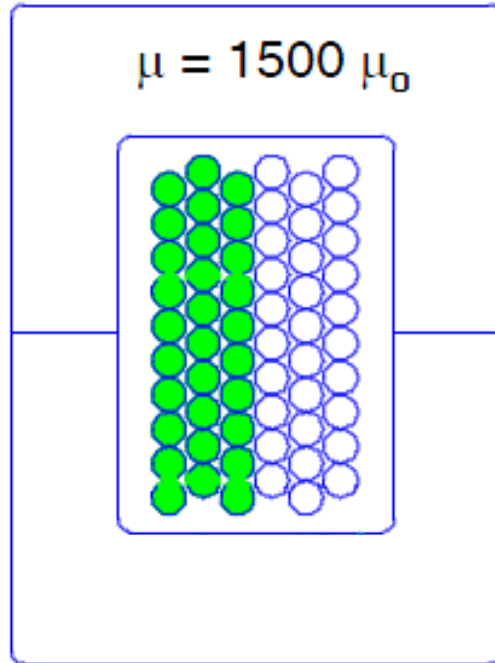
$$G_2(\varphi) = \frac{\sinh(\varphi) \cos(\varphi) + \cosh(\varphi) \sin(\varphi)}{\cosh(2\varphi) - \cos(2\varphi)}$$



Interleaving



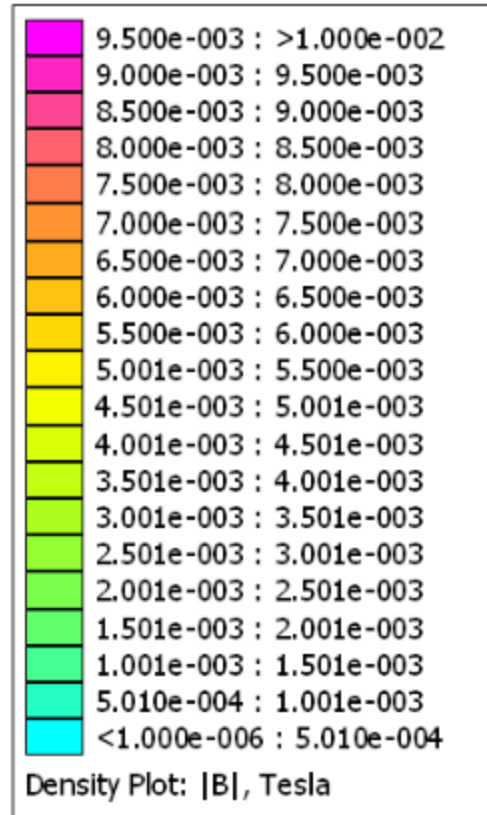
Simulation Example



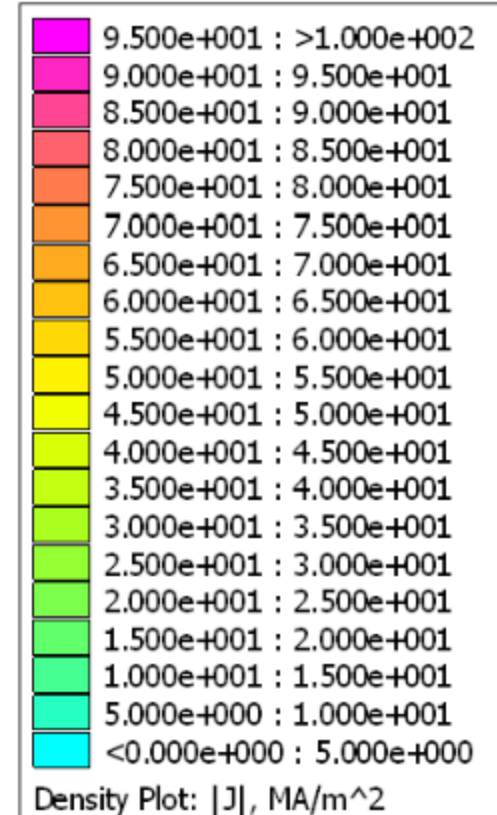
- AWG#30 copper wire
 - Diameter $d = 0.294$ mm
 - $d = \delta$ at around 50 kHz
- 1:1 transformer
 - Primary and secondary are the same, 30 turns in 3 layers
- Sinusoidal currents,
 $I_{1rms} = I_{2rms} = 1$ A

Numerical field and current density solutions using FEMM (Finite Element Method Magnetics), a free 2D solver, <http://www.femm.info/wiki/HomePage>

Flux density magnitude

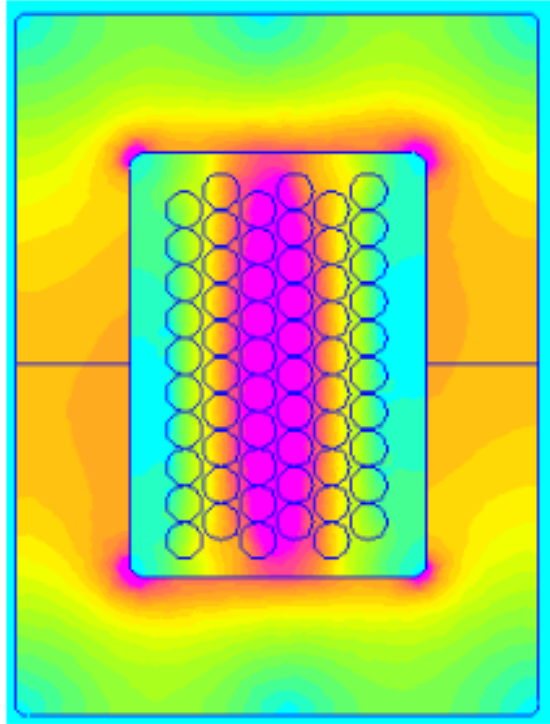


Current density magnitude

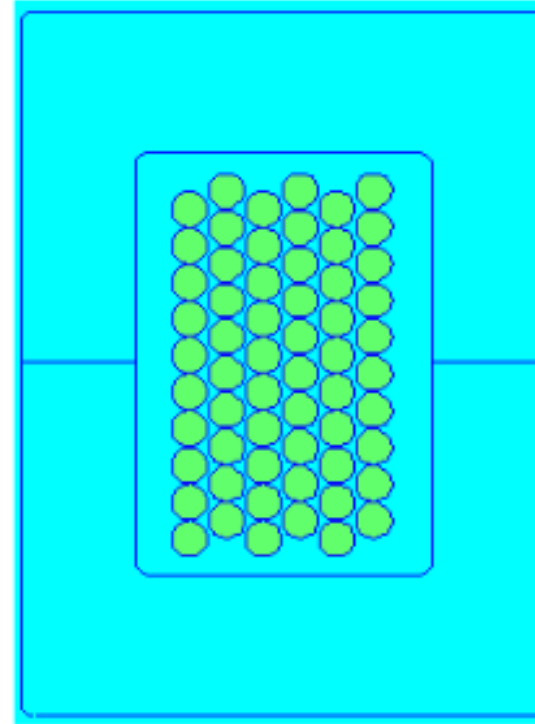


Frequency: 1 kHz

Flux density

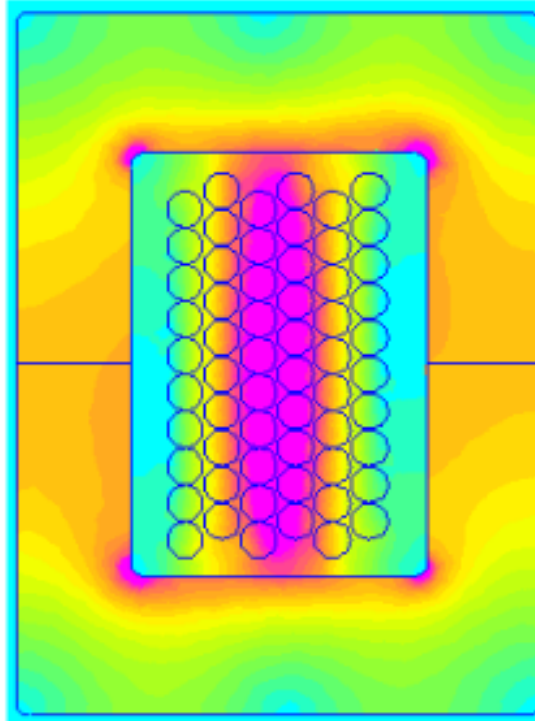


Current Density

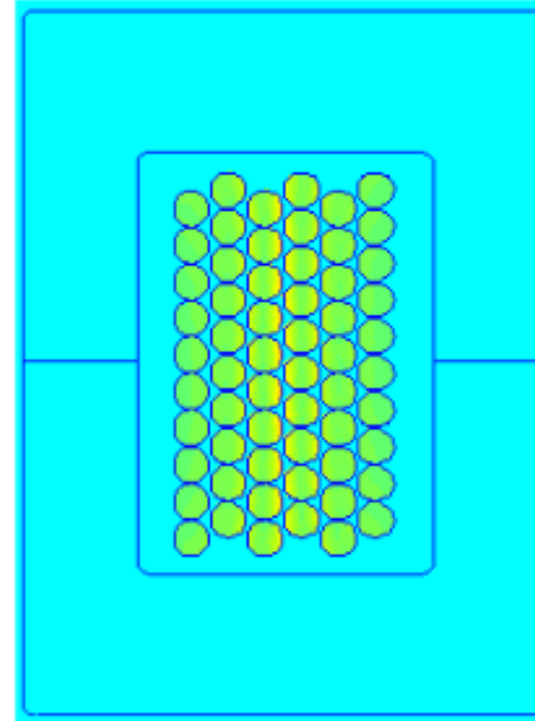


Frequency: 100 kHz

Flux density



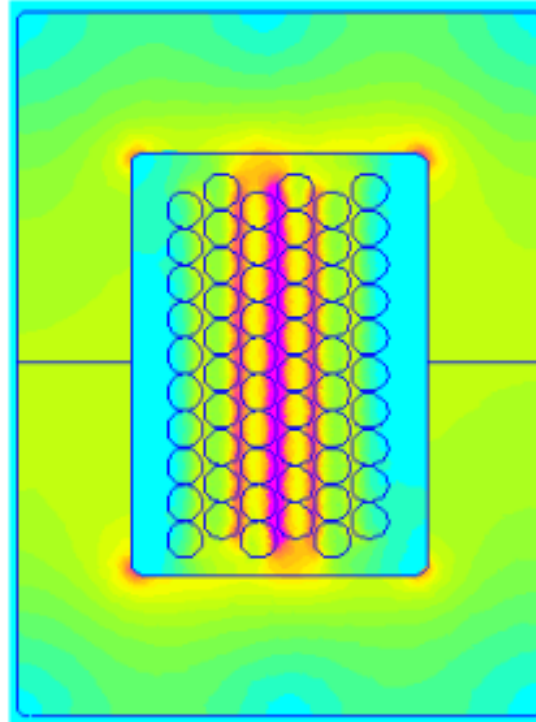
Current Density



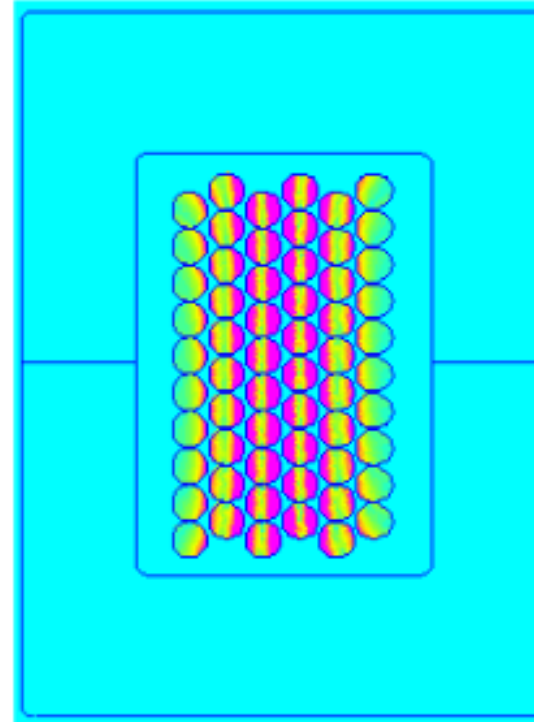
Total copper losses 1.8 larger than at 1 kHz

Frequency: 1 MHz

Flux density



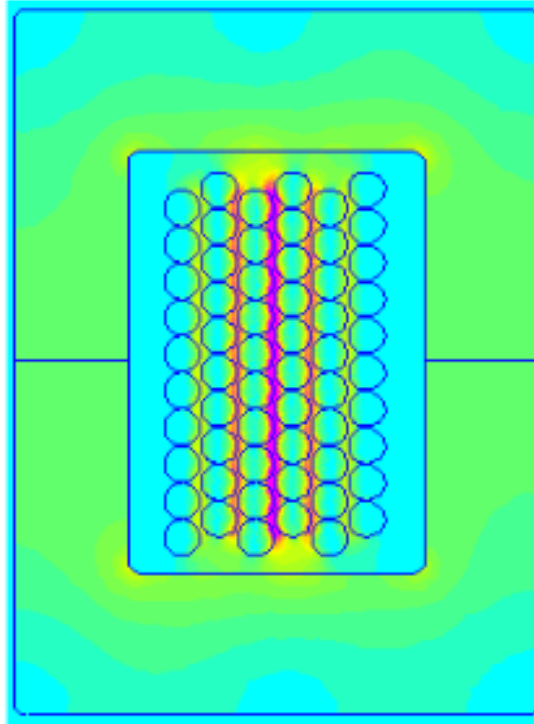
Current Density



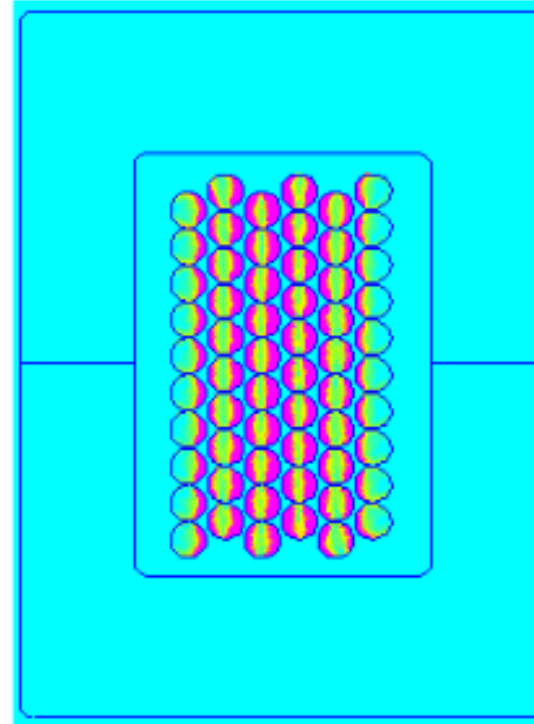
Total copper losses 20 times larger than at 1 kHz

Frequency: 10 MHz

Flux density



Current Density



Very significant proximity effect
Total copper losses = 65 times larger than at 1 KHz