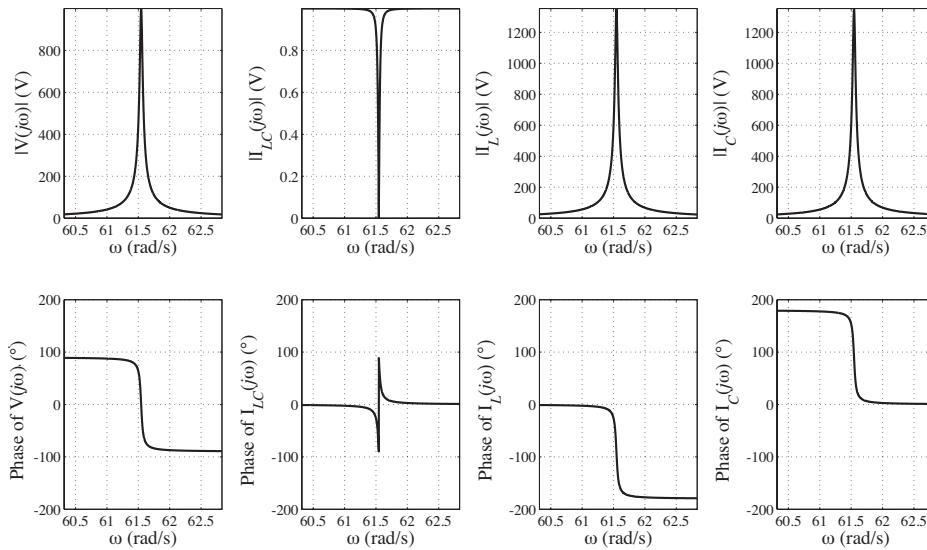


Chapter 16 Answers to Assigned Homework

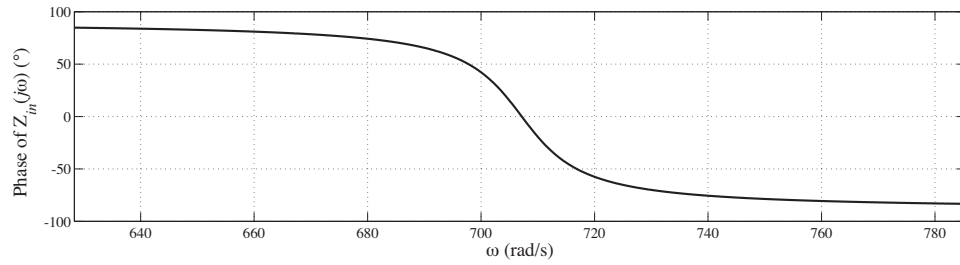
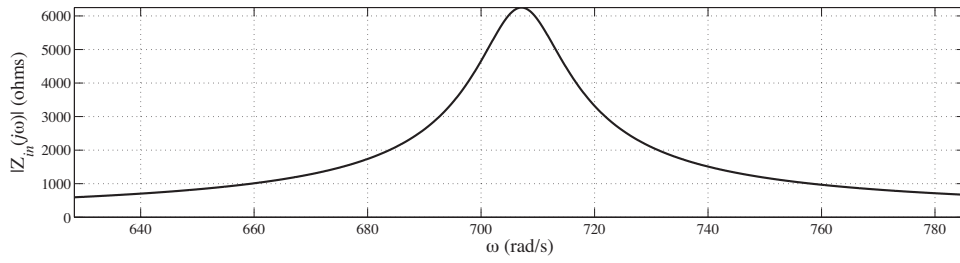
2. (a) $\alpha = 0.0227 \text{ s}^{-1}$ $\omega_0 = 61.5457 \text{ s}^{-1}$
 $\zeta = 3.69 \times 10^{-4}$ $f_0 = 9.7953 \text{ Hz}$
 $\omega_d = 61.5457$

(b)
$$I_{LC} = I_L + I_C = \frac{s^2 + 1/LC}{s^2 + s/RC + 1/LC}$$

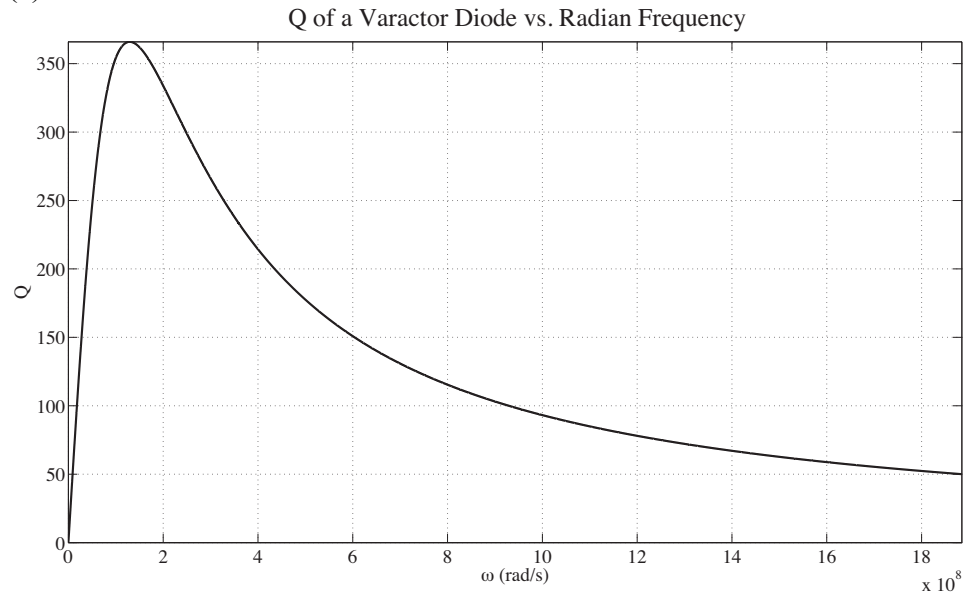


- (c) At resonance, $I_L = -I_C$.
5.
$$Z_{in}(j\omega) = 2 \frac{-\omega^2 + j5 \times 10^4 \omega + 2.5 \times 10^5}{-\omega^2 + j16\omega + 5 \times 10^5}$$

Close examination of a graph of the magnitude of the input impedance reveals that the frequency of maximum magnitude is 707.1. The same as the ideal to 4 significant figures.



9. (a)



$$(b) \quad \frac{dQ}{d\omega} = C_j R_p \frac{1 - \omega^2 R_s R_p C_j^2}{(1 + \omega^2 C_j^2 R_p R_s)^2}$$

$$\omega = 1.2943 \times 10^8 \text{ rad/s}$$

At that frequency $Q = 365.96$.

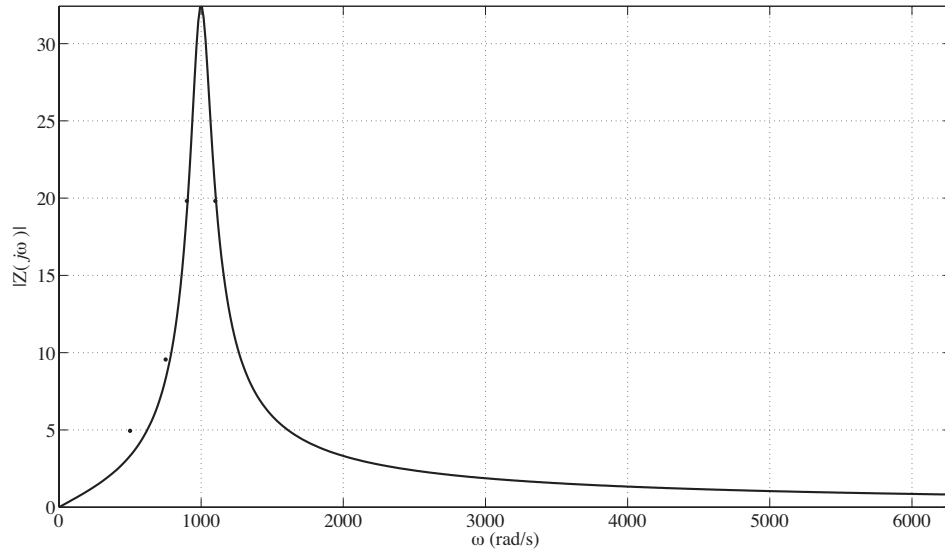
11. (a) 500 rad/s $Z(j\omega) = 4.94 \angle 81.25^\circ$

(b) 750 rad/s $Z(j\omega) = 9.558 \angle 72.9^\circ$

(c) 900 rad/s $Z(j\omega) = 19.816 \angle 52.4^\circ$

(d) 1100 rad/s $Z(j\omega) = 19.816 \angle -52.4^\circ$

(e)



14. (a) 2.5 ohms.

(b) $C = 1 \text{ F}$

(c) $Q_0 = 2.5$

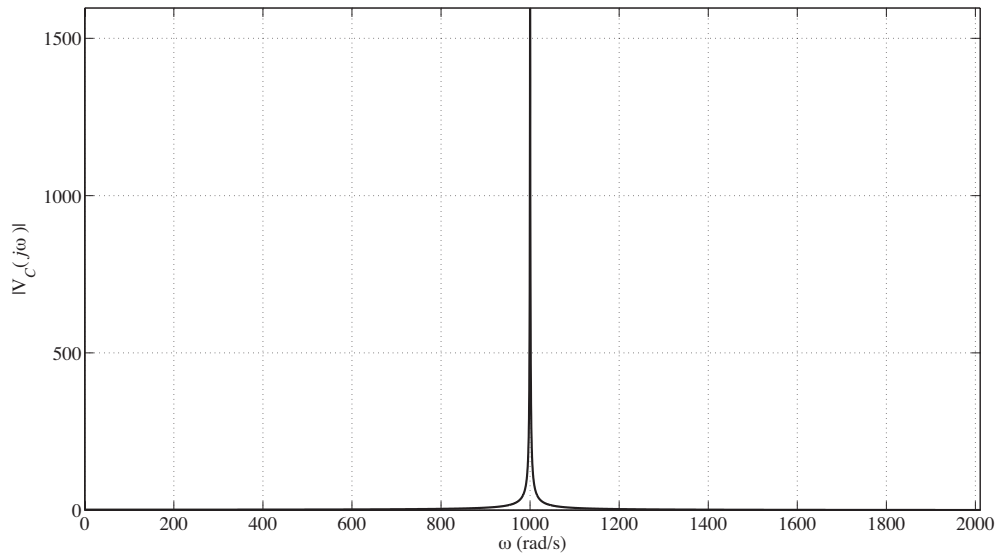
$$B = 0.4 \Rightarrow \omega_1 = 0.8198 \text{ and } \omega_2 = 1.2198$$

15. (a) $[Z_{RLC}]_{\omega=500 \times 10^6} = 750,000 \angle 89.99^\circ$

(b) $I = 3.92 \angle -89.99^\circ \mu\text{A}$

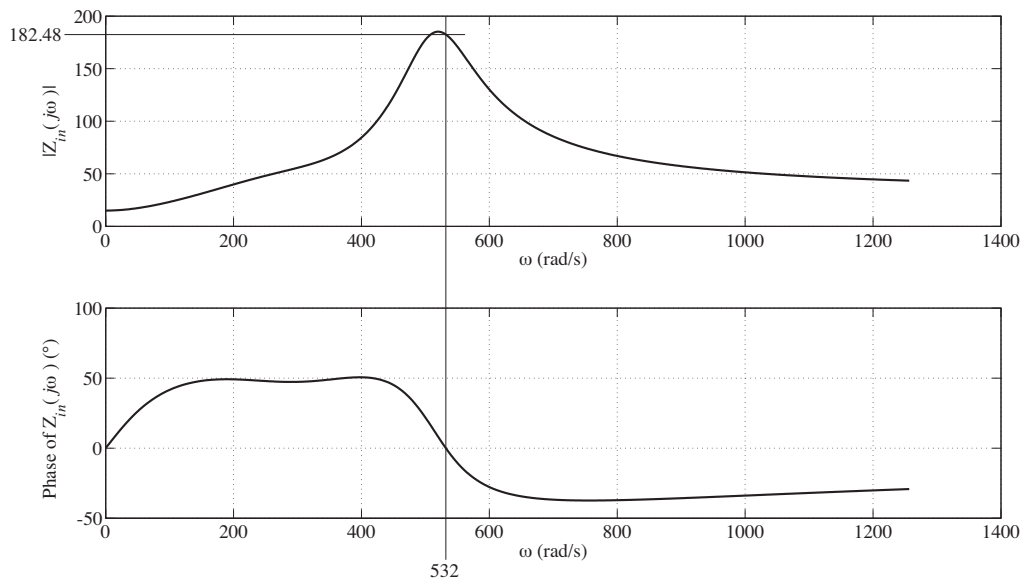
$$i(t) = 3.92 \cos(425 \times 10^6 t - 89.99^\circ) \mu\text{A}$$

18. $V_C = 4 \times 10^6 I_1 / j\omega = \frac{6 \times 10^6}{-4\omega^2 + j3.75\omega + 4 \times 10^6}$



20. (a) At $\omega = 700$ rad/s, $Z_{in} = 61.56 \angle 0.044^\circ$.
 At $\omega = 800$ rad/s, $Z_{in} = 62.03 \angle -0.0535^\circ$.

23.
$$Z_{in}(j\omega) = 32 \frac{\omega^4 - j1248\omega^3 - 548125\omega^2 + j1.7959 \times 10^8 \omega + 1.25 \times 10^{10}}{\omega^4 - j503.3s^3\omega - 4.16 \times 10^5 \omega^2 + j1.12 \times 10^8 \omega + 2.667 \times 10^{10}}$$



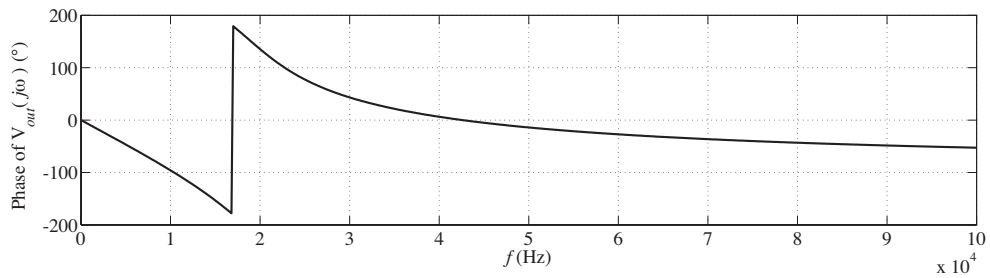
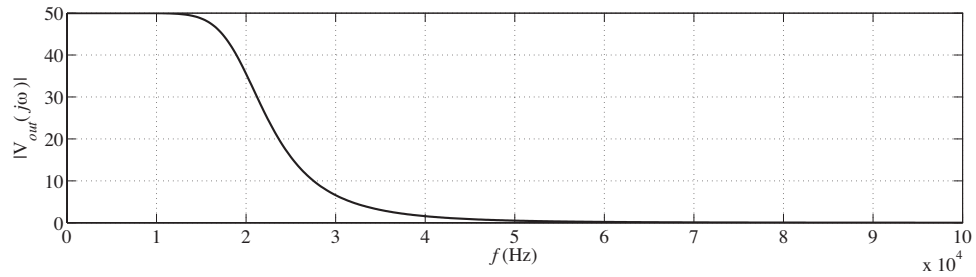
25. (a) $R = 1\Omega$ $L = 265.3\text{nH}$ $C = 2.39\mu\text{F}$
 (b) $R = 500\text{k}\Omega$ $L = 166.67\text{kH}$ $C = 6\mu\text{F}$
 (c) $R = 25\Omega$ $L = 1.768\mu\text{H}$ $C = 25.46\text{nF}$

26. (a) $R = 1\Omega$ $L = 74.03\mu\text{H}$ $C = 1.85\text{mF}$
 (b) $R = 100\Omega$ $L = 20\text{H}$ $C = 50\text{mF}$
 (c) $R = 15\text{k}\Omega$ $L = 6.367\text{mH}$ $C = 707.4\text{pF}$

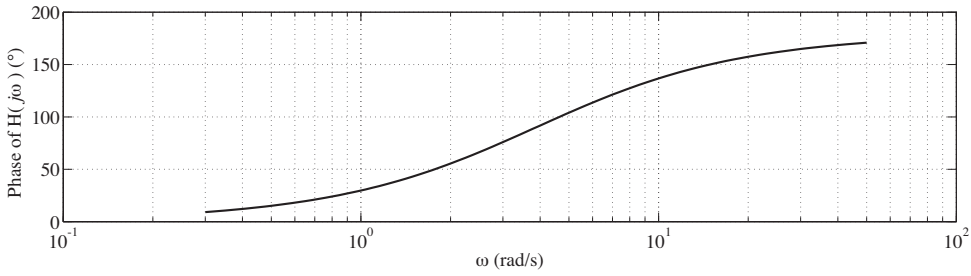
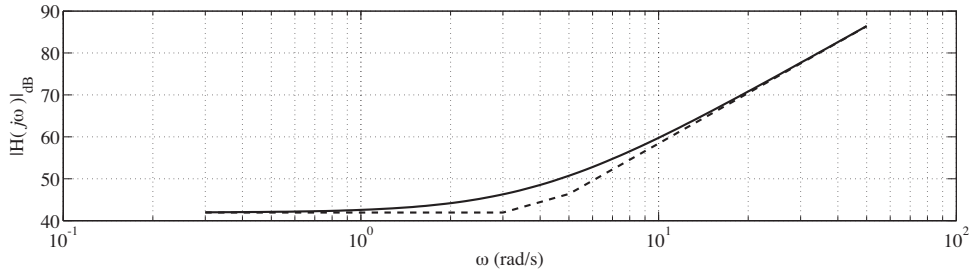
28. (a)
- | | | |
|---------------------|-------------------|---------------------|
| $245.5 \mu\text{H}$ | $795 \mu\text{H}$ | $245.5 \mu\text{H}$ |
| 257 nF | 257 nF | |

(b)

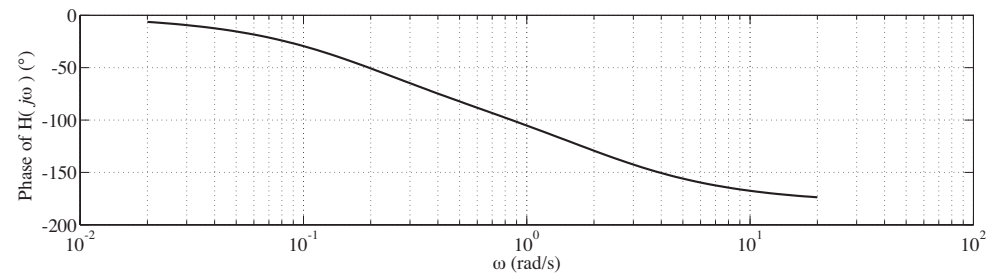
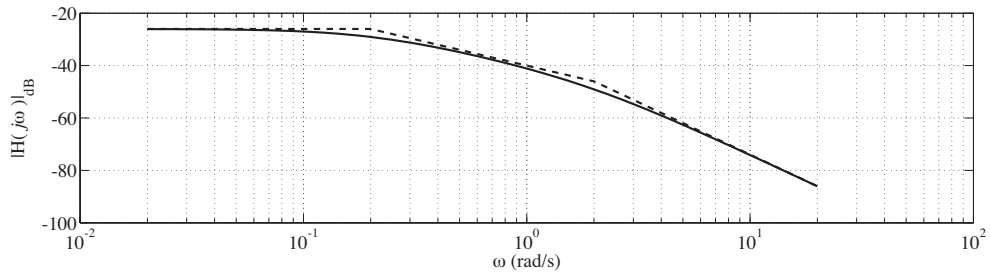
$$V_{out} = 50I_3 = \frac{1.58 \times 10^{27}}{s^5 + 4.073 \times 10^5 s^4 + 8.296 \times 10^{10} s^3 + 1.044 \times 10^{16} s^2 + 8.121 \times 10^{20} s + 3.16 \times 10^{25}}$$



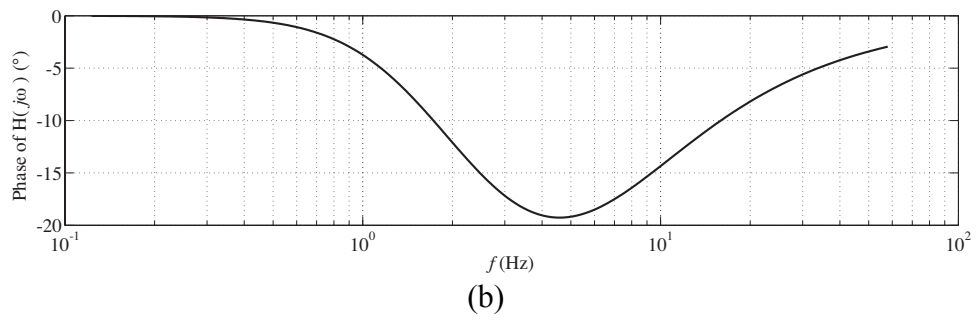
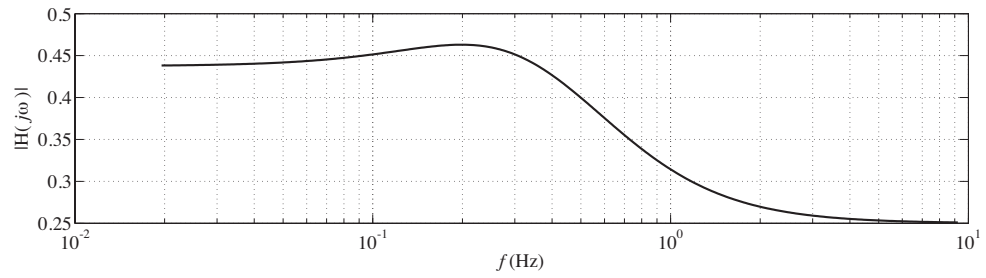
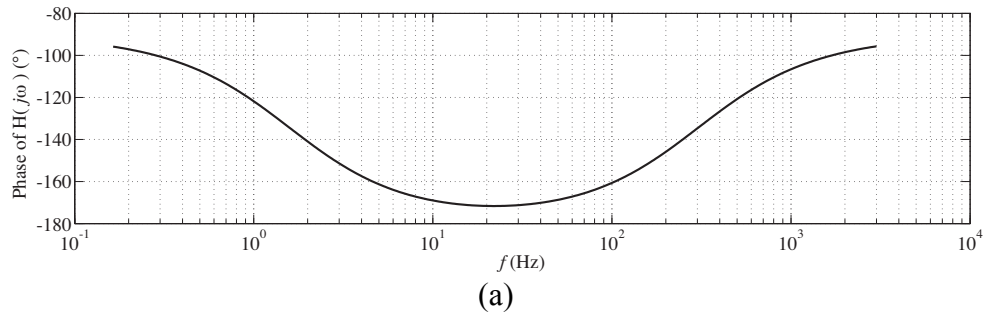
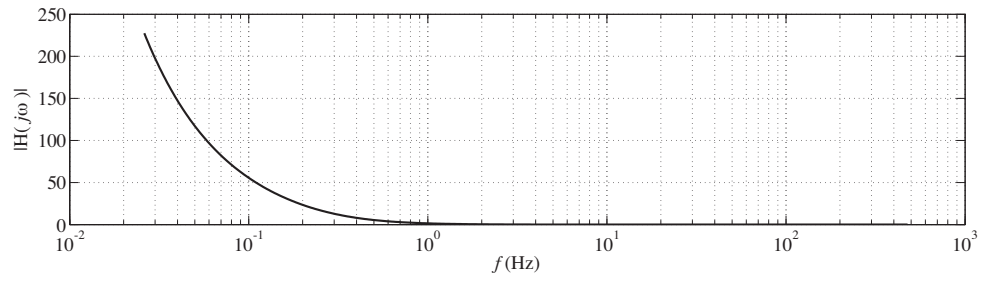
31.

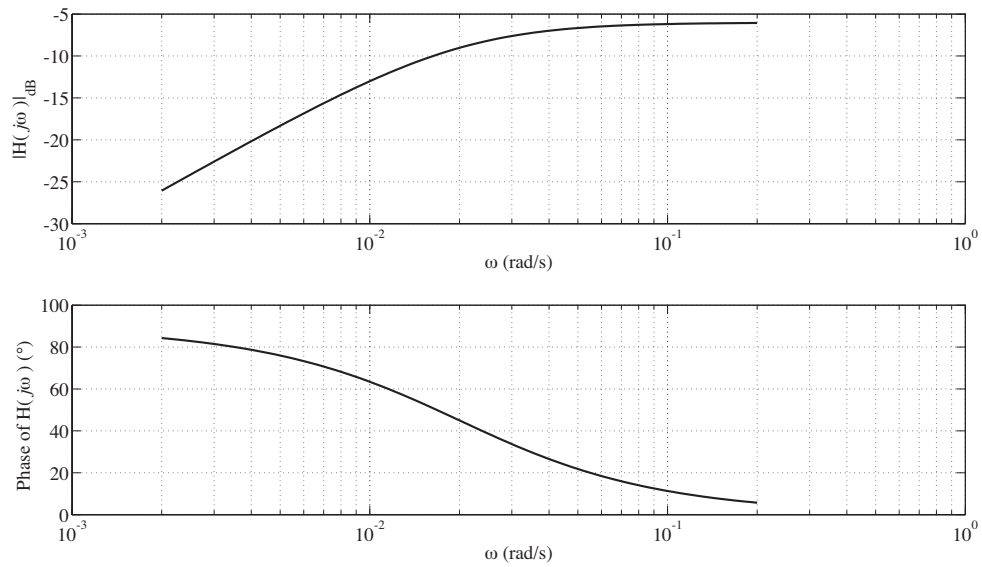


(a)

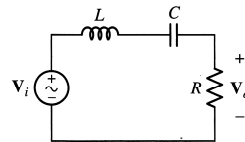


(b)





42.



■ FIGURE 16.39 A simple bandpass filter.

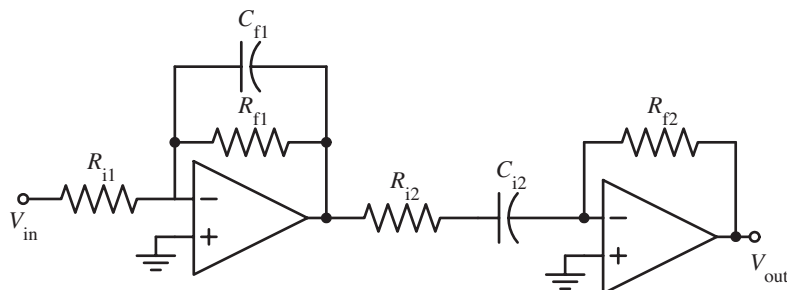
$$\omega_L = 500\pi \quad C = 2.476 \mu\text{F}$$

44.

$$H(j\omega) = \frac{1/LC - \omega^2}{1/LC - \omega^2 + j\omega R/L}$$

Choosing C arbitrarily to be $100\mu\text{F}$, we get $L = 70.36 \text{ mH}$.

47.



$$H(s) = \frac{R_{f2}}{R_{i1}C_{f1}R_{i2}} \frac{s}{(s + 1/R_{f1}C_{f1})(s + 1/R_{i2}C_{i2})}$$

Pick R_{i2} arbitrarily to be 1000 ohms. Then

$$C_{i2} = 10 \mu\text{F}$$

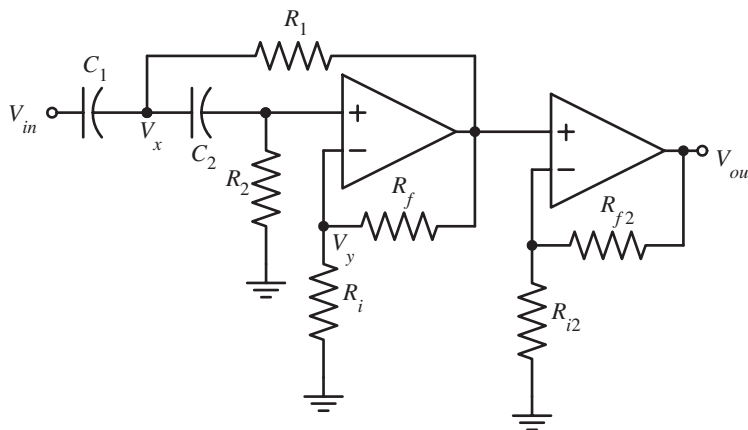
Pick R_{f2} so that the high-frequency gain is $\sqrt{10}$. Then $R_{f2} = 3162 \Omega$.

Set the high corner frequency of the lowpass filter to 1100 rad/s. Pick $R_{f1} = 1000 \Omega$. Then

$$C_{f1} = 909.1 \text{ nF}$$

Pick R_{i1} so that the low-frequency gain is $\sqrt{10}$. Then $R_{i1} = 316.2 \Omega$.

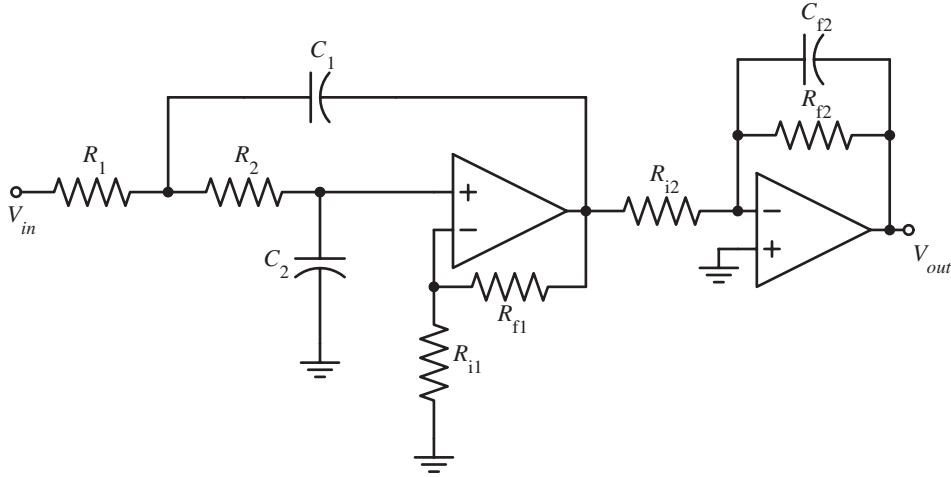
54. This can be done with one Sallen-Key highpass filter stage followed by an amplifier to set the overall gain.



$$H(s) = \frac{V_{out}(s)}{V_{in}(s)} = K \frac{s^2}{s^2 + s \left[\frac{C_1 + C_2}{R_2 C_1 C_2} + \frac{1 - K}{R_1 C_1} \right] + \frac{1}{R_1 R_2 C_1 C_2}}$$

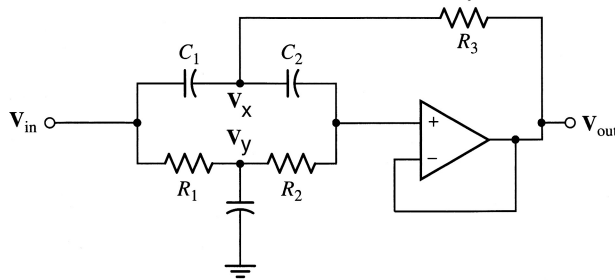
Let $R_1 = R_2 = R$ and $C_1 = C_2 = C$. Then $K = 1.5858$. Let $R = 10^7$ and $C = 10^{-7}$. Frequency scale to make $\omega_c = 4000\pi$. We can do this by reducing R by a factor of 4000π to become 795.7 ohms. Let $R_{i2} = 1000\Omega$. Then $R_{f2} = 60\Omega$.

56. This can be designed using a Sallen-Key stage followed by a single-pole stage.



Let $R_1 = R_2 = R$ and $C_1 = C_2 = C$. Then $K = 2$. Also $RC = 1$. Let $R = 10^7 \Omega$ and $C = 10^{-7} \text{ F}$. Then in the second stage let $R_{f2} = 10^7 \Omega$. Then $C_{f2} = 10^{-7} \text{ F}$ or 100nF. Set $R_{i2} = 4.478\text{M}\Omega$. Now, scale to the correct corner frequency of 1800 Hz or 11310 rad/s. Then $R = 884.2 \Omega$, $R_{f2} = 884.2 \Omega$ and $R_{i2} = 396 \Omega$.

60.

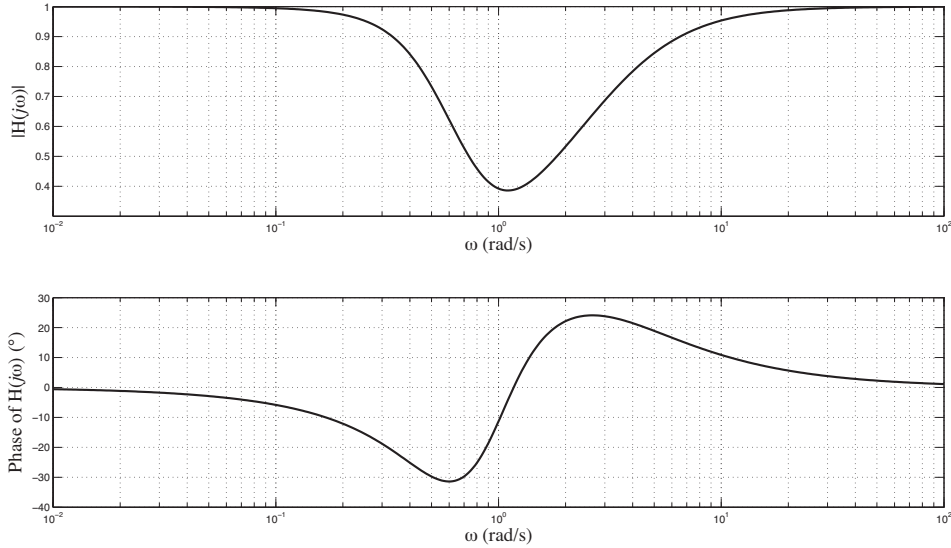


■ FIGURE 16.61

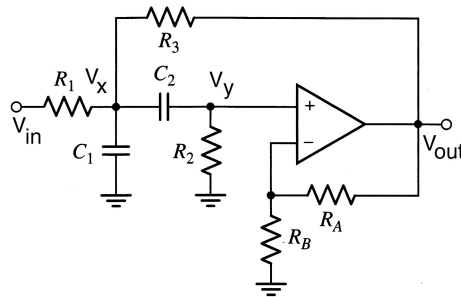
$$\frac{V_{out}}{V_{in}} = \frac{s^3 + s^2 \frac{G_1 + G_2}{C_3} + s \frac{G_1 G_2}{C_3} \frac{C_1 + C_2}{C_1 C_2} + \frac{G_1 G_2 G_3}{C_1 C_2 C_3}}{s^3 + s^2 \left[G_2 \frac{C_1 + C_2}{C_1 C_2} + \frac{G_1 + G_2}{C_3} \right] + s G_2 \left[\frac{G_3}{C_1 C_2} + \frac{G_1}{C_3} \frac{C_1 + C_2}{C_1 C_2} \right] + \frac{G_1 G_2 G_3}{C_1 C_2 C_3}}$$

No numerical component values were given. Let all resistors be 1 ohm and all capacitors be 1 F. Then

$$V_{out} = \frac{s^3 + 2s^2 + 2s + 1}{s^3 + 4s^2 + 3s + 1}$$



63.



■ FIGURE 16.62

$$\frac{V_{out}}{V_{in}} = \frac{KG_1}{C_1} \frac{s}{s^2 + s \left[\frac{G_1 + G_2}{C_1} + \frac{G_2}{C_2} + \frac{G_3}{C_1} (1 - K) \right] + \frac{G_1 + G_3}{C_1 C_2} G_2}$$

No numerical component values were given. Let all resistors be 1 ohm and all capacitors be 1 F. Then

$$\frac{V_{out}}{V_{in}} = \frac{2s}{s^2 + 2s + 2}$$

