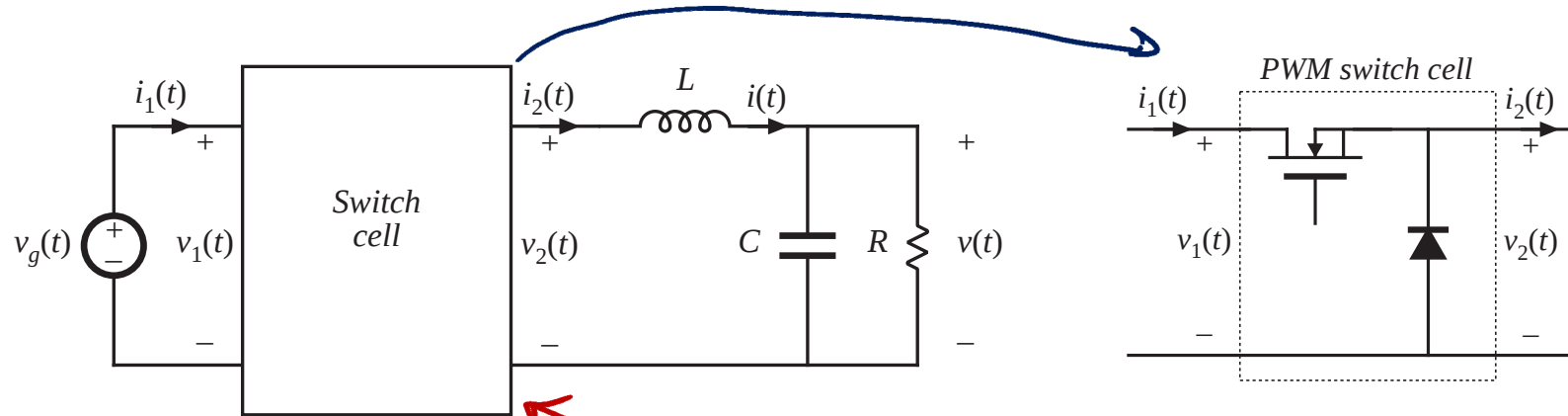
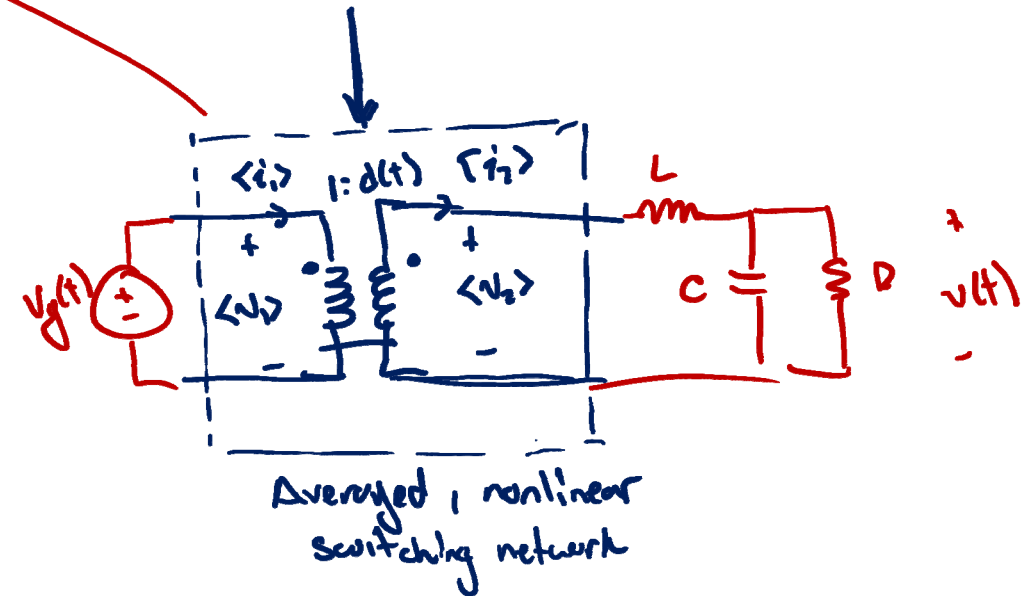


Identification of Resonant Switch

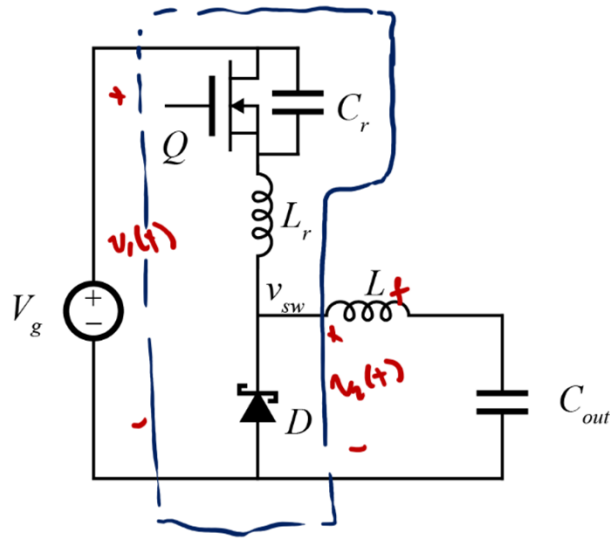


Review Switch Averaging

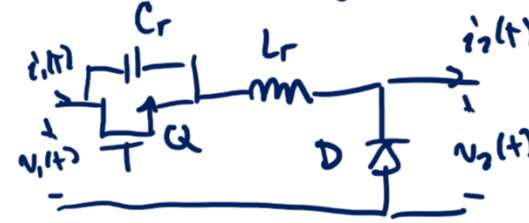
$$\begin{aligned}\langle v_2 \rangle &= \langle v_1 \rangle d(t) \\ \langle i_1 \rangle &= d(t) \langle i_2 \rangle\end{aligned}$$



Switching Cell Conversion Ratio



zvs-QR switching Network



By volt-sec balance on L_f , $\langle v_2(t) \rangle = V$
 $\langle v_2 \rangle = M \langle v_1 \rangle$ ← unique to buck converter
 M solved earlier for zvs-QR Buck converter

Generalize to

$$\langle v_2 \rangle = m(t) \langle v_1 \rangle$$

$$\langle i_1 \rangle = m(t) \langle i_2 \rangle$$

$m(t)$ is the "switch cell conversion ratio"

To get the behavior of any 2-switch PWM converter with a resonant switching network, take
 PWM result & replace $d(t)$ with $m(t)$

E.g. for zvs-QR boost, take $m = \frac{\text{PWM}}{1 - d(t)} \rightarrow M = \frac{1}{1 - m(t)}$

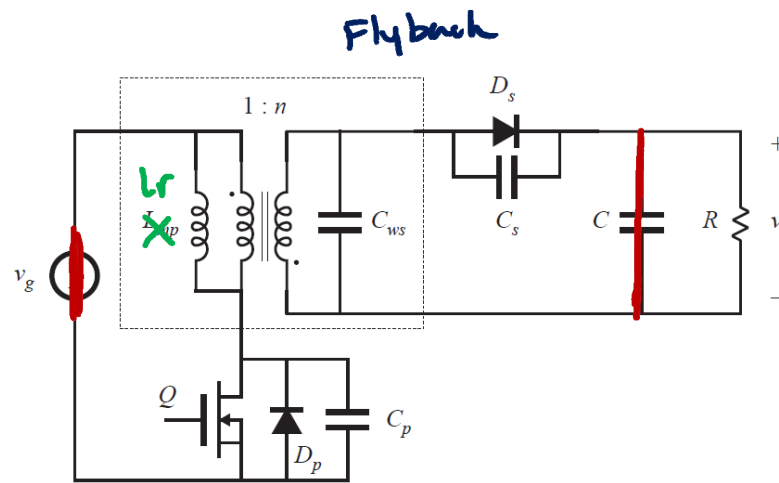
Conversion Ratios of Various Switch Cells

$$P_{1/2}(x) = \frac{1}{2\pi} \left[\frac{1}{2}x + \pi + \sin^{-1}x + \frac{1}{x} \left(1 - \sqrt{1 - x^2} \right) \right]$$

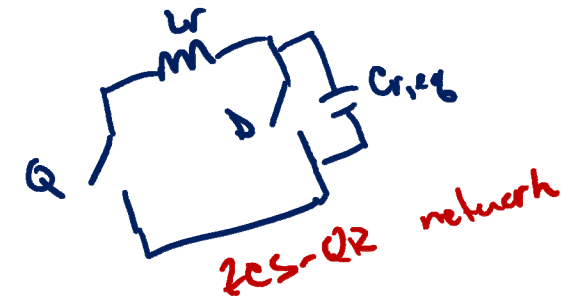
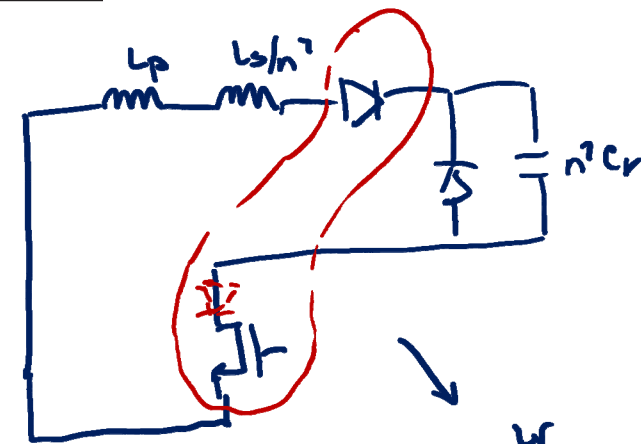
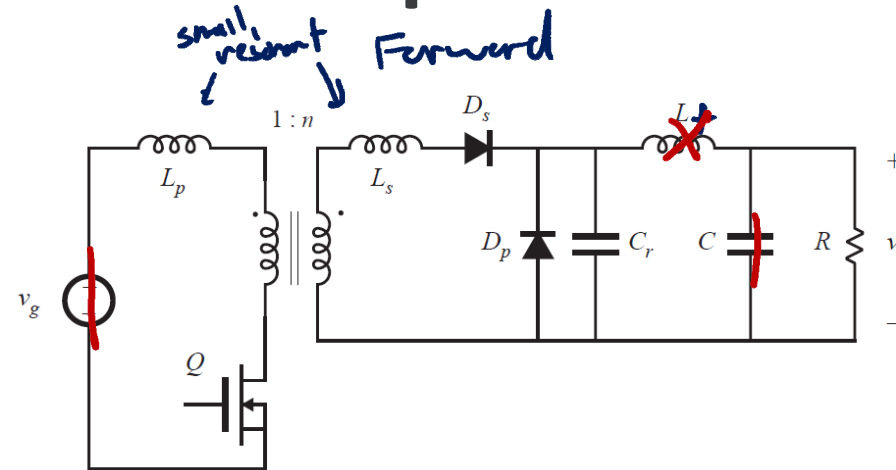
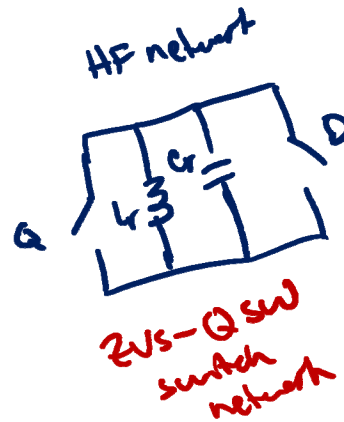
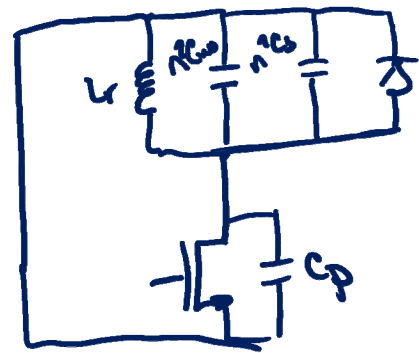
$$P_1(x) = \frac{1}{2\pi} \left[\frac{1}{2}x + 2\pi + \sin^{-1}x + \frac{1}{x} \left(1 - \sqrt{1 - x^2} \right) \right] \approx 1$$

Switch Cell	Conv. Ratio μ	Current Range	Conv. Ratio Range	Requirements on Q
PWM	D	N/A	$0 \leq \mu \leq 1$	
ZVS-QR (half)	$1 - FP_{\frac{1}{2}}\left(\frac{1}{J_L}\right)$	$1 \leq J_L \leq \infty$	$0 \leq \mu \leq 1$	
ZVS-QR (full)	$1 - FP_1\left(\frac{1}{J_L}\right)$	$1 \leq J_L \leq \infty$	$0 \leq \mu \leq 1$	Bidirectional voltage ✓
ZCS-QR (half)	$FP_{\frac{1}{2}}(J_L)$	$0 \leq J_L \leq 1$	$0 \leq \mu \leq 1$	Unidirectional Current* ✗
ZCS-QR (full)	$FP_1(J_L)$	$0 \leq J_L \leq 1$	$0 \leq \mu \leq 1$	

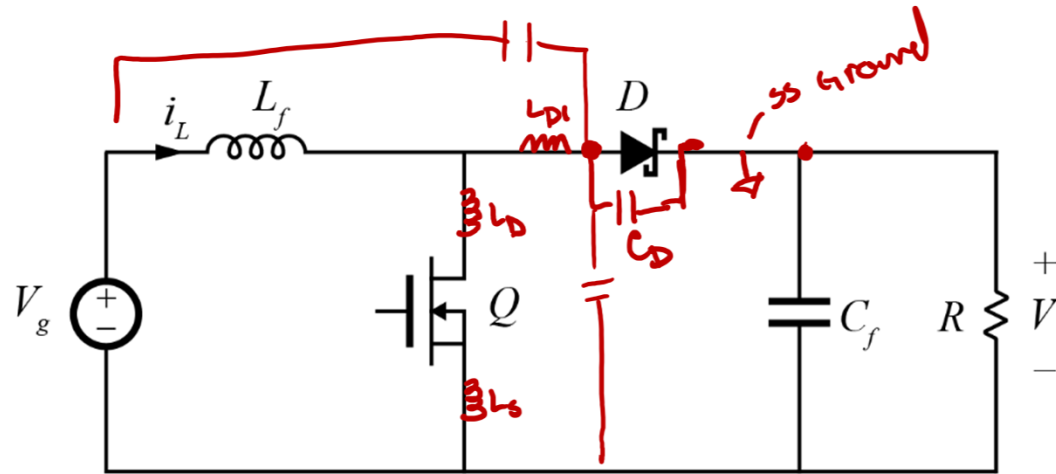
Resonant Switch Identification Examples



HF switching network



ZCS-QR Boost

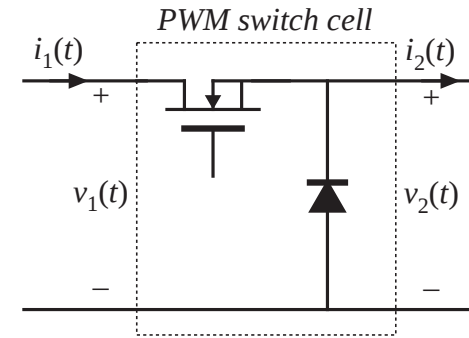
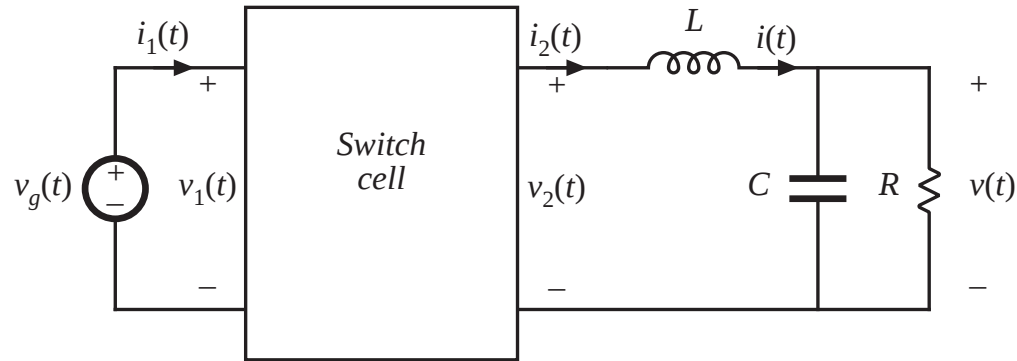


pwm solution to boost $M = \frac{1}{D} = \frac{1}{1-d(t)}$

ZCS-QR solution $M = \frac{1}{1-u(t)}$

$u(t) = \text{FP}_{1/2}(\omega)$ for half-wave operation

SSM - PWM Parent

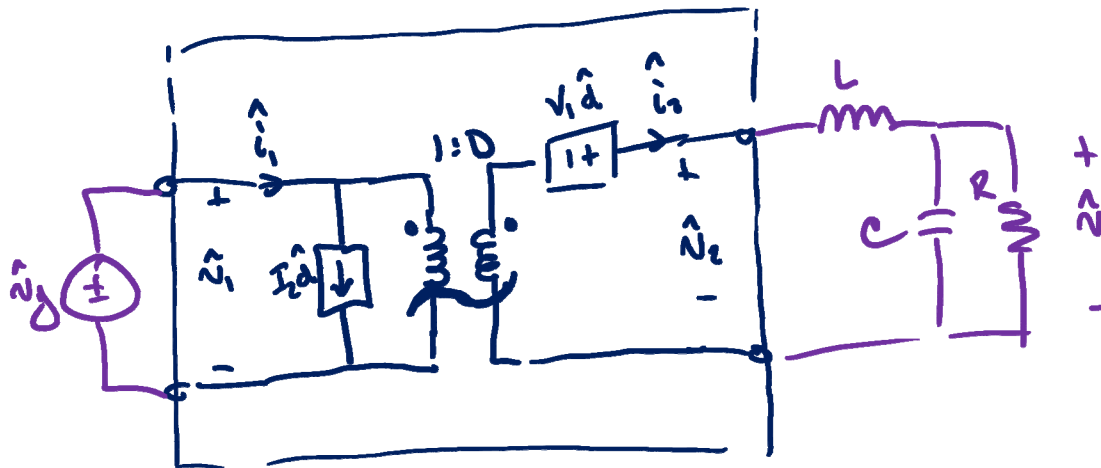


Review $\rightarrow$ SSM linearization

$$\begin{aligned} \langle v_2 \rangle &= d(t) \langle v_1 \rangle \\ \langle i_1 \rangle &= d(t) \langle i_2 \rangle \end{aligned} \quad \left. \vphantom{\begin{aligned} \langle v_2 \rangle &= d(t) \langle v_1 \rangle \\ \langle i_1 \rangle &= d(t) \langle i_2 \rangle \end{aligned}} \right\} \text{Averaged, nonlinear}$$

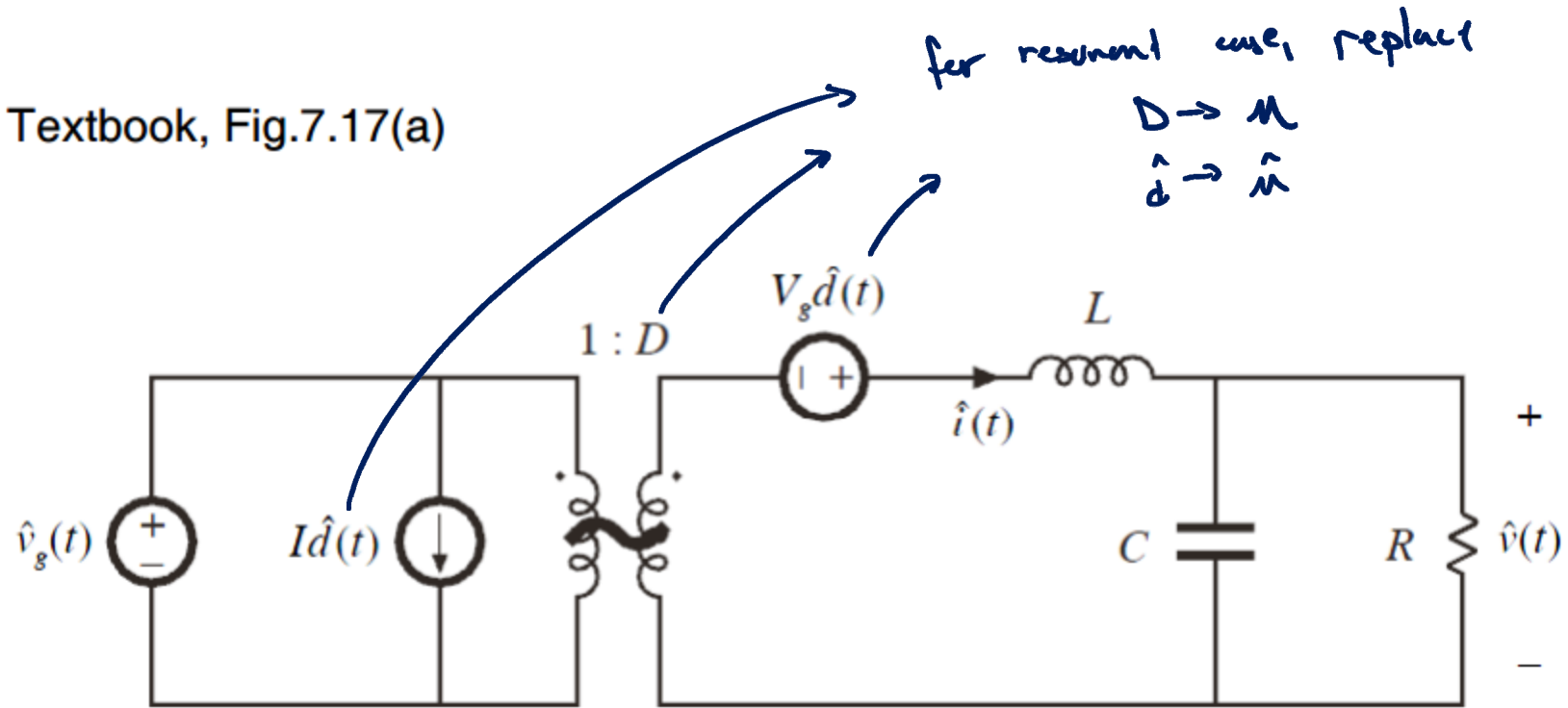
$\downarrow$ SSM linearization

$$\begin{aligned} \hat{v}_2 &= D \hat{v}_1 + \hat{d} V_1 \\ \hat{i}_1 &= \underbrace{D \hat{i}_2}_{\text{XF}} + \underbrace{\hat{d} I_2}_{\text{controlled source}} \end{aligned} \quad \left. \vphantom{\begin{aligned} \hat{v}_2 &= D \hat{v}_1 + \hat{d} V_1 \\ \hat{i}_1 &= D \hat{i}_2 + \hat{d} I_2 \end{aligned}} \right\} \text{Averaged, linear}$$



SSM, PWM Case

Textbook, Fig.7.17(a)



ZVS-QR Switch Cell SSM

$$\mu = 1 - F \cdot \frac{1}{2\pi} \left[\frac{1}{2\lambda} + \pi + \sin^{-1}\left(\frac{1}{\lambda}\right) + \sqrt{\lambda^2 - 1} + \lambda \right] = 1 - F \cdot P_{42}\left(\frac{1}{\lambda}\right)$$
$$F = \frac{f_s}{f_o} \quad \lambda = \frac{I_L R_o}{V_g}$$

$$\mu = f(I_L, V_g, f_s)$$