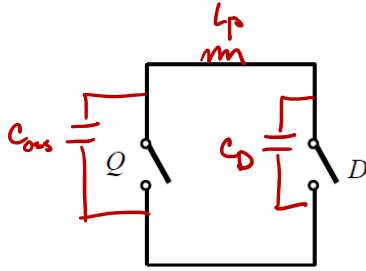
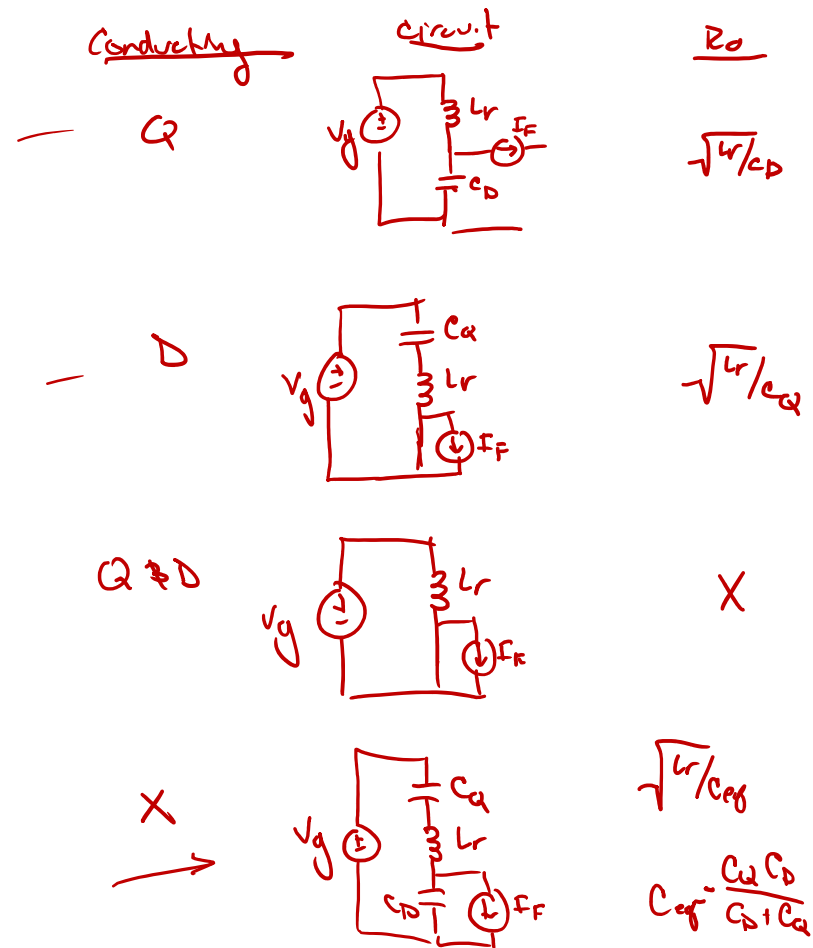
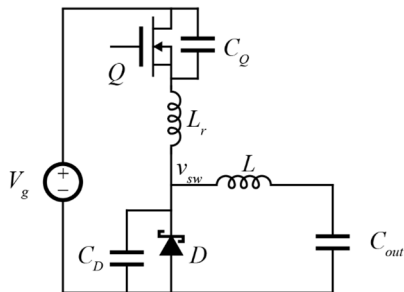


# Wishlist: Multi-Resonant



# ZVS-MR Buck



# Operating Modes

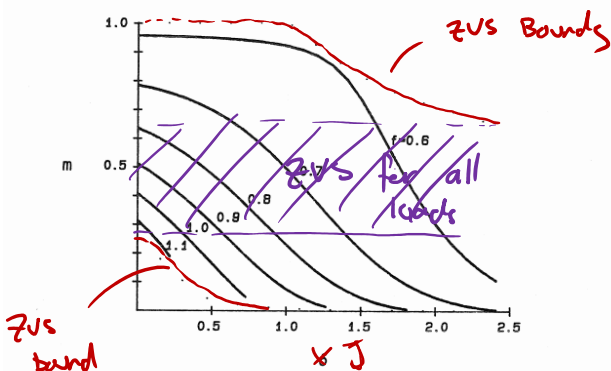


Figure 10.7: Load-to-output DC characteristics of a ZV-MR converter operating in modes (I, II)<sub>1</sub>.

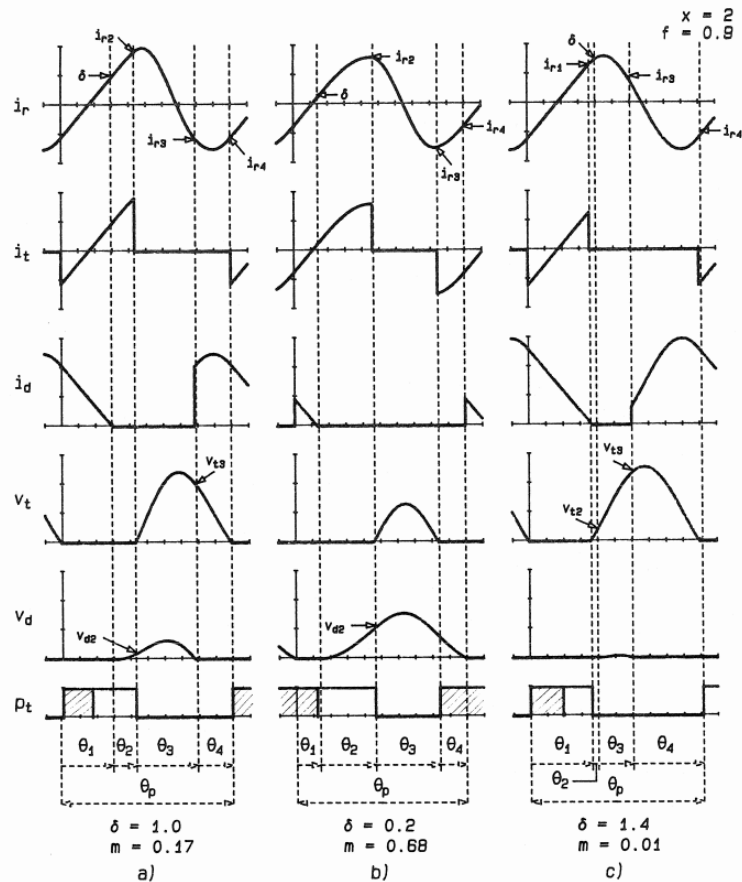
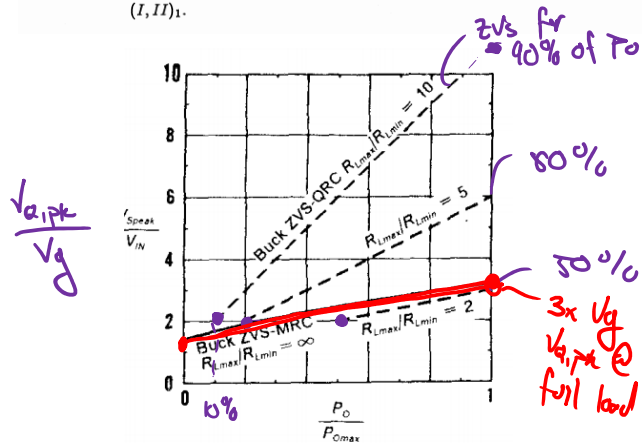
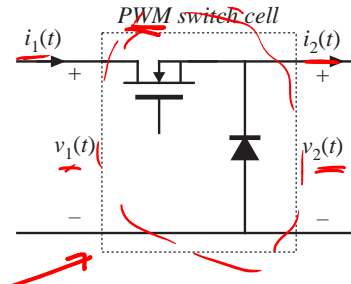
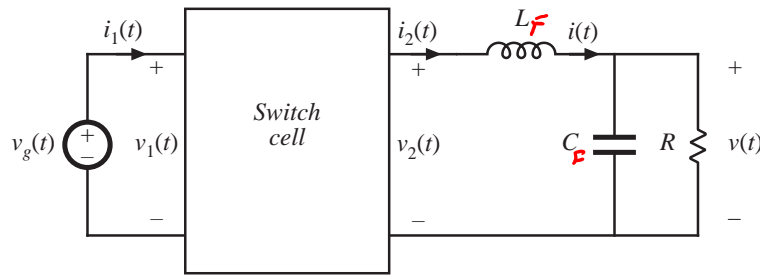


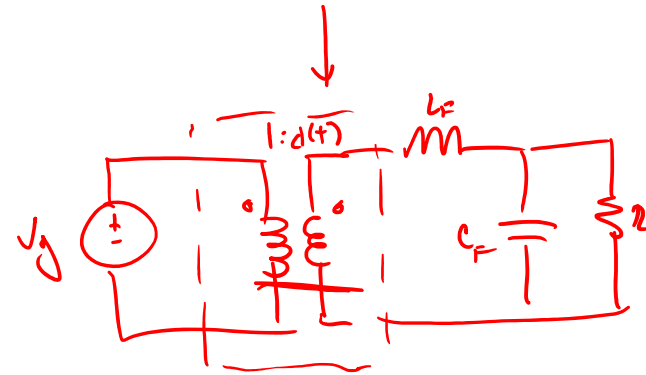
Figure 10.9: Typical waveforms for a ZV-MR converter operating in modes I<sub>1</sub> (a), II<sub>1</sub> (b) or III<sub>1</sub> (c).

# Identification of Resonant Switch



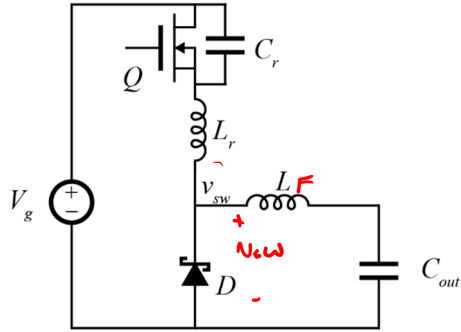
$$\langle v_2 \rangle = D \langle v_1 \rangle$$

$$\langle i_1 \rangle = \underline{D} \langle i_2 \rangle$$



Average, nonlinear model

# Switching Cell Conversion Ratio



$$M = 1 - FR_{1/2} \left( \frac{1}{\beta_c} \right) = f(F, \beta_c)$$

In buck converter

$$M = \frac{\langle v_{out} \rangle}{\langle v_{in} \rangle} = \frac{\langle v_{sw} \rangle}{\langle v_{in} \rangle} = \text{switch cell conversion ratio}$$

$$M = \frac{\langle v_o \rangle}{\langle v_i \rangle} = \frac{\langle v_o \rangle}{\langle v_{in} \rangle}$$

For every other topology

$$M = \text{switch cell conversion ratio} \quad (\neq M \text{ for non-Buck})$$

$M$  for resonant network is equivalent to  $d$  for PWM case

for any other parent topology,

take  $M = f(d)$  from PWM analysis

replace  $d \rightarrow M$

$M = f(M)$  for resonant switch implementation

$$\text{Boost} : M = \frac{1}{1-d}, \text{ PWM}$$

ZVS-QR Boost

$$M = \frac{1}{1-M}, \quad M = 1 - FR_{1/2} \left( \frac{1}{\beta_c} \right)$$

# Conversion Ratios of Various Switch Cells

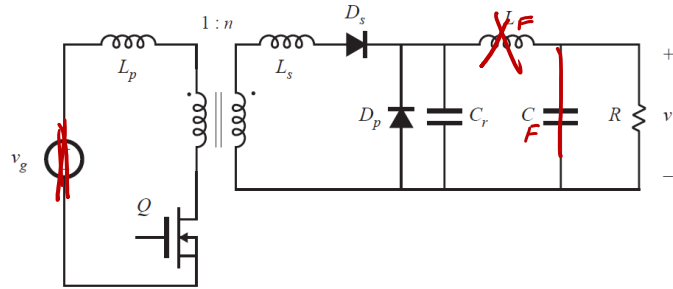
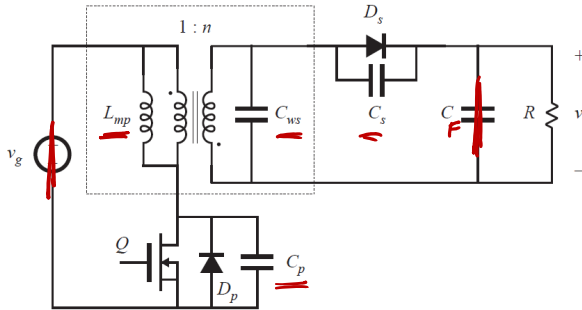
$$\rightarrow P_{1/2}(x) = \frac{1}{2\pi} \left[ \frac{1}{2}x + \pi + \sin^{-1}x + \frac{1}{x} \left( 1 - \sqrt{1-x^2} \right) \right]$$

$$\rightarrow P_1(x) = \frac{1}{2\pi} \left[ \frac{1}{2}x + 2\pi - \sin^{-1}x + \frac{1}{x} \left( 1 - \sqrt{1-x^2} \right) \right] \approx 1$$

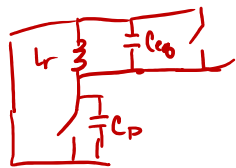
Switch Cell	Conv. Ratio $\mu$	Current Range	Conv. Ratio Range	Requirements on $Q$
PWM	$D$	N/A	$0 \leq \mu \leq 1$	
ZVS-QR (half)	$1 - FP_{\frac{1}{2}}\left(\frac{1}{J_L}\right)$	$1 \leq J_L \leq \infty$	$0 \leq \mu \leq 1$	
ZVS-QR (full)	$1 - FP_1\left(\frac{1}{J_L}\right)$	$1 \leq J_L \leq \infty$	$0 \leq \mu \leq 1$	Bidirectional voltage
ZCS-QR (half)	$FP_{\frac{1}{2}}(J_L)$	$0 \leq J_L \leq 1$	$0 \leq \mu \leq 1$	Unidirectional Current*
ZCS-QR (full)	$FP_1(J_L)$	$0 \leq J_L \leq 1$	$0 \leq \mu \leq 1$	

$\frac{1}{J_L}$

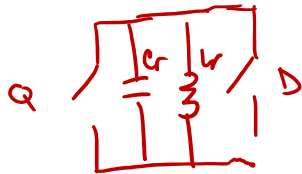
# Resonant Switch Identification Examples



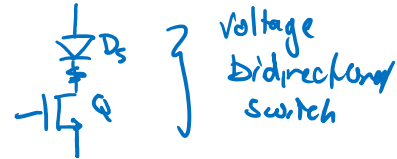
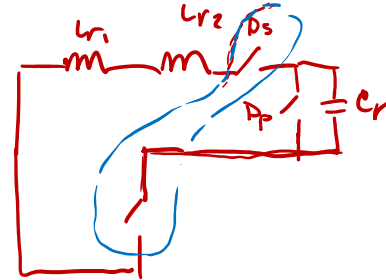
Flyback



$$C_{eq} = n^2(C_s + C_{we})$$

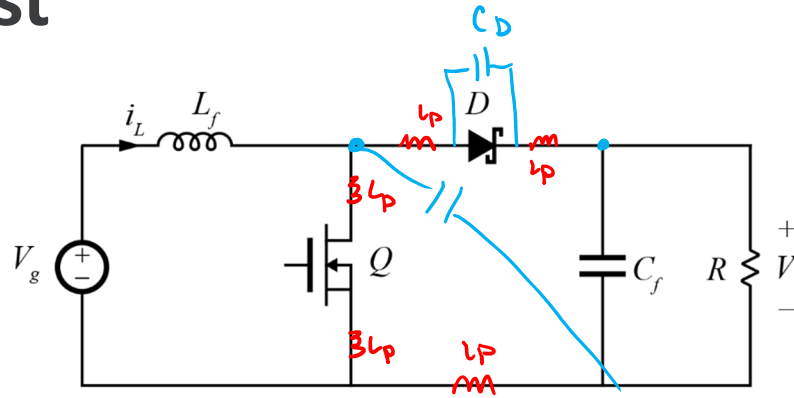


$\therefore$  This is the ZVS-QR flyback



this is ZCS-QR forward converter

# ZCS-QR Boost

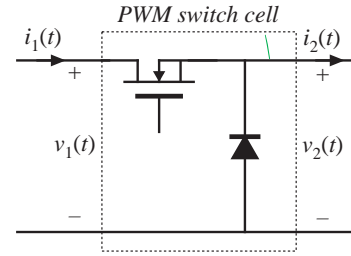
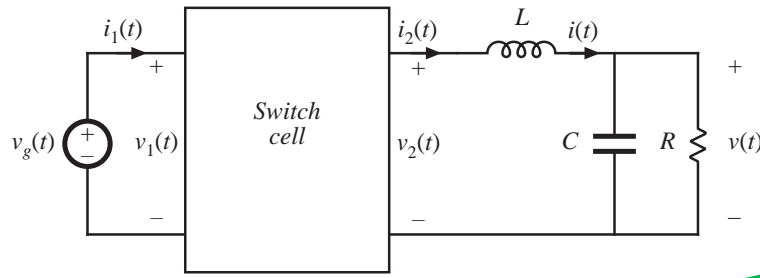


PWM Boost  $\rightarrow M = \frac{1}{1-D(t)}$

ZCS-QR Boost  $\rightarrow m = \frac{1}{1-M(t)}$



# SSM - PWM Parent

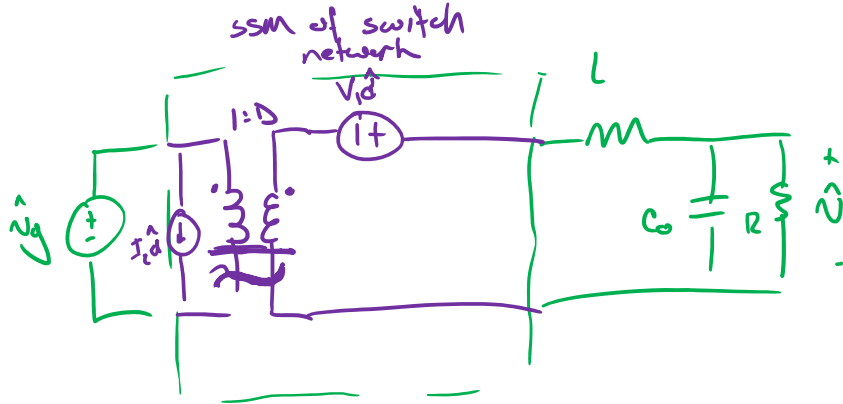


↓ Averaging

$$\begin{cases} \langle v_2 \rangle = d(t) \langle v_1 \rangle \\ \langle i_1 \rangle = d(t) \langle i_2 \rangle \end{cases}$$

↓ Small signal linearization

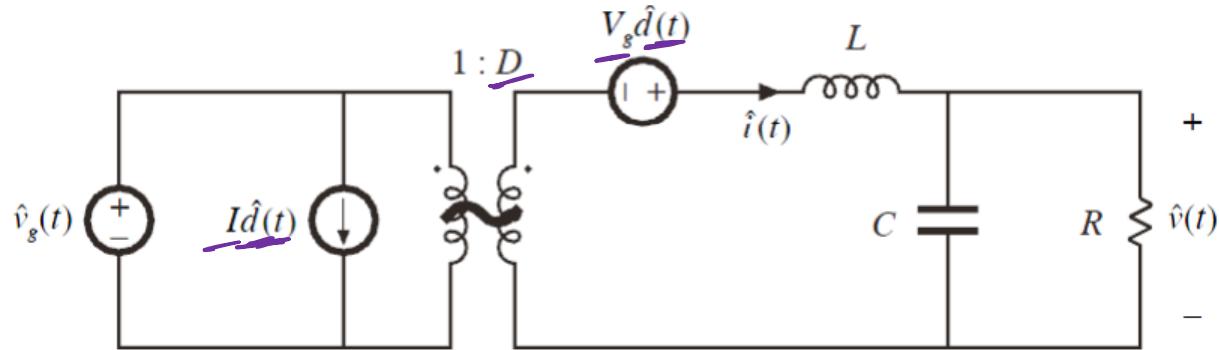
$$\begin{cases} \hat{v}_2 = D \hat{v}_1 + v_1 \hat{d} \\ \hat{i}_1 = D \hat{i}_2 + I_2 \hat{d} \end{cases}$$



equiv circuit

# SSM, PWM Case

Textbook, Fig.7.17(a)



# ZVS-QR Switch Cell SSM

Replace  $d(t)$  with  $u(t)$

$$u = 1 - \frac{F}{2\pi} \left[ \frac{1}{2\zeta_c} + \pi + \sin^{-1}\left(\frac{1}{\zeta_c}\right) + \sqrt{\zeta_c^2 - 1} + \zeta_c \right] = u = f(\zeta_c, F) = f(i_L, v_g, f_s)$$

$$F = \frac{f_s}{f_0}$$

$$\zeta_c = \frac{I_L}{V_g} R_0$$

$$u = f(i_L, v_g, f_s)$$

$$\hat{u} = k_i \hat{i}_L + k_v \hat{v}_g + k_c \hat{f}_s \rightarrow \text{small-signal linearization}$$

# Complete Solution

$$M = 1 - \frac{F}{2\pi} \left[ \frac{\theta_1}{2} + \beta + \mathcal{J}_2 + \mathcal{J}_F \right]$$

$$M = 1 - \frac{F}{2\pi} \left[ \frac{1}{2\mathcal{J}_F} + \pi + \sin^{-1} \left( \frac{1}{\mathcal{J}_F} \right) + \sqrt{\mathcal{J}_F^2 - 1} + \mathcal{J}_F \right]$$

→ same as 20.46 in 2<sup>nd</sup> Ed Fundamentals of Power Elec  
 " 23.46 in 3<sup>rd</sup> Ed

$$M = 1 - F P_{1/2} \left( \frac{1}{\mathcal{J}_L} \right)$$

$$\frac{1}{\mathcal{J}_L} \leftrightarrow \mathcal{J}_L$$

$1 - FP_{1/2}(\mathcal{J}_L) = M$  of ZCS-QR Buck

This is half-wave ZVS-QR Buck



Full-wave ZVS-QR Buck



→ Requires bidirectional voltage blocking Q

